## General Description

The ADT7350 is a step-down converter with integrated switching MOSFET. It operates wide input supply voltage range from 4.5 V to 24 V with 1.2A peak output current. It includes current limiting protection and thermal shutdown.
It reduces design complexity and external component count. The ADT7350 is available in


Package outline of the ADT7350 small outline SOT23-6L package.

## Features

- Current mode buck regulator with 1.4 MHz fixed frequency
- Input voltage range : 4.5 V to 24 V
- Adjustable output range : 0.81 V to 15 V


## Applications

- Distributed Power Systems
- Surveillance Camera module
- Navigation
- Infra-Red LED Drivers


## Typical Application Circuit



[^0]
## Part List

| Component | Type | Value (Model) | Manufacturer |
| :---: | :--- | :--- | :--- |
| U1 | IC | ADT7350 | ADTech |
| D1 | Schottky barrier diode | RSX101M-30 | ROHM |
| L1 | Chip inductor | $4.7 \mathrm{uH} / 1.6 \mathrm{~A}$ | TDK |
| C2 | MLCC | $22 \mu \mathrm{~F} / 16 \mathrm{~V}$ | - |
| C1 | MLCC | $10 \mu \mathrm{~F} / 50 \mathrm{~V}$ | - |
| CB | MLCC | 10 nF | - |
| R1 | Chip resistor | $49.9 \mathrm{k} \Omega / 1 \%$ | - |
| R2 | Chip resistor | $16.2 \mathrm{k} \Omega / 1 \%$ | - |

-R1,R2 : VOUT 3.3V typical case

## Pin Description

| Pin <br> No. | Name | Pin Description |
| :---: | :---: | :--- |
| 1 | BST | Bootstrap. This pin acts as the power supply for internal Power Transistor driver. <br> Connect a 10nF capacitor between this pin and SW pin. |
| 2 | GND | Ground and heat sink. Connect this pin to a large copper PCB area. |
| 3 | FB | Feedback voltage input. The voltage of this pin is regulated to 0.81V. The shielding in the <br> PCB layout is option and connect resistor divider as close as possible. |
| 4 | EN | Chip enable input. The Enable voltage range is higher than 1.2 V and disable voltage is <br> lower than 0.4V. |
| 5 | VIN | Power supply input. The input bypass capacitor should be as close as GND pin. |
| 6 | SW | Switching node. The free-wheeling diode is connected between this pin and GND. |



[^1]Functional Block Diagram


## Absolute Maximum Ratings (Note1)

| Parameter | Symbol | Min. | Typ. | Max. | Unit |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Power supply voltage | $\mathrm{V}_{\mathrm{IN}}$ | - | - | 26 | V |
| SW pin voltage | $\mathrm{V}_{\mathrm{SW}}$ | - | - | 27 | V |
| BST pin voltage | $\mathrm{V}_{\mathrm{BST}}$ | - | - | $\mathrm{V}_{\mathrm{SW}}+6 \mathrm{~V}$ | V |
| All Other Pins | - | -0.3 | - | +6 | V |
| Max. power dissipation (Ta=25 ${ }^{\circ} \mathrm{C}$ ) (Note2) | $\mathrm{P}_{\mathrm{D}}$ | - | - | 400 | mW |
| Thermal resistance (Note3) | $\Theta_{\mathrm{JA}}$ | - | 250 | - | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| Storage temperature | $\mathrm{T}_{\mathrm{STG}}$ | -65 | - | +150 | ${ }^{\circ} \mathrm{C}$ |
| Junction temperature | $\mathrm{T}_{\mathrm{Jmax}}$ | - | - | +150 | ${ }^{\circ} \mathrm{C}$ |

Note1. Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. Note2. derate $4.0 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $+25^{\circ} \mathrm{C}$. This is recommended to operate under this power dissipation specification.
Note3. Measured on JESD51-7, 4-layer PCB

## Operating Ratings

| Parameter | Symbol | Min. | Typ. | Max. | Unit |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Power supply voltage | $\mathrm{V}_{\text {IN }}$ | 4.5 | 12.0 | 24.0 | V |
| Output voltage | $\mathrm{V}_{\mathrm{OUT}}$ | 0.81 | - | 15 | V |
| Operating temperature | $\mathrm{T}_{\mathrm{OPR}}$ | -40 | - | +85 | ${ }^{\circ} \mathrm{C}$ |
| Junction temperature | $\mathrm{T}_{\mathrm{J}}$ | - | - | +125 | ${ }^{\circ} \mathrm{C}$ |

[^2]Electrical Characteristics ( $\mathbf{T a}=\mathbf{2 5}{ }^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{iN}}=\mathbf{1 2 V}$, unless otherwise noted)

| Parameters | Condition | Min. | Typ. | Max. | Unit | Note |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Supply current (shutdown) | $\mathrm{V}_{\text {EN }}=0 \mathrm{~V}$ | - | 0.2 | - | $\mu \mathrm{A}$ |  |
| Supply current (quiescent) | $\mathrm{V}_{\mathrm{EN}}=2 \mathrm{~V}, \mathrm{~V}_{\mathrm{FB}}=1 \mathrm{~V}$ | - | - | 1 | mA |  |
| Feedback voltage ( $\mathrm{V}_{\mathrm{FB}}$ ) | $4.5 \mathrm{~V} \leq \mathrm{VIN} \leq 24 \mathrm{~V}$ | 0.79 | 0.81 | 0.83 | V |  |
| Feedback current ( $\mathrm{I}_{\mathrm{FB}}$ ) | $\mathrm{V}_{\mathrm{FB}}=0.81 \mathrm{~V}$ | - | 0.1 | - | $\mu \mathrm{A}$ |  |
| Switch On Resistance ( $\mathrm{R}_{\text {DS.on }}$ ), (Note5) | - | - | 0.4 | - | $\Omega$ |  |
| Switch Leakage | $\mathrm{V}_{\mathrm{EN}}=0 \mathrm{~V}, \mathrm{~V}_{\text {SW }}=0 \mathrm{~V}$ | - | - | 30 | $\mu \mathrm{A}$ |  |
| Current Limit (Note5) | - | - | 1.8 | - | A |  |
| Oscillator frequency ( $\mathrm{F}_{\text {SW }}$ ) | $\mathrm{V}_{\mathrm{FB}}=0.81 \mathrm{~V}$ | 1.2 | 1.4 | 1.7 | MHz |  |
| Fold-back frequency | $\mathrm{V}_{\mathrm{FB}}=0 \mathrm{~V}$ | - | 460 | - | kHz |  |
| Maximum Duty cycle | $\mathrm{V}_{\mathrm{FB}}=0.81 \mathrm{~V}$ | - | 85 | - | \% |  |
| UVLO rising threshold | - |  |  | 4 | V |  |
| UVLO threshold hysteresis | - | - | 200 | - | mV |  |
| EN input low voltage | - | - | - | 0.4 | V |  |
| EN input high voltage | - | 1.2 | - | - | V |  |
|  | $\mathrm{V}_{\text {EN }}=2 \mathrm{~V}$ | - | 5 | - | $\mu \mathrm{A}$ |  |
| EN input current | $\mathrm{V}_{\mathrm{EN}}=0 \mathrm{~V}$ | - | 0.1 | - | $\mu \mathrm{A}$ |  |
| Thermal shutdown (Note5) | - | - | 140 | - | ${ }^{\circ} \mathrm{C}$ |  |

Note5. guaranteed by design.

[^3]
## TYPICAL PERFORMANCE CHARACTERISTICS

$\mathrm{V}_{\text {IN }}=12 \mathrm{~V}, \mathrm{~V}_{\text {out }}=5 \mathrm{~V}, \mathrm{~L} 1=4.7 \mu \mathrm{H}, \mathrm{C} 1=10 \mu \mathrm{~F}, \mathrm{C} 2=22 \mu \mathrm{~F}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$, Unless otherwise mentioned

Efficiency vs. Load current


Switching Frequency vs. Temperature


Efficiency vs. Load current


Feedback voltage vs. Temperature


[^4]
## TYPICAL PERFORMANCE CHARACTERISTICS

$\mathrm{V}_{\mathrm{IN}}=12 \mathrm{~V}, \mathrm{~V}_{\text {out }}=3.3 \mathrm{~V}, \mathrm{~L} 1=4.7 \mu \mathrm{H}, \mathrm{C} 1=10 \mu \mathrm{~F}, \mathrm{C} 2=22 \mu \mathrm{~F}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$, Unless otherwise noted

Transient Load Response


Shut down through Enable


Shut down through Enable


Start-up through Enable


Start-up through Enable


Dimming control


* This specifications are subject to be changed without notice


## DETAILED DESCRIPTION

## Architecture and Operation

The ADT7350 is a current mode step-down converter uses a PWM control and integrated Power Switch. An internal error amplifier establishes an integrated error voltage which is compared with slope compensated ramp signal in order to make internal Power Switch control signal. The ADT7350 has not only voltage feedback loop but also current feed-forward loop. The peak current detector limit the over current in cycle by cycle basis and gives the information of load current to the PWM controller.

At the starting of a cycle, Power Switch is off, the input and output is isolated. The fixed frequency generator start to operate. After oscillator working, every cycle of beginning, the Power Switch is turned on. The current flows into inductor thus input is shorted to output via to inductor.

The Power Switch is turned off when ;

1) The ramp signal which is the sum of the current sensing amplifier output and artificial compensated signal to prevent from sub-harmonic oscillation crosses the output of integrator .
2) The inductor current reaches the maximum allowed level (1.8A)
3) The protection signals are trigged.

When the Power Switch is off, the inductor current can not change abruptly. The external Schottky diode conducts the inductor current.

## Enable(EN)

The ADT7350 features a low-power shut down mode to decrease total current consumption. A logic Low at EN shuts down the controller and internal Power Switch. The typical value of VIH is 1.2 V , VIL is 0.4 .

## Soft-Start

The internal soft-start circuitry make a reference voltage of error amplifier gradually ramps up to reduce input surge current during start-up.

## Bootstrap

In order to generate enough large control voltage of Power Switch. The purpose of this block is to decrease Power Switch turn on resistance.

## Protection (OVP/OTP/OCP)

The ADT7350 monitors the output voltage of controller and limit the maximum allowed feedback voltage in order not to abruptly change output voltage. The output feed-back voltage over 0.95 V , this function is trigged. The ADT7350 monitors the temperature variation using PTAT(Proportional to Absolute Temperature) current. When the die temperature exceeds typical $+140{ }^{\circ} \mathrm{C}$, the Over temperature function is trigged. The maximum allowed inductor current is well controlled to prevent from falling into over-load or shorted output conditions. The maximum allowed load current is typically 1.8 A .

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## DETAILED DESCRIPTION

## UVLO (Under Voltage Lock Out)

ADT7350 can monitor the battery output voltage. If the battery output voltage is under 3.8 V . The ADT7350 goes to the UVLO status to save the power consumption of applications.

## Oscillator

The ADT7350 has a fixed frequency generator (typical 1.4 MHz ) and also automatically changes the operating frequency in accordance with feedback voltage $(0.81 \mathrm{~V})$ for protecting the load. If the feedback voltage is under 0.4 V , the operating frequency changes to one third of fixed operating frequency. (Typical $460 \mathrm{KHz})$

## Error Amplifier

The high gain error amplifier integrate the error signal from which the difference of reference voltage and feedback voltage. The summed output Error signal compares with PWM Ramp signal. It provides accuracy of the voltage feedback loop regulation. The feedback voltage of ADT7350 is typical 0.81 V .

## Current sense Amplifier

The current sense amplifier output is proportional to the current flowing into the inductor. This output goes to the comparator to make a proper control signal. The ADT7350 has well compensated current slope which is added into ramp generator circuit.

## Compensation Network

The fixed frequency peak current mode control scheme to provide easy compensation for stability and fast transient response. The cycle-by-cycle basis monitored peak current of inductor is compared with error amplifier output voltage. The duty cycle is modulated based on the inductor's peak current. The inductor can be emulated to current source. So the inductor's pole frequency goes to high frequency. It seems to be one-pole system. The ADT7350 has a inner compensation network in error amplifier block.

[^5]
## APPLICATION INFORMATION

## Output Voltage Setting



Step 1) Determine R1 value
The current of feedback resistors is

$$
\mathrm{I}_{\mathrm{FB}_{-} \mathrm{RES}}=\frac{\mathrm{V}_{\mathrm{OUT}}}{\mathrm{R}_{1}+\mathrm{R}_{2}}
$$

In order to get the good sensitivity, increase this current in spite of such amount of current consumption.
(Refer to following table for reasonable start)
Step 2) Determine R2 value the basis of following equation.

$$
\mathrm{R}_{2}=\frac{\mathrm{R}_{1}}{\frac{\mathrm{~V}_{\mathrm{OUT}}}{\mathrm{~V}_{\mathrm{REF}}}-1} @ \mathrm{~V}_{\mathrm{REF}}=810 \mathrm{mV}
$$

The voltage of feedback should be equal to internal reference voltage in well-designed negative feedback loop.

| VOUT (V) | $\mathrm{R} 1(\mathrm{~K} \Omega)$ | $\mathrm{R} 2(\mathrm{~K} \Omega)$ | $\mathrm{L}(\mu \mathrm{H})$ |
| :---: | :---: | :---: | :---: |
| 1.2 | 80.6 | 160 | 2.2 |
| 1.8 | 80.6 | 64.9 | 2.2 |
| 2.5 | 49.9 | 23.7 | 4.7 |
| 3.3 | 49.9 | 16.2 | 4.7 |
| 5 | 49.9 | 9.53 | 4.7 |
| 7 | 49.9 | 6.53 | 4.7 |
| 10 | 80.6 | 6.98 | 10 |
| 15 | 80.6 | 4.53 | 10 |

## Enable resistor Setting (R3)



Above figure is simplified input stage of ADT7350.

$$
\begin{aligned}
\mathrm{V}_{\mathrm{EN}}=\frac{\mathrm{R}_{\mathrm{INT} 1}+\mathrm{R}_{\mathrm{INT} 2}}{\mathrm{R}_{\mathrm{INT} 1}+\mathrm{R}_{\mathrm{INT} 2}+\mathrm{R}_{3}} \times \mathrm{V}_{\mathrm{IN}} \\
\quad \mathrm{R} \_ \text {int } 1=1 \mathrm{~K} \Omega, \mathrm{R} \_ \text {Int } 2=1 \mathrm{M} \Omega
\end{aligned}
$$

Example) VIN $=12 \mathrm{~V}$

1) $\mathrm{R} 3=1 \mathrm{M} \Omega$

$$
\mathrm{V}_{\mathrm{EN}}=\frac{1 \mathrm{~K} \Omega+1 \mathrm{M} \Omega}{1 \mathrm{~K} \Omega+1 \mathrm{M} \Omega+1 \mathrm{M} \Omega} \times \mathrm{V}_{\mathrm{IN}} \approx 5.8 \mathrm{~V}
$$

The zener diode clips this voltage into 5.8 V
2) $R 3=2 \mathrm{M} \Omega$

$$
\mathrm{V}_{\mathrm{EN}}=\frac{1 \mathrm{~K} \Omega+1 \mathrm{M} \Omega}{1 \mathrm{~K} \Omega+1 \mathrm{M} \Omega+2 \mathrm{M} \Omega} \times \mathrm{V}_{\mathrm{IN}}=4 \mathrm{~V}
$$

3) $\mathrm{R} 3=5 \mathrm{M} \Omega$

$$
\mathrm{V}_{\mathrm{EN}}=\frac{1 \mathrm{~K} \Omega+1 \mathrm{M} \Omega}{1 \mathrm{~K} \Omega+1 \mathrm{M} \Omega+5 \mathrm{M} \Omega} \times \mathrm{V}_{\mathrm{IN}}=2 \mathrm{~V}
$$

[^6]
## Package ; SOT23-6L



| Symbol | Dimensions In Millimeters |  | Dimensions In Inches |  |
| :---: | :---: | :---: | :---: | :---: |
|  | Min | Max | Min | Max |
| A | 1.050 | 1.250 | 0.041 | 0.049 |
| A1 | 0.000 | 0.100 | 0.000 | 0.004 |
| A2 | 1.050 | 1.150 | 0.041 | 0.045 |
| b | 0.300 | 0.500 | 0.012 | 0.020 |
| c | 0.100 | 0.200 | 0.004 | 0.008 |
| D | 2.820 | 3.020 | 0.111 | 0.119 |
| E | 1.500 | 1.700 | 0.059 | 0.067 |
| E1 | 2.650 | 2.950 | 0.104 | 0.116 |
| e | 0.950(BSC) |  | $0.037(\mathrm{BSC})$ |  |
| e1 | 1.800 | 2.000 | 0.071 | 0.079 |
| L | 0.300 | 0.600 | 0.012 | 0.024 |
| $\boldsymbol{\theta}$ | $0^{\text {o }}$ | $8^{\circ}$ | $0^{\circ}$ | $8^{\text {o }}$ |

[^7]
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[^4]:    * This specifications are subject to be changed without notice

[^5]:    * This specifications are subject to be changed without notice

[^6]:    * This specifications are subject to be changed without notice

[^7]:    * This specifications are subject to be changed without notice

