

STEPPER MOTOR CONTROLLER

April 2009

FEATURES:

- Controls Bipolar and Unipolar Motors
- Cost-effective, **low current**, pin compatible replacement for **L297**
- Torque ripple compensated half-steps - **LS8297CT**
- Half and full step modes
- Normal/wave drive
- Direction control
- Reset input
- Step control input
- Enable input
- PWM chopper circuit for current control
- Two over current sensor comparators with external references input
- All inputs and outputs TTL/CMOS compatible (TTL for 5V operation)
- **Supply current < 400uA**
- 4.75 to 7V Operation ($V_{DD} - V_{SS}$).
- **LS8297** (DIP), **LS8297-S** (SOIC), **LS8297-TS** (TSSOP)
- **LS8297CT** (DIP), **LS8297CT-S** (SOIC), **LS8297CT-TS** (TSSOP)

– See Figure 1 –

DESCRIPTION:

The **LS8297** Stepper Motor Controller generates four phase drive signal outputs for controlling two phase Bipolar and four phase Unipolar motors. The outputs are used to drive two H-bridges for the two motor windings in the Bipolar motor or the four driver transistors for the two center-tapped windings in the Unipolar motor. The motor can be driven in full step mode either in normal drive (two-phase-on) or wave drive (one-phase-on) and half step mode. The **LS8297** provides two inhibit outputs which are used to control the driver stages of each of the motor phases. The circuit uses STEP, FRD/REV and HALF/FULL inputs in a translator to generate controls for the output stages.

A dual PWM chopper circuit using an on-chip oscillator, latches and voltage comparators are used to regulate the current in the motor windings. For each pair of phase driver outputs (PHA, PHB, and PHC, PHD) each pulse of the common internal oscillator sets the latch and enables the output. If the current in the motor winding causes the voltage across a sense resistor to exceed the reference voltage, V_{REF} , at the comparator inputs, the latch is reset disabling the output until the next oscillator pulse.

The CONTROL input determines whether the chopper acts on the phase driver outputs or the inhibit outputs. When the phase lines are chopped, the non-active phase line of each pair (PHA, PHB or PHC, PHD) is activated rather than de-activating the active line to reduce dissipation in the load sensing resistors. Refer to Figure 5B for Bipolar motors. If PHA is high and PHB is low, current flows through Q1, motor winding, Q4 and sense resistor R_s . When chopping occurs, PHB is brought high and circulating current flows through Q1 and D3 and not through R_s resulting in less power dissipation in R_s . Current decay is slow using this method. When the Control input is brought low, chopping occurs by bringing INH1 low. In this case circulating current flows through D2, motor winding and D3 and through the power supply to ground causing the current to decay rapidly. For Unipolar motors, only inhibit chopping is used. Refer to Figure 6. When INH1 is brought low

PIN ASSIGNMENT TOP VIEW

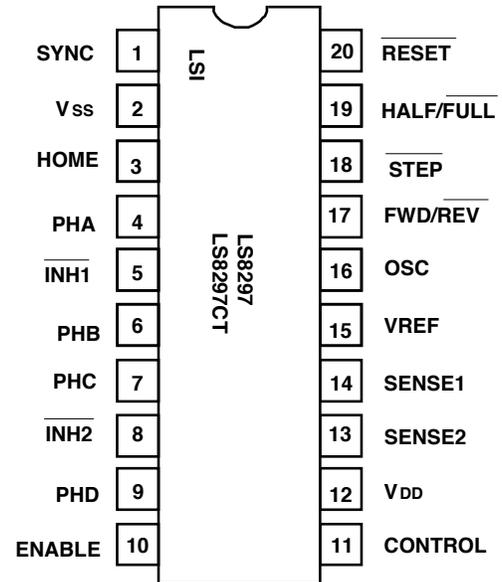


FIGURE 1

current in either half of the center tapped motor winding recirculates through the diode across it.

LS8297CT is the torque ripple compensated version of the **LS8297**. Torque imbalance resulting from alternating “one-phase on”, “two-phase on” sequence of the half-step mode (see Figure 4) is eliminated in the **LS8297CT** by switching the sense reference voltage between 100% and 70.7% in alternate steps.

INPUT/OUTPUT DESCRIPTION:

OSC Input

An RC input with the resistor connected to V_{DD} and the capacitor connected to ground determines the oscillator chopper rate. When connected as an oscillator, the oscillator output appears as a negative-going pulse at the Sync pin. If the Oscillator pin is tied to ground, the Sync pin becomes an input. Osc frequency, $f_{osc} = 1/0.69RC$

SYNC

As an output the Sync can be used to drive Sync pins of other **LS8297s**. This eliminates the need for RC components for any other **LS8297** controllers used in the system. As an input the Sync can be driven by the **LS8297** that has the RC oscillator components or by any other system external clock.

ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
Vs	Supply Voltage	10	V
Vi	Input Signals	7	V
TSTG, Tj	Storage and Junction Temperatures	-40 to +150	°C

ELECTRICAL CHARACTERISTICS: (Refer to Block Diagram, Figure 2, and Timing Diagram, Figure 3)
 TA = +25°C, VDD = +5V unless otherwise specified.

Parameter	Symbol	Minimum	Typical	Maximum	Unit	Condition
(Pin 12)						
Supply Voltage	VDD	4.75	-	7	V	-
Quiescent Supply Current	IDD	-	300	400	uA	Outputs floating
(Pins 11, 17, 18, 19, 20)						
Input Voltage Low	VIL	0	-	0.75	V	-
Input Voltage High	VIH	2	-	VDD	V	-
Input Current	Ii	-	-	50	nA	Vi = VIL
Input Current	Ii	-	-	50	nA	Vi = VIH
(Pin 10)						
Enable Input Voltage Low	VENL	0	-	1.3	V	-
Enable Input Voltage High	VENH	2	-	VDD	V	-
Enable Input Current	IEN	-	-	50	nA	VEN = VENL
Enable Input Current	IEN	-	-	50	nA	VEN = VENH
(Pins 4, 6, 7, 9)						
Phase Output Voltage Low	VOL	-	-	0.5	V	Io = -10mA
Phase Output Voltage High	VOH	4.0	-	-	V	Io = 5mA
(Pins 5, 8)						
Inhibit Output Voltage Low	VinhL	-	-	0.5	V	Io = -10mA
Inhibit Output Voltage High	VinhH	4.0	-	-	V	Io = 5mA
(Pin 3)						
Leakage Current	ILeak	-	-	1	uA	VO = VDO = 7V
Saturation Voltage	VSat	-	-	0.4	V	I = 5mA
(Pins 13, 14, 15)						
Comparators Offset Voltage	Voff	-	5	-	mV	VREF = 1V
Comparator Bias Current	IO	100	-	10	uA	-
(Pin 15)						
Input Reference Voltage	VREF	0	-	3	V	-
Input Current	IREF	-	-	8	uA	VREF = 3V
Clock Time Step Pulse Width	tstp	0.5	-	-	us	-
Set up time	tS	1	-	-	us	-
Hold time	tH	4	-	-	us	-
Reset time	tR	1	-	-	us	-
Reset to Step delay	tRStp	1	-	-	us	-
(Pin 16)						
Oscillator:						
Sawtooth Low	VSOL	-	2.1	-	V	-
Sawtooth High	VSOH	-	3.65	-	V	-
Frequency	fOSC	-	30	-	kHz	R = 22kΩ, C = 3.3nF

Parameter	Symbol	Minimum	Typical	Maximum	Unit	Condition
(Pin 1)						
Sync:						
Sync Output Voltage Low	V_{SyncL}	-	-	0.8	V	$I_o = -5\text{mA}$
Sync Output Voltage High	V_{SyncH}	3.0	-	-	V	$I_o = 5\text{mA}$
Sync Input Pulse Width	T_{SPW}	-	3.3	-	us	$R = 22\text{k}\Omega, C = 3.3\text{nF}$
Sync Input Switching Point	T_{SSP}	-	2.0	-	V	$\text{Pin } 16 \leq 1.0\text{V}$
Sync Input Current	I_{IS}	-	-425	-	uA	$\text{Pin } 16 \leq 1.0\text{V}, V_{\text{IN}} = V_{\text{DD}}$

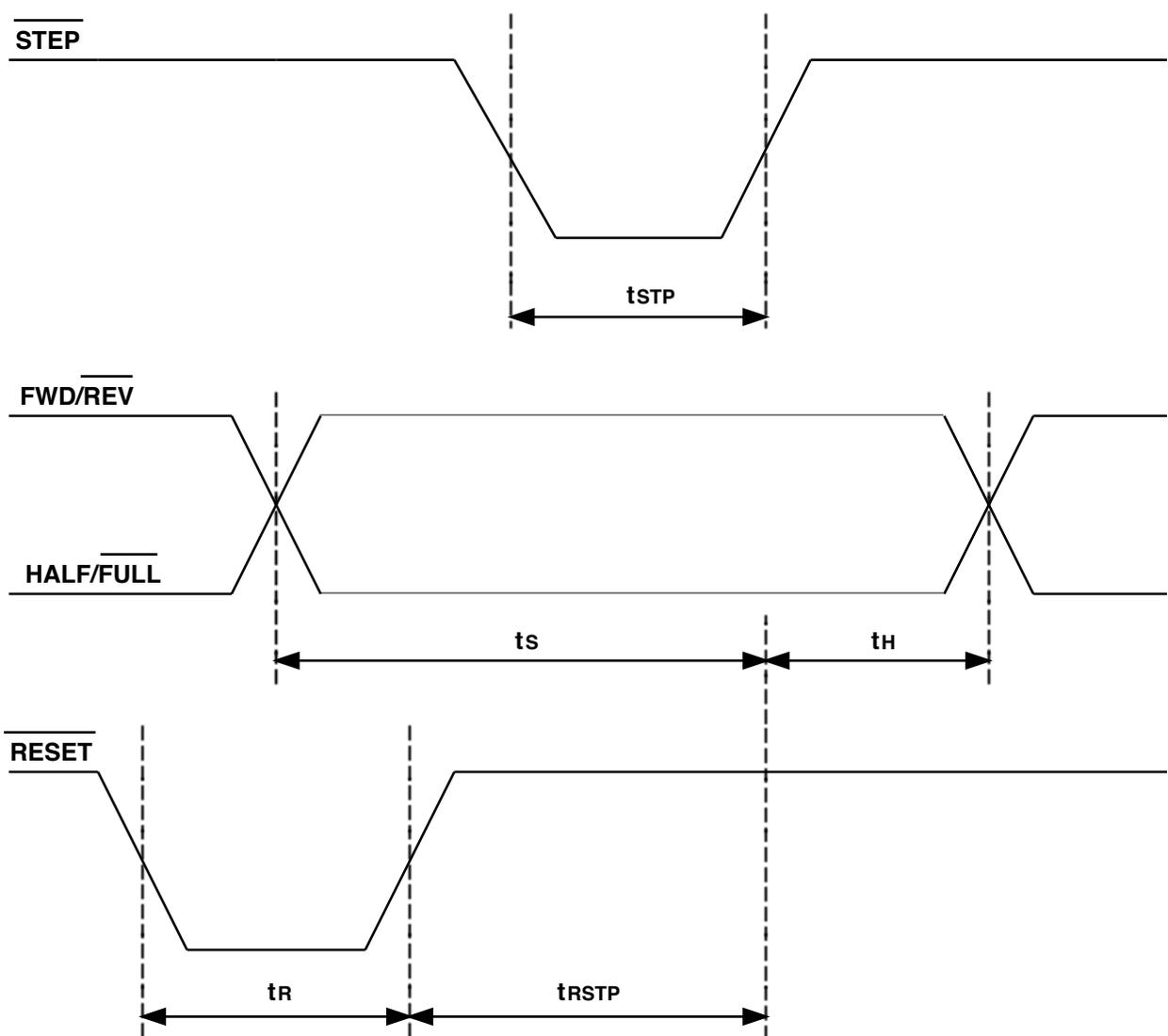


FIGURE 3. Input Timing Diagram

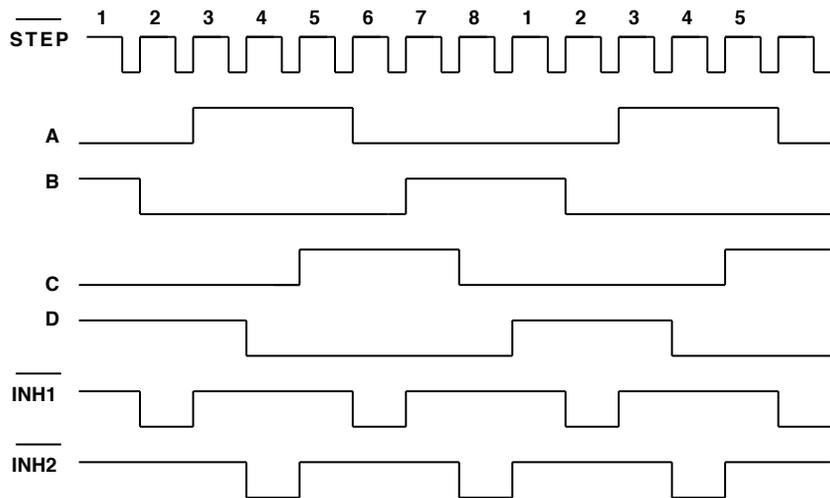
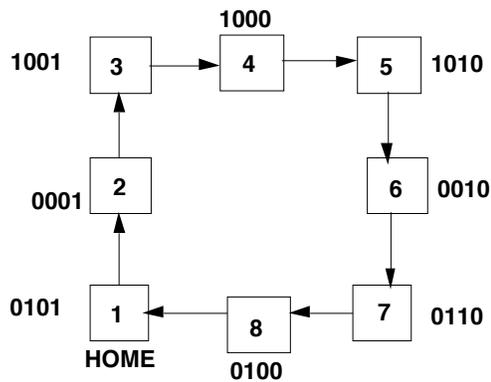


FIGURE 4A. HALF-STEP MODE

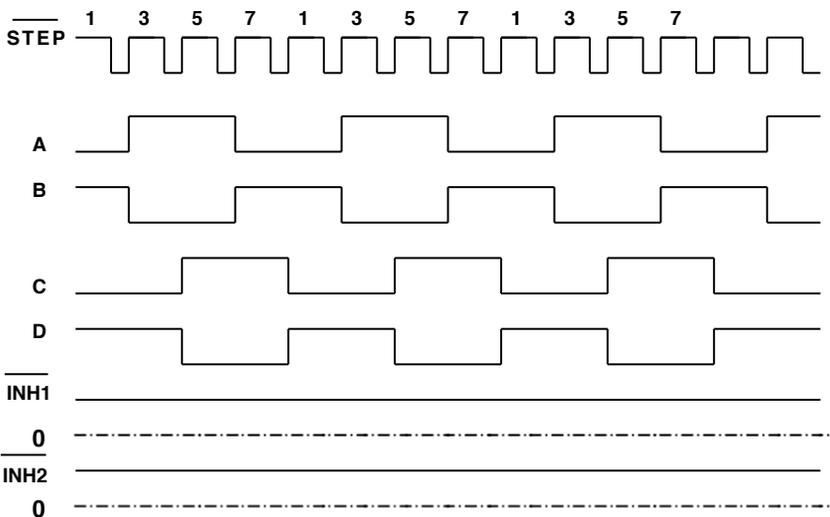
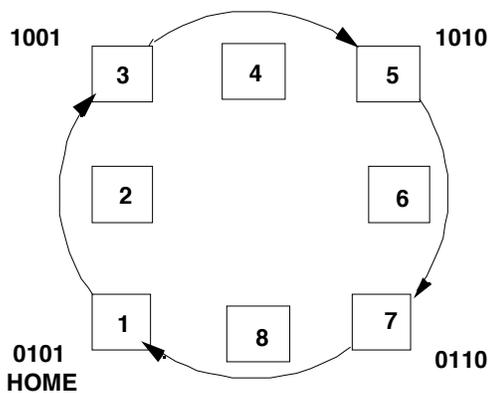


FIGURE 4B. NORMAL DRIVE MODE (TWO-PHASE-ON)

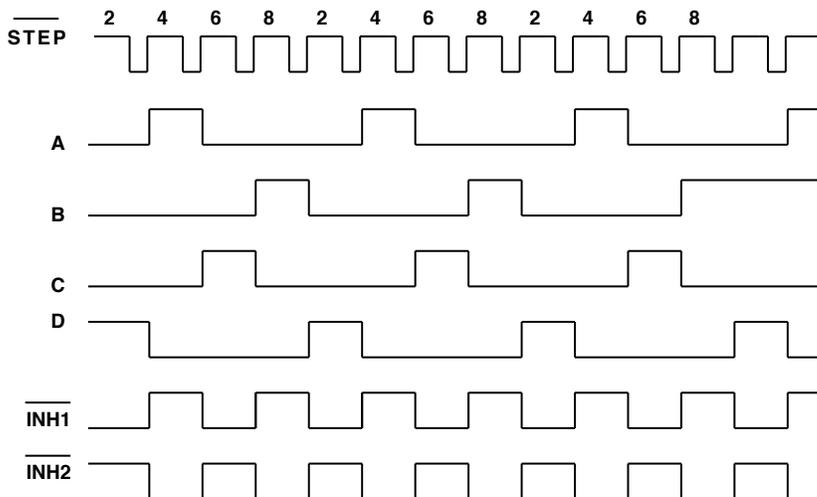
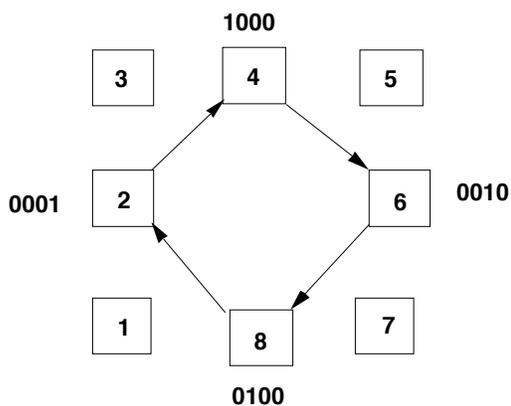
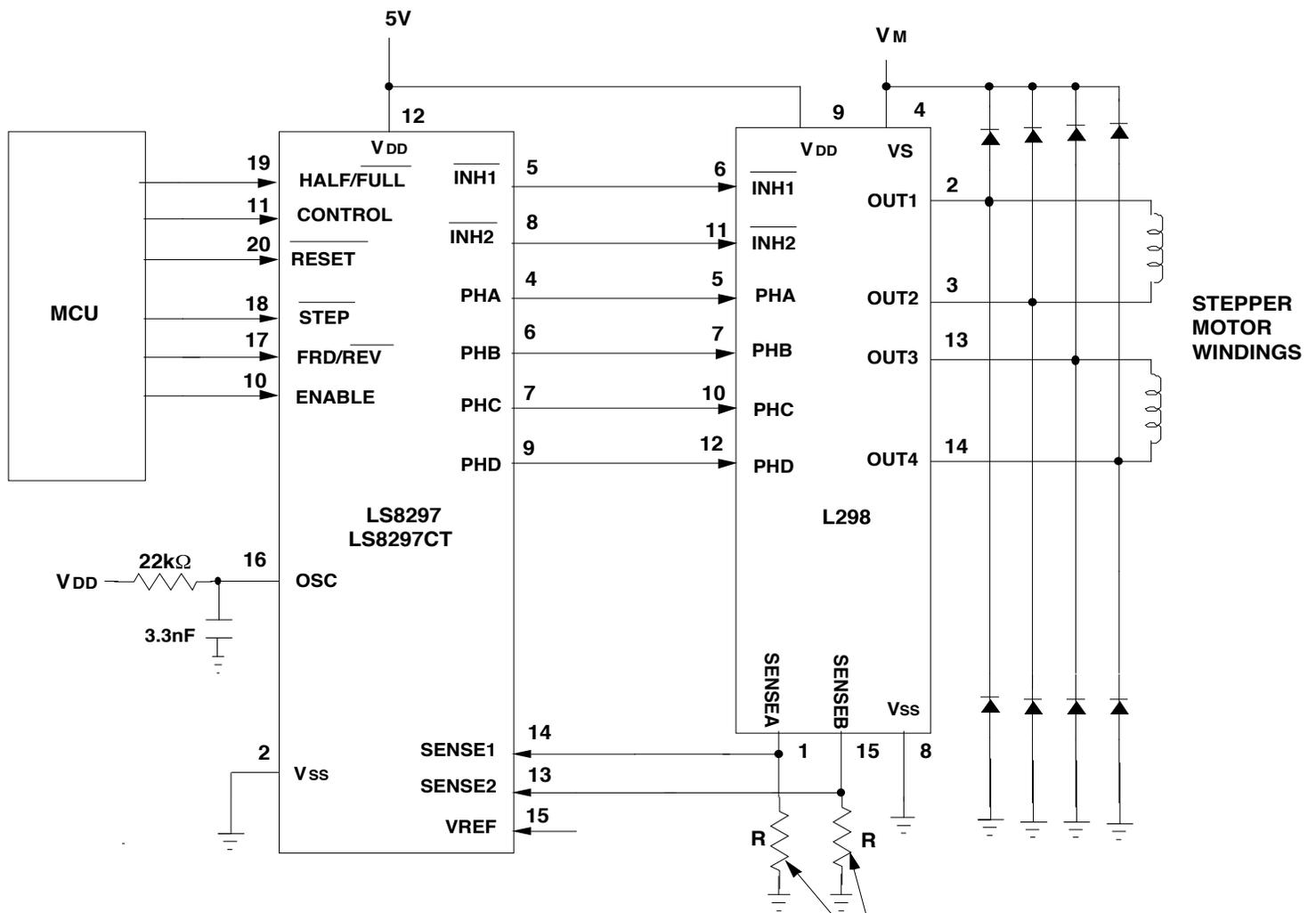


FIGURE 4C. WAVE DRIVE MODE (ONE-PHASE-ON)

FIGURE 4. MOTOR DRIVING SEQUENCES

The LS8297 generates phase sequences for half-step mode, normal drive mode and wave drive mode. Advancing occurs on the positive edge of the STEP input signal. HOME is defined as PHA, PHB, PHC, PHD being 0101, respectively. The State Diagrams showing the phase output polarities for all states are shown above for clockwise rotation. For counter clockwise rotation, the sequences are reversed. RESET restores the phases to 0101 and State 1.



Note: The SENSE resistors on L298 should be chosen so that $I_{MAX} = V_{REF}/R$, where I_{MAX} is the maximum motor winding current.

See Note

FIGURE 5A. Typical Application Schematic for a Two-Phase Bipolar Motor Using a Single Motor Driver IC

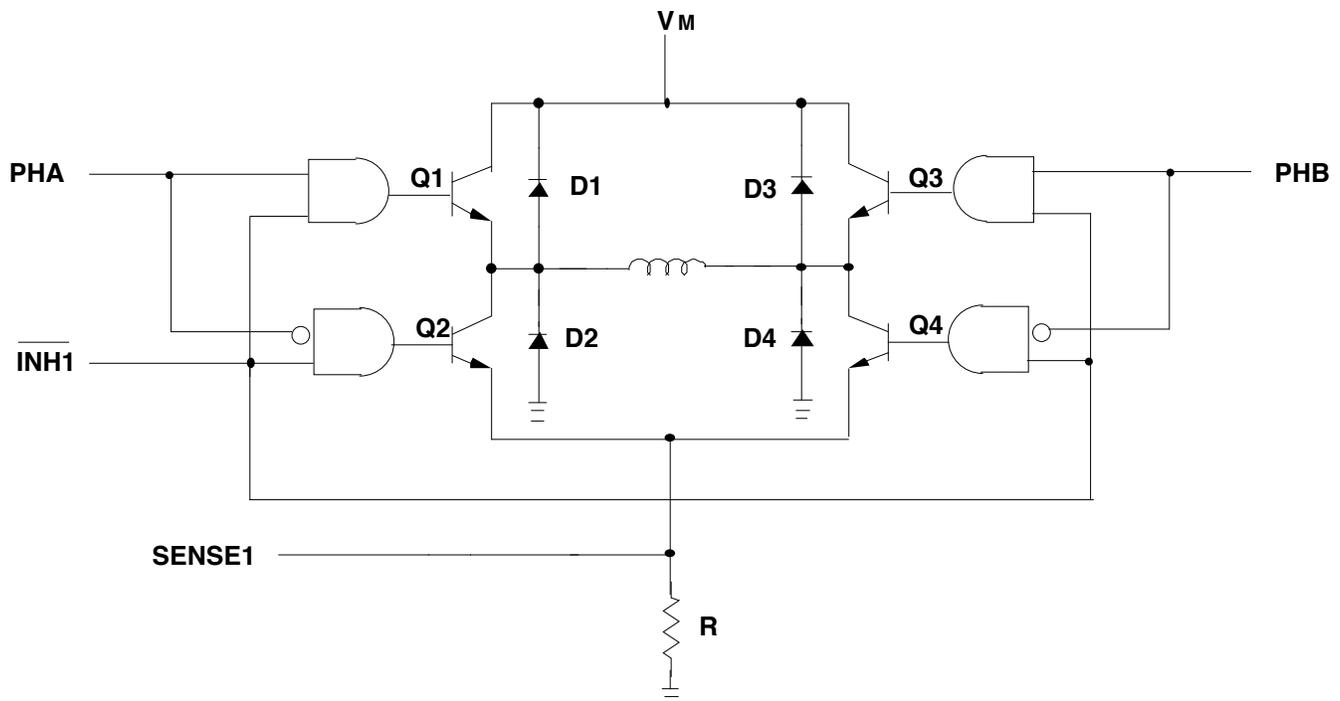
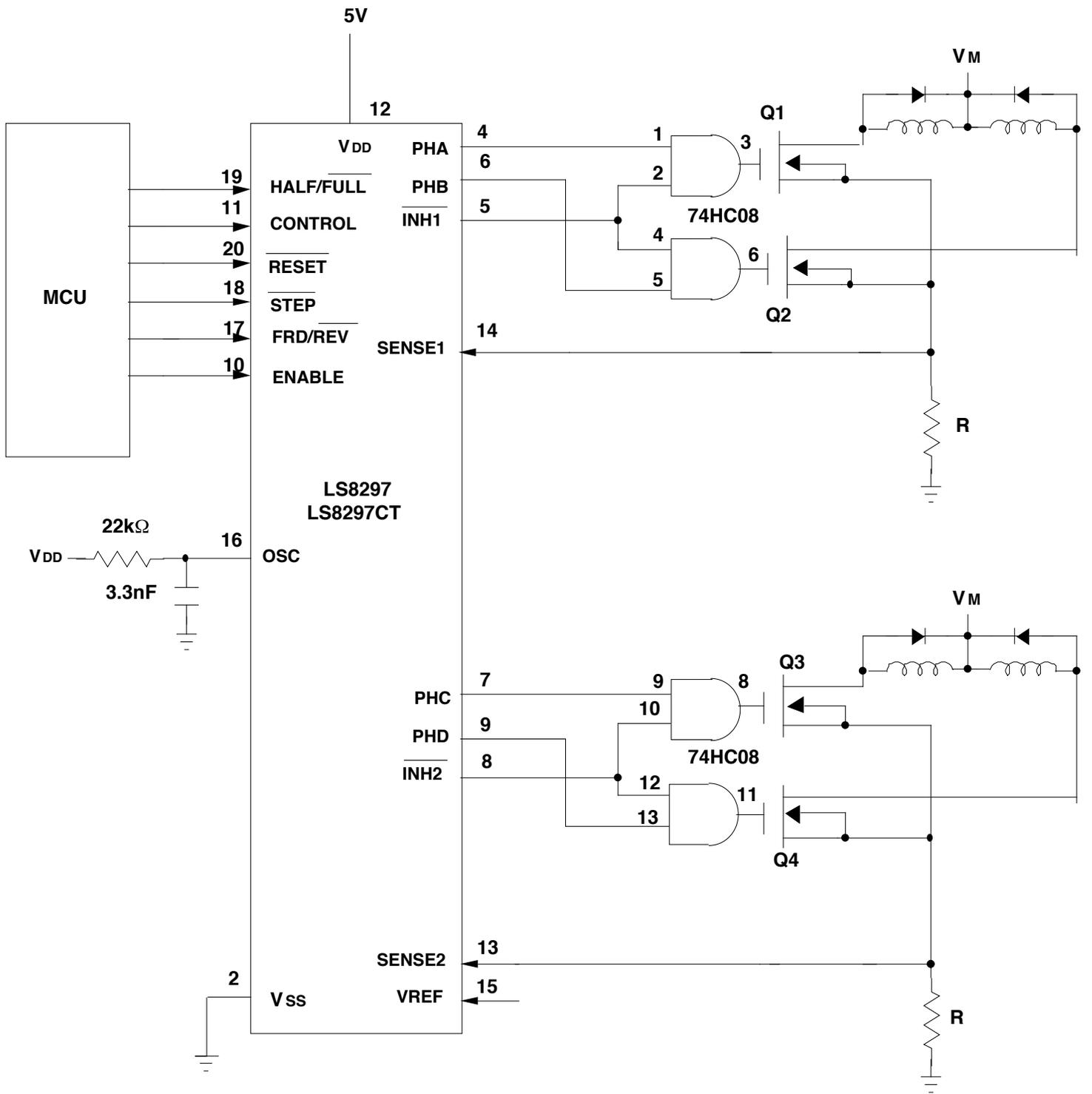


FIGURE 5B. One half of L298 Drive Stage



NOTE: Q1, Q2, Q3, Q4 are MOSFET Power Transistors suitable for 5V Gate Drive
 Typical P/Ns = IRLZ44N and IRF3708

FIGURE 6. TYPICAL APPLICATION SCHEMATIC FOR A FOUR-PHASE UNIPOLAR MOTOR USING DISCRETE MOSFET TRANSISTORS

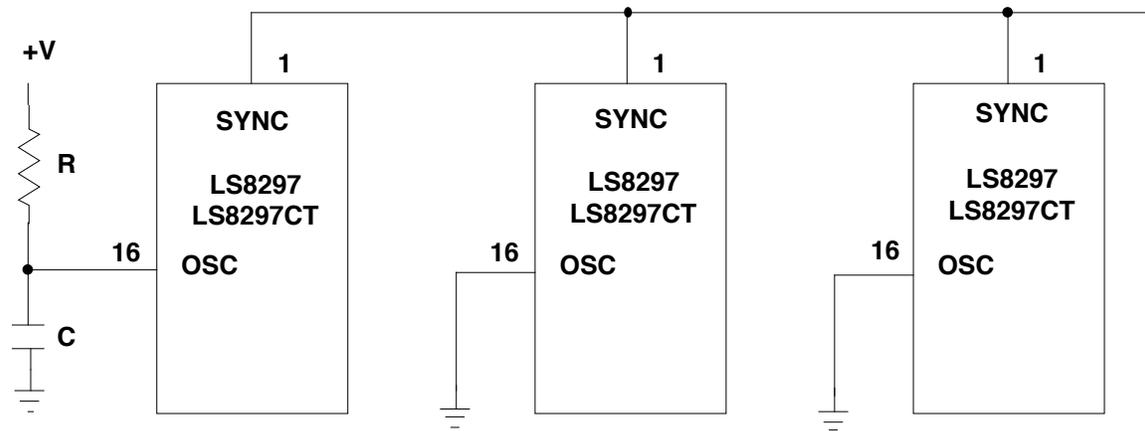


FIGURE 7. Synchronizing Multiple LS8297s

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