# BMA250 <br> Digital, triaxial acceleration sensor 

## Data sheet

## Bosch Sensortec

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## BMA250 Data sheet

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# BMA250 <br> Digital, triaxial $\pm 2 \mathrm{~g}$ to $\pm 16 \mathrm{~g}$ acceleration sensor with intelligent on-chip motion-triggered interrupt controller 

## Key features

- Ultra-Small package
- Digital interface
- Programmable functionality
- On-chip interrupt controller

LGA package (12 pins), footprint $2 \mathrm{~mm} \times 2 \mathrm{~mm}$, height 0.95 mm
SPI (4-wire, 3 -wire), ${ }^{2} \mathrm{C}$, 2 interrupt pins
$\mathrm{V}_{\text {Ddo }}$ voltage range: 1.2 V to 3.6 V
Acceleration ranges $\pm 2 \mathrm{~g} / \pm 4 \mathrm{~g} / \pm 8 \mathrm{~g} / \pm 16 \mathrm{~g}$
Low-pass filter bandwidths $1 \mathrm{kHz}-<8 \mathrm{~Hz}$
Motion-triggered interrupt-signal generation for

- new data
- any-motion (slope) detection
- tap sensing (single tap / double tap)
- orientation recognition
- flat detection
- low-g/high-g detection

Stand-alone capability (no microcontroller needed)
Low current consumption, short wake-up time, Advanced features for system power management

- RoHS compliant, halogen-free


## Typical applications

- Display profile switching
- Menu scrolling, tap / double tap sensing
- Gaming
- Pedometer / step counting
- Free-fall detection
- E-compass tilt compensation
- Drop detection for warranty logging
- Advanced system power management for mobile applications


## General description

The BMA250 is a triaxial, low-g acceleration sensor with digital output for consumer market applications. It allows measurements of acceleration in three perpendicular axes. An evaluation circuitry (ASIC) converts the output of a micromechanical acceleration-sensing structure (MEMS) that works according to the differential capacitance principle.

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Package and interfaces of the BMA250 have been defined to match a multitude of hardware requirements. Since the sensor features an ultra-small footprint and a flat package it is ingeniously suited for mobile applications.

The BMA250 offers a variable $\mathrm{V}_{\text {DDIo }}$ voltage range from 1.2 V to 3.6 V and can be programmed to optimize functionality, performance and power consumption in customer specific applications. In addition it features an on-chip interrupt controller enabling motion-based applications without use of a microcontroller.

The BMA250 senses tilt, motion and shock vibration in cell phones, handhelds, computer peripherals, man-machine interfaces, virtual reality features and game controllers.

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## 1. Specification

If not stated otherwise, the given values are over lifetime and full performance temperature and voltage ranges, minimum/maximum values are $\pm 3 \sigma$.

Table 1: Parameter Specification

| Operating Conditions |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Symbol | Condition | Min | Typ | Max | Units |
| Acceleration Range | gFS2g | Selectable via serial digital interface |  | $\pm 2$ |  | g |
|  | gFS4g |  |  | $\pm 4$ |  | g |
|  | g ${ }_{\text {FS8g }}$ |  |  | $\pm 8$ |  | g |
|  | $\mathrm{g}_{\text {FS16g }}$ |  |  | $\pm 16$ |  | g |
| Supply Voltage Internal Domains | $\mathrm{V}_{\mathrm{DD}}$ |  | 1.62 | 2.4 | 3.6 | V |
| Supply Voltage I/O Domain | $\mathrm{V}_{\text {DDIO }}$ |  | 1.2 | 2.4 | 3.6 | V |
| Voltage Input Low Level | $\mathrm{V}_{\text {IL }}$ | SPI \& ${ }^{2} \mathrm{C}$ |  |  | $0.3 \mathrm{~V}_{\text {DDIO }}$ | - |
| Voltage Input High Level | $\mathrm{V}_{\mathrm{IH}}$ | SPI \& ${ }^{2} \mathrm{C}$ | $0.7 \mathrm{~V}_{\text {DDIO }}$ |  |  | - |
| Voltage Output Low Level | $\mathrm{V}_{\text {OL }}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{DDIO}}=1.62 \mathrm{~V} \\ & \mathrm{I}_{\mathrm{OL}}=3 \mathrm{~mA}, \mathrm{SPI} \& I^{2} \mathrm{C} \end{aligned}$ |  |  | $0.2 \mathrm{~V}_{\text {DDIO }}$ | - |
|  |  | $\begin{aligned} & \mathrm{V}_{\mathrm{DDIO}}=1.2 \mathrm{~V} \\ & \mathrm{I}_{\mathrm{OL}}=3 \mathrm{~mA}, \mathrm{SPI} \& \mathrm{I}^{2} \mathrm{C} \end{aligned}$ |  |  | $\begin{gathered} 0.23 \\ \mathrm{~V}_{\text {DDIO }} \end{gathered}$ | - |
| Voltage Output High Level | $\mathrm{V}_{\mathrm{OH}}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{DDIO}}=1.62 \mathrm{~V} \\ & \mathrm{I}_{\mathrm{OL}}=2 \mathrm{~mA}, \mathrm{SPI} \& \mathrm{I}^{2} \mathrm{C} \end{aligned}$ | $0.8 \mathrm{~V}_{\text {DDIO }}$ |  |  | - |
|  |  | $\begin{aligned} & \mathrm{V}_{\mathrm{DDIO}}=1.2 \mathrm{~V} \\ & \mathrm{I}_{\mathrm{OL}}=2 \mathrm{~mA}, \mathrm{SPI} \& \mathrm{I}^{2} \mathrm{C} \end{aligned}$ | $\begin{aligned} & 0.62 \\ & \mathrm{~V}_{\text {DDIO }} \end{aligned}$ |  |  | - |
| Supply Current in Normal Mode | $\mathrm{I}_{\mathrm{DD}}$ | Nominal $\mathrm{V}_{\mathrm{DD}}$ supplies $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{bw}=1 \mathrm{kHz}$ |  | 139 |  | $\mu \mathrm{A}$ |
| Supply Current in Low-Power Mode | $\mathrm{I}_{\text {DDIp }}$ | $\begin{aligned} & \text { Nominal } \mathrm{V}_{\mathrm{DD}} \text { supplies } \\ & \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C} \text {, bw }=1 \mathrm{kHz} \\ & \text { sleep duration } \geq 25 \mathrm{~ms} \end{aligned}$ |  | 7 |  | $\mu \mathrm{A}$ |
| Supply Current in Suspend Mode | $\mathrm{I}_{\text {DSsm }}$ | Nominal $\mathrm{V}_{\mathrm{DD}}$ supplies $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |  | 0.5 |  | $\mu \mathrm{A}$ |
| Wake-Up Time | $\mathrm{t}_{\text {w_up }}$ | from Low-Power Mode or Suspend Mode, bw $=1 \mathrm{kHz}$ |  | 0.8 |  | ms |
| Start-Up Time | $\mathrm{t}_{\text {S_up }}$ | POR, bw = 1kHz |  | 2 |  | ms |
| Operating Temperature | $\mathrm{T}_{\mathrm{A}}$ |  | -40 |  | +85 | ${ }^{\circ} \mathrm{C}$ |


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| Output Signal |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Symbol | Condition | Min | Typ | Max | Units |
| Device Resolution | $\mathrm{D}_{\text {res }}$ | $\mathrm{g}_{\text {FS2g }}$ |  | 3.91 |  | mg |
| Sensitivity | $\mathrm{S}_{2 \mathrm{~g}}$ | $\mathrm{g}_{\mathrm{FS2g}}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |  | 256 |  | LSB/g |
|  | $\mathrm{S}_{48}$ | $\mathrm{g}_{\mathrm{FS4} 4}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |  | 128 |  | LSB/g |
|  | $\mathrm{S}_{88}$ | $\mathrm{g}_{\mathrm{Fs} 8 \mathrm{~g}}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |  | 64 |  | LSB/g |
|  | $\mathrm{S}_{16 \mathrm{~g}}$ | $\mathrm{g}_{\text {FS16g, }} \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |  | 32 |  | LSB/g |
| Sensitivity Temperature Drift | TCS | $\begin{aligned} & \mathrm{g}_{\mathrm{FS} 2 \mathrm{~g},}-40^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq+85^{\circ} \mathrm{C} \\ & \text { Nominal } \mathrm{V}_{\mathrm{DD}} \text { supplies } \\ & \hline \end{aligned}$ |  | $\pm 0.02$ |  | \%/K |
| Zero-g Offset | Off | $\begin{aligned} & \mathrm{g}_{\mathrm{FS2g}} \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C} \\ & \text { Nominal } \end{aligned}$ |  | $\pm 80$ |  | mg |
| Zero-g Offset <br> Temperature Drift | TCO | $\mathrm{g}_{\mathrm{FS} 2 \mathrm{~g}},-40^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq+85^{\circ} \mathrm{C}$ <br> Nominal $\mathrm{V}_{\mathrm{DD}}$ supplies |  | $\pm 1$ |  | mg/K |
| Bandwidth | $\mathrm{bw}_{8}$ | $1^{\text {st }}$ order filter, selectable via serial digital interface |  | 8 |  | Hz |
|  | $\mathrm{bw}_{16}$ |  |  | 16 |  | Hz |
|  | $\mathrm{bw}_{31}$ |  |  | 31 |  | Hz |
|  | $\mathrm{bw}_{63}$ |  |  | 63 |  | Hz |
|  | $\mathrm{bW}_{125}$ |  |  | 125 |  | Hz |
|  | $\mathrm{bw}_{250}$ |  |  | 250 |  | Hz |
|  | $\mathrm{bw}_{500}$ |  |  | 500 |  | Hz |
|  | $\mathrm{bw}_{1000}$ |  |  | 1000 |  | Hz |
| Nonlinearity | NL | best fit straight line |  | $\pm 0.5$ |  | \%FS |
| Output Noise | $\mathrm{n}_{\text {rms }}$ | $\mathrm{g}_{\text {FS22 }}, \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ <br> Nominal $\mathrm{V}_{\mathrm{DD}}$ supplies <br> Normal mode |  | 0.8 |  | $\mathrm{mg} / \sqrt{ } \mathrm{Hz}$ |
| Power Supply Rejection Rate | PSRR | $\begin{aligned} & \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C} \\ & \text { Nominal } \mathrm{V}_{\mathrm{DD}} \text { supplies } \end{aligned}$ |  |  | 20 | $\mathrm{mg} / \mathrm{V}$ |
| Temperature Sensor Measurement Range | $\mathrm{T}_{\text {s }}$ | $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ <br> Nominal $\mathrm{V}_{\mathrm{DD}}$ supplies | -40 |  | +87.5 | ${ }^{\circ} \mathrm{C}$ |
| Temperature Sensor Slope | $\mathrm{d}_{\mathrm{s}}$ | $T_{A}=25^{\circ} \mathrm{C}$ <br> Nominal $\mathrm{V}_{\mathrm{DD}}$ supplies |  | 0.5 |  | LSB/K |
| Temperature Sensor Offset | OTs | $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ <br> Nominal $\mathrm{V}_{\mathrm{DD}}$ supplies |  | $\pm 5$ |  | K |
| Mechanical Characteristics |  |  |  |  |  |  |
| Parameter | Symbol | Condition | Min | Typ | Max | Units |
| Cross Axis Sensitivity | S | relative contribution between any two of the three axes |  | 1 |  | \% |
| Alignment Error | $\mathrm{E}_{\mathrm{A}}$ | relative to package outline |  | $\pm 0.5$ |  | - |


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## 2. Absolute maximum ratings

Table 2: Absolute maximum ratings

| Parameter | Condition | Min | Max | Units |
| :--- | :--- | :---: | :---: | :---: |
| Voltage at Supply Pin | $\mathrm{V}_{\mathrm{DD}}$ Pin | -0.3 | 4.25 | V |
|  | $\mathrm{~V}_{\text {DDIO }}$ Pin | -0.3 | 4.25 | V |
| Voltage at any Logic Pad | Non-Supply Pin | -0.3 | $\mathrm{~V}_{\text {DDIO }}+0.3$ | V |
| Passive Storage Temp. Range | $\leq 65 \%$ rel. H. | -50 | +150 | ${ }^{\circ} \mathrm{C}$ |
|  | Duration $\leq 200 \mu \mathrm{~s}$ |  | 10,000 | g |
|  | Duration $\leq 1.0 \mathrm{~ms}$ |  | 2,000 | g |
|  | Free fall <br> onto hard surfaces |  | 1.8 | m |
| ESD | HBM, at any Pin |  | 2 | kV |
|  | CDM |  | 500 | V |


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## 3. Block diagram

Figure 1 shows the basic building blocks of the BMA250:


Figure 1: Block diagram of BMA250

## 4. Functional description

Note: Default values for registers can be found in chapter 5 .

### 4.1 Power management

The BMA250 has two distinct power supply pins:

- $V_{D D}$ is the main power supply for all internal analog and digital functional blocks;
- $\mathrm{V}_{\text {DDIO }}$ is a separate power supply pin, exclusively used for the supply of the digital interface.

There are no limitations on the voltage levels of both pins relative to each other, as long as each of them lies within its operating range. Furthermore, the device can be completely switched off $\left(\mathrm{V}_{\mathrm{DD}}=0 \mathrm{~V}\right)$ while keeping the $\mathrm{V}_{\text {DDIO }}$ supply on $\left(\mathrm{V}_{\text {DDIO }}>0 \mathrm{~V}\right)$. To switch off the interface supply $\left(\mathrm{V}_{\text {DDIO }}=0 \mathrm{~V}\right)$ and keep the internal supply on ( $\mathrm{V}_{\mathrm{DD}}>0 \mathrm{~V}$ ) is safe only in normal mode. If the device is in low-power mode or suspend mode while $\mathrm{V}_{\mathrm{DDIO}}=0 \mathrm{~V}$, there is a risk of excess current consumption on the $\mathrm{V}_{\mathrm{DD}}$ supply (non-destructive).

It is absolutely prohibited to keep any interface at a logical high level when $V_{\text {DDIo }}$ is switched off. Such a configuration will permanently damage the device (i.e. if $\mathrm{V}_{\text {DDIO }}=0 \rightarrow$ [SDI \& SDO \& SCK \& CSB] $\neq$ high $)$.

The device contains a power-on reset (POR) generator. It resets the logic part and the register values after powering-on $V_{D D}$ and $V_{D D I O}$. There is no limitation on the sequence of switching on

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both supply voltages. In case the $I^{2} \mathrm{C}$ interface shall be used, a direct electrical connection between $\mathrm{V}_{\text {DDIO }}$ supply and the PS pin is needed in order to ensure reliable protocol selection (see section 4.2 Operational modes).

### 4.2 Operational modes

Depending on the configuration the BMA250 is able to operate in two different operational modes:

- General mode: The device is acting as a slave on a digital interface (SPI or ${ }^{2} \mathrm{C}$ ) and is controlled by the external bus master (e.g. $\mu \mathrm{C}$ ). The master gets measurement data and status information from the device through the digital interface. In particular, the master can configure the interrupt controller and read out the interrupt status registers. Moreover, it can freely configure and use the two interrupt pins (INT1, INT2). Several interrupts may be enabled in parallel.
- Dedicated mode: The dedicated mode allows the sensor to be operated as a standalone device in a simple $\mu \mathrm{C}$-less system without abandon of the interrupt functionality. No digital interface is needed and, as a consequence, no measurement data can be read from the device. Instead of the digital interface the internal interrupt engine with its default setting is used. The interrupt status is mapped onto dedicated output pins. One out of three different sub-modes can be chosen: A) orientation recognition, B) tap sensing or C) slope (any-motion) detection. Only one interrupt at a time can be assigned.

The selection of the operational mode is done during start-up or reset by the state of the PS pin. If PS is floating, the dedicated mode is selected. A defined digital state selects the general mode. All pads are in input mode (no output driver active) during the start-up sequence until the operational mode and, in case of the general mode, the interface type is selected. The start-up sequence is run after power-up and after reset.

Figure 2 illustrates the selection of the different operational modes:


Figure 2: Operational mode selection

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### 4.2.1 General mode

A defined digital state at the PS pin selects the general mode. Its polarity determines the kind of interface to be used:

- PS = GND enables the digital SPI interface
- $P S=V_{\text {DDo }}$ enables the digital $I^{2} C$ interface
- PS = float enables the dedicated mode


### 4.2.2 Dedicated mode ( $\mu \mathrm{C}$-less or stand-alone mode)

The dedicated mode operates with pre-defined settings of the interrupt engine in order to generate the motion-triggered interrupt-signals, i.e. bandwidth, sleep time, low-power mode, threshold, and hysteresis are use case optimized. Nevertheless some minor configurations can be selected by the user. The dedicated mode is entered if the device is connected according to table 3. During the start-up / power on sequence the PS pin (\#11) must float.

Table 3: Entering and operating dedicated mode

| VDDIO <br> Pin\#3 | NC <br> Pin\#4 | VDD <br> Pin\#7 | GNDIO | Pin\#8 | GND |
| :---: | :---: | :---: | :---: | :---: | :---: |
| V Pin\#9 $^{\text {PDI }}$ | NC | V $_{\text {DD }}$ | GND | PS |  |

Depending on the configuration of the other device pins according to table 4 the corresponding sub-mode of the dedicated mode is entered. In table 4 and table 5 the unshaded entries represent necessary input values for the corresponding sub-mode selection while the shaded entries represent corresponding output parameters of the events to be detected.

Table 4: Sub-mode selection and specific outputs of the dedicated mode

| Sub-Mode | SDO <br> Pin\#1 | SDx <br> Pin\#2 | INT1 <br> Pin\#5 | INT2 <br> Pin\#6 | CSB <br> Pin\#10 | SCx <br> Pin\#12 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Orientation | output <br> orient1-detect | output <br> orient0-detect | output <br> orient2-detect | output <br> flat-detect | select <br> orient sleep | GND |
| Taublput | output <br> doubletect <br> single-detect | GND | select <br> tap type | select <br> tap sleep | $\mathrm{V}_{\mathrm{DD}}$ |  |
| Slope | GND | output <br> motion-detect | $\mathrm{V}_{\text {DD }}$ | GND | select <br> slope sleep | $\mathrm{V}_{\mathrm{DD}}$ |

Table 5 contains state and description details of the parameters introduced in table 4. Unshaded entries represent input values to be set, shaded entries represent output parameters to be detected.

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Table 5: Description of the parameters of table 4

| Sub-Mode | Parameter see Table 4 | State | Description |
| :---: | :---: | :---: | :---: |
| Orientation$\mathrm{BW}=62.5 \mathrm{~Hz}$ | output orient0-detect | low | "upright" for portrait / "left" for landscape |
|  |  | high | "upside-down" for portrait / "right" for landscape |
|  | output orient1-detect | low | portrait |
|  |  | high | landscape |
|  | outputorient2-detect | low | $z$-axis upward looking i.e. $\|\theta\|<90^{\circ}$ (Fig. 8) |
|  |  | high | $z$-axis downward looking i.e. $\|\theta\|>90^{\circ}$ (Fig. 8) |
|  | output flat-detect | low | non flat i.e. $\|\theta\|>19,5^{\circ}$ (Fig. 8) |
|  |  | high | flat i.e. $\|\theta\|<19,5^{\circ}$ (Fig. 8) |
|  | select orient sleep | GND | Low-Power mode enabled, sleep time $=100 \mathrm{~ms}$ |
|  |  | $V_{\text {DD }}$ | Low-Power mode enabled, sleep time $=1 \mathrm{~s}$ |
| Tap$\mathrm{BW}=1 \mathrm{k} \mathrm{~Hz}$ | output double-detect | low | currently no Double-Tap event |
|  |  | high | Double-Tap event detected |
|  | output single-detect | low | currently no single-tap event |
|  |  | high | Single-Tap event detected |
|  | select tap type | GND | Single-Tap detection enabled |
|  |  | $V_{D D}$ | Double-Tap detection enabled |
|  | select tap sleep | GND | Low-Power Mode disabled |
|  |  | $V_{D D}$ | Low-Power Mode enabled, sleep time $=10 \mathrm{~ms}$ |
| Slope$\mathrm{BW}=125 \mathrm{~Hz}$ | output motion-detect | low | currently no Any-Motion event |
|  |  | high | Any-Motion event detected |
|  | select slope sleep | GND | Low-Power mode enabled, sleep time $=50 \mathrm{~ms}$ |
|  |  | $V_{\text {DD }}$ | Low-Power mode enabled, sleep time $=1 \mathrm{~s}$ |

$$
\text { low }=\text { GND, high }=V_{\text {DDIO }}
$$

For more details, refer to chapter 4.3 Power modes and 4.8 Interrupt Controller

- Orientation recognition sub mode
- Tap sensing sub mode
- Any-motion (slope) detection) sub mode
$\rightarrow$ refer to chapter 4.8.7
$\rightarrow$ refer to chapter 4.8.6
$\rightarrow$ refer to chapter 4.8.5

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### 4.3 Power modes

The BMA250 has three different power modes. Besides normal mode, which represents the fully operational state of the device, there are two special energy saving modes: low-power mode and suspend mode.

The possible transitions between the power modes are illustrated in figure 3:


Figure 3: Power mode transition diagram
In normal mode, all parts of the electronic circuit are held powered-up and data acquisition is performed continuously.

In contrast to this, in suspend mode the whole analog part, oscillators included, is powered down. No data acquisition is performed, the only supported operations are reading registers (latest acceleration data are kept) and writing to the (0x11) suspend bit or (0x14) softreset register. Suspend mode is entered (left) by writing ' 1 ' (' 0 ') to the ( $0 x 11$ ) suspend bit.

In low-power mode, the device is periodically switching between a sleep phase and a wake-up phase. The wake-up phase essentially corresponds to operation in normal mode with complete power-up of the circuitry. During the sleep phase the analog part except the oscillators is powered down. Low-power mode is entered (left) by writing ' 1 ' ('0') to the ( $0 x 11$ ) lowpower_en bit.

During the wake-up phase the number of samples required by any enabled interrupt is processed. If an interrupt is detected, the device stays in the wake-up phase as long as the interrupt condition endures (non-latched interrupt), or until the latch time expires (temporary interrupt), or until the interrupt is reset (latched interrupt). If no interrupt is detected, the device enters the sleep phase.

The duration of the sleep phase is set by the (0x11) sleep_dur bits as shown in the following table:

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| :--- | :---: | :---: |

Table 6: Sleep phase duration settings

| (0x11) <br> sleep_dur | Sleep Phase <br> Duration <br> $\boldsymbol{t}_{\text {sleep }}$ |
| :---: | :---: |
| 0000b | 0.5 ms |
| 0001 b | 0.5 ms |
| 0010 b | 0.5 ms |
| 0011 b | 0.5 ms |
| 0100 b | 0.5 ms |
| 0101 b | 0.5 ms |
| 0110 b | 1 ms |
| 0111 b | 2 ms |
| 1000 b | 4 ms |
| 1001 b | 6 ms |
| 1010 b | 10 ms |
| 1011 b | 25 ms |
| 1100 b | 50 ms |
| 1101 b | 100 ms |
| 1110 b | 500 ms |
| 1111 b | 1 s |

The current consumption of the BMA250 can be calculated according to this formula:

$$
\mathrm{I}_{\mathrm{DDlp}} \approx \frac{\mathrm{t}_{\text {sleep }} \cdot \mathrm{I}_{\mathrm{DDsm}}+\mathrm{t}_{\text {active }} \cdot \mathrm{I}_{\mathrm{DD}}}{\mathrm{t}_{\text {sleep }}+\mathrm{t}_{\text {active }}}
$$

When making an estimation about the length of the wake-up phase $t_{\text {active }}$, the wake-up time, $t_{w_{-} \text {up }}$, has to be considered. Therefore, $t_{\text {active }}=t_{u t}+t_{w_{-} u p}$, where $t_{u t}$ is given in table 8. During the wake-up phase all analog modules are held powered-up, while during the sleep phase most analog modules are powered down. As a consequence, a wake-up time of less than 1 ms (typ. value 0.8 ms ) is needed to settle the analog modules in order to get reliable acceleration data.

Table 7 gives an overview of the resulting average supply currents $I_{\text {Dolpe }}$ for the different sleep phase durations and a selected bandwidth of 1000 Hz , assuming no interrupt is active and thus only one sample per wake-up phase is taken:

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| :--- | :---: | :---: |

Table 7: Average current consumption in low-power mode

| Sleep phase <br> duration | Average <br> current <br> consumption |
| :---: | :---: |
| 0.5 ms | $100.5 \mu \mathrm{~A}$ |
| 1 ms | $78.8 \mu \mathrm{~A}$ |
| 2 ms | $55.0 \mu \mathrm{~A}$ |
| 4 ms | $34.5 \mu \mathrm{~A}$ |
| 6 ms | $25.2 \mu \mathrm{~A}$ |
| 10 ms | $16.4 \mu \mathrm{~A}$ |
| 25 ms | $7.4 \mu \mathrm{~A}$ |
| 50 ms | $4.0 \mu \mathrm{~A}$ |
| 100 ms | $2.3 \mu \mathrm{~A}$ |
| 500 ms | $0.9 \mu \mathrm{~A}$ |
| 1 s | $0.7 \mu \mathrm{~A}$ |

### 4.4 Sensor data

### 4.4.1 Acceleration data

The width of acceleration data is 10 bits given in two's complement representation. The 10 bits for each axis are split into an MSB upper part (one byte containing bits 9 to 2) and an LSB lower part (one byte containing bits 1 and 0 of acceleration and a ( $0 \times 02,0 \times 04,0 \times 06$ ) new_data flag). Reading the acceleration data registers shall always start with the LSB part. The content of an MSB register is updated by reading the corresponding LSB register (shadowing procedure). The shadowing procedure can be disabled (enabled) by writing ' 1 ' ('0') to the bit shadow_dis. With disabled shadowing, the content of both MSB and LSB registers is updated by a new value immediately. Unused bits of the LSB registers are fixed to 0 . The ( $0 \times 02,0 \times 04$, 0x06) new_data flag of each LSB register is set if the data registers are updated, it is reset if either the corresponding MSB or LSB part is read.

Two different streams of acceleration data are available, unfiltered and filtered. The unfiltered data is sampled with 2 kHz . The sampling rate of the filtered data depends on the selected filter bandwidth; it is twice the bandwidth. Which kind of data is stored in the acceleration data registers depends on bit ( $0 \times 13$ ) data_high_bw. If ( $0 \times 13$ ) data_high_bw is ' 0 ' ('1'), then filtered (unfiltered) data is stored in the registers. Both data streams are separately offsetcompensated. Both kinds of data can be processed by the interrupt controller.

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| :--- | :---: | :---: |

The bandwidth of filtered acceleration data is determined by setting the ( $0 \times 10$ ) bw bit as followed:

Table 8: Bandwidth configuration

| bw | Bandwidth | Update Time <br> $\mathbf{t}_{\text {ut }}$ |
| :---: | :---: | :---: |
| 00 xxx | *) | - |
| 01000 | 7.81 Hz | 64 ms |
| 01001 | 15.63 Hz | 32 ms |
| 01010 | 31.25 Hz | 16 ms |
| 01011 | 62.5 Hz | 8 ms |
| 01100 | 125 Hz | 4 ms |
| 01101 | 250 Hz | 2 ms |
| 01110 | 500 Hz | 1 ms |
| 01111 | 1000 Hz | 0.5 ms |
| 1 xxxx | $\boldsymbol{*})$ | - |

*) Note: Settings $00 x x x$ result in a bandwidth of 7.81 Hz ; settings 1 xxxx result in a bandwidth of 1000 Hz . It is recommended to actively use the range from '01000b' to '01111b' only in order to be compatible with future products.

The BMA250 supports four different acceleration measurement ranges. A measurement range is selected by setting the ( $0 x 0 F$ ) range bits as follows:

Table 9: Range selection

| Range | Acceleration <br> measurement <br> range | Resolution |
| :---: | :---: | :---: |
| 0011 | $\pm 2 \mathrm{~g}$ | $3.91 \mathrm{mg} / \mathrm{LSB}$ |
| 0101 | $\pm 4 \mathrm{~g}$ | $7.81 \mathrm{mg} / \mathrm{LSB}$ |
| 1000 | $\pm 8 \mathrm{~g}$ | $15.62 \mathrm{mg} / \mathrm{LSB}$ |
| 1100 | $\pm 16 \mathrm{~g}$ | $31.25 \mathrm{mg} / \mathrm{LSB}$ |
| others | reserved | - |

### 4.4.2 Temperature data

The width of temperature data is 8 bits given in two's complement representation. Temperature values are available in the ( $0 \times 08$ ) temp register.

The slope of the temperature sensor is $0.5 \mathrm{~K} / \mathrm{LSB}$, its center temperature is $24^{\circ} \mathrm{C}[(0 \times 08)$ temp $=$ $0 \times 00]$. Therefore, the typical temperature measurement range is $-40^{\circ} \mathrm{C}$ up to $87.5^{\circ} \mathrm{C}$.

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| :--- | :---: | :--- |

### 4.5 Self-test

This feature permits to check the sensor functionality by applying electrostatic forces to the sensor core instead of external accelerations. By actually deflecting the seismic mass, the entire signal path of the sensor can be tested. Activating the self-test results in a static offset of the acceleration data; any external acceleration or gravitational force applied to the sensor during active self-test will be observed in the output as a superposition of both acceleration and self-test signal.

The self-test is activated individually for each axis by writing the proper value to the (0x32) self_test_axis bits ('01b' for $x$-axis, '10b' for $y$-axis, ' 11 b ' for z -axis, '00b' to deactivate selftest). It is possible to control the direction of the deflection through bit ( $0 \times 32$ ) self_test_sign. The excitation occurs in positive (negative) direction if (0x32) self_test_sign = '0b' ('1 $1 \bar{b}^{\prime}$ ).

In order to ensure a proper interpretation of the self-test signal it is recommended to perform the self-test for both (positive and negative) directions and then to calculate the difference of the resulting acceleration values. Table 10 shows the minimum differences for each axis. The actually measured signal differences can be significantly larger.

Table 10: Self-test difference values

|  | x -axis signal | y -axis signal | z -axis signal |
| :---: | :---: | :---: | :---: |
| resulting <br> minimum <br> difference signal | +0.8 g | +0.8 g | +0.4 g |

It is recommended to perform a reset of the device after self-test. If the reset cannot be performed, the following sequence must be kept to prevent unwanted interrupt generation: disable interrupts, change parameters of interrupts, wait for at least $600 \mu \mathrm{~s}$, enable desired interrupts.

### 4.6 Offset compensation

Offsets in measured signals can have several causes but they are always unwanted and disturbing in many cases. Therefore, the BMA250 offers an advanced set of four digital offset compensation methods which are closely matched to each other. These are slow, fast, and manual compensation, and inline calibration.

The compensation is performed for unfiltered and filtered data independently. It is done by adding a compensation value to the acceleration data coming from the ADC. The result of this computation is saturated if necessary to prevent any overflow errors (the smallest or biggest possible value is set, depending on the sign). However, the public registers used to read and write compensation values have only a width of 8 bits.

An overview of the offset compensation principle is given in figure 4:

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| :--- | :--- | :--- |



Figure 4: Principle of offset compensation
The meaning of both public and internal registers is the same for all acceleration measurement ranges. Therefore, with measurement ranges other than $\pm 2 \mathrm{~g}$, one or more lower significant bits of the internal registers are lost when added to an acceleration value, or are set to zero when the internal compensation value is computed. If a compensation value is too small or too big to fit into the corresponding internal register, it is saturated to prevent an overflow error.

In a similar way the conversion of the internal register value to the public register value (10bit to 8bit) uses saturation.

Summarized, in dependence to the measurement range which has been set, the compensation value, which has been written into the public register will correct the data output according to figure 4.
e.g. $\pm 2 \mathrm{~g}$ range:
public register $=00000001 \mathrm{~b} \rightarrow$ add to acceleration data $= \pm 7.8 \mathrm{mg} \quad=+2 \mathrm{LSB}$
public register $=00000010 \mathrm{~b} \rightarrow$ add to acceleration data $=+15.6 \mathrm{mg} \quad=+4 \mathrm{LSB}$
public register $=00000101 \mathrm{~b} \rightarrow$ add to acceleration data $=+39.1 \mathrm{mg} \quad=+10$ LSB

The public registers are image registers of EEPROM registers. With each image update (see section 4.7 Non-volatile memory for details) the contents of the non-volatile EEPROM registers are written to the public registers. At any time the public register can be over-written by the user. After changing the contents of the public registers by either an image update or manually, all 8 bit values are widened to 10 bit values and stored in the corresponding internal registers. In the opposite direction, if the value of an internal register changes due to the computation performed by a compensation algorithm, it is converted to an 8bit value and stored in the public register.

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| :--- | :---: | :---: |

For slow and fast offset compensation, the compensation target can be chosen by setting the bits (0x37) offset_target_x, (0x37) offset_target_y, and (0x37) offset_target_z according to table 11:

Table 11: Offset target settings

| (0x37) <br> offset_target_x/y/z | Target value |
| :---: | :---: |
| 00 b | 0 g |
| 01 b | +1 g |
| 10 b | -1 g |
| 11 b | 0 g |

By writing ' 1 ' to the ( $0 \times 36$ ) offset_reset bit, all offset compensation registers are reset to zero.

### 4.6.1 Slow compensation

Slow compensation is a quasi-continuous process which regulates the acceleration value of each axis towards the target value by comparing the current value with the target and adding or subtracting a fixed value depending on the comparison.

The algorithm in detail: If an acceleration value is larger (smaller) than the target value (0x37) offset_target_x/y/z for a number of samples (given by the parameter Offset Period see table 12), the internal offset compensation value ( $0 \times 38,0 \times 039,0 \times 3 A$ ) offset_filt_ $x / y / z$ or ( $0 \times 3 B$, $0 \times 03 C, 0 \times 3 D$ ) offset_unfilt_ $x / y / z$ is decremented (incremented) by 4 LSB.

The public registers ( $0 x 38,0 x 039,0 x 3 A$ ) offset_filt_ $x / y / z$ and ( $0 \times 3 B, 0 \times 03 C, 0 \times 3 D$ ) offset_unfilt_ $x / y / z$ are not used for the computations but they are updated with the contents of the internal registers (using saturation if necessary) and can be read by the user.

The compensation period offset_period is set by the (0x37) cut_off bit as represented in table 12:

Table 12: Compensation period settings

| (0x37) <br> cut_off | Offset <br> Period |
| :---: | :---: |
| Ob | 8 |
| 1 b | 16 |

The slow compensation can be enabled (disabled) for each axis independently by setting the bits (0x36) hp_x_en, hp_y_en, hp_z_en to '1' ('0'), respectively.

Slow compensation should not be used in combination with low-power mode. In low-power mode the conditions (availability of necessary data) for proper function of slow compensation are not fulfilled.

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| :--- | :---: | :--- |

### 4.6.2 Fast compensation

Fast compensation is a one-shot process by which the compensation value is set in such a way that when added to the raw acceleration, the resulting acceleration value of each axis equals the target value.

The algorithm in detail: An average of 16 consecutive acceleration values is computed and the difference between target value and computed value is written to ( $0 x 38,0 \times 39,0 \times 3 A$ ) offset_filt_x/y/z or ( $0 x 3 B, 0 \times 3 C, 0 \times 3 D$ ) offset_unfilt_x/y/z The public registers ( $0 x 38,0 x 39$, $0 x 3 A$ ) offset_filt_ $x / y / z$ and ( $0 \times 3 B, 0 \times 3 C, 0 x 3 D$ ) offset_unfilt_ $x / y / z$ are updated with the contents of the internal registers (using saturation if necessary) and can be read by the user.

Fast compensation is triggered for each axis individually by setting the ( $0 \times 36$ ) cal_trigger bits as shown in table 13:

Table 13: Fast compensation axis selection

| (0x36) <br> cal_trigger | Selected Axis |
| :---: | :---: |
| 00 b | none |
| 01 b | x |
| 10 b | y |
| 11 b | z |

The register (0x36) cal_trigger keeps its non-zero value while the fast compensation procedure is running. Slow compensation is blocked as long as fast compensation endures. Bit (0x36) cal_rdy is ' 0 ' when ( $0 \times 36$ ) cal_trigger is not ' 00 '.

Fast compensation should not be used in combination with low-power mode. In low-power mode the conditions (availability of necessary data) for proper function of fast compensation are not fulfilled.

### 4.6.3 Manual compensation

As explained above, the contents of the public compensation registers ( $0 \times 38,0 \times 39,0 \times 3 \mathrm{~A}$ ) offset_filt_ $x / y / z$ and ( $0 x 3 B, 0 \times 3 C, 0 \times 3 D$ ) offset_unfilt_ $x / y / z$ can be set manually via the digital interface. It is recommended to write into these registers immediately after a new data interrupt in order not to disturb running offset computations.

Writing to the offset compensation registers is not allowed if slow compensation is enabled or if the fast compensation procedure is running.

### 4.6.4 Inline calibration

For a given application, it is often desirable to calibrate the offset once and to store the compensation values permanently. This can be achieved by using one of the aforementioned offset compensation methods to determine the proper compensation values and then storing

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| :--- | :---: | :--- |

these values permanently in the non-volatile memory (EEPROM). See section 4.7 Non-volatile memory for details of the storing procedure.

Each time the device is reset, the compensation values are loaded from the non-volatile memory into the image registers and used for offset compensation until they are possibly overwritten using one of the other compensation methods.

### 4.7 Non-volatile memory

The entire memory of the BMA250 consists of three different kinds of registers: hard-wired, volatile, and non-volatile. Non-volatile memory is implemented as EEPROM. Part of it can be both read and written by the user. Access to non-volatile memory is only possible through (volatile) image registers.

Altogether, there are eight registers (bytes) of EEPROM which are accessible by the customer. The addresses of the image registers range from $0 \times 38$ to $0 \times 3 F$. While the addresses up to $0 \times 3 \mathrm{D}$ are used for offset compensation (see 4.6 Offset Compensation), addresses $0 \times 3 \mathrm{E}$ and $0 \times 3 F$ are general purpose registers not linked to any sensor-specific functionality.

The content of the EEPROM is loaded to the image registers after a reset (either POR or softreset) or after a user request which is performed by writing ' 1 ' to bit ( $0 \times 33$ ) nvm_load. As long as the image update is not yet complete, bit ( $0 \times 33$ ) nvm_load is ' 1 ', otherwise it is ' 0 '.

The image registers can be read and written like any other register.
Writing to the EEPROM is a three-step procedure:

1. Write the new contents to the image registers.
2. Write ' 1 ' to bit ( $0 \times 33$ ) nvm_prog_mode in order to unlock the EEPROM.
3. Write ' 1 ' to bit ( $0 \times 33$ ) nvm_prog_trig and keep ' 1 ' in bit ( $0 \times 33$ ) nvm_prog_mode in order to trigger the write process.

Writing to the EEPROM always renews the entire EEPROM contents. It is possible to check the write status by reading bit ( $0 \times 33$ ) nvm_rdy. While ( $0 \times 33$ ) nvm_rdy $=$ ' 0 ', the write process is still enduring; if ( $0 \times 33$ ) nvm_rdy = '1', then writing is completed. As long as the write process is ongoing, no power mode change and no image update is allowed. It is forbidden to write to the EEPROM while the image update is running, in low-power mode, and in suspend mode.

### 4.8 Interrupt controller

Seven interrupt engines are integrated in the BMA250. Each interrupt can be independently enabled and configured. If the condition of an enabled interrupt is fulfilled, the corresponding status bit is set to ' 1 ' and the selected interrupt pin is activated. There are two interrupt pins, INT1 and INT2; interrupts can be freely mapped to any of these pins. The pin state is a logic 'or' combination of all mapped interrupts.

The interrupt status registers are updated together with writing new data into the acceleration data registers. If an interrupt is disabled, all active status bits and pins are immediately reset.

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| :--- | :---: | :---: |

All time constants are based upon the typical frequency of the internal oscillator. This is reflected by the bandwidths (bw) as specified in table 1.

### 4.8.1 General features

An interrupt is cleared depending on the selected interrupt mode, which is common to all interrupts. There are three different interrupt modes: non-latched, latched, and temporary. The mode is selected by the (0x21) latch_int bits according to table 14

Table 14: Interrupt mode selection

| (0x21) <br> latch_int | Interrupt mode |
| :---: | :---: |
| 0000b | non-latched |
| 0001 b | temporary, 250 ms |
| 0010 b | temporary, 500 ms |
| 0011b | temporary, 1 s |
| 0100b | temporary, 2 s |
| 0101b | temporary, 4 s |
| 0110b | temporary, 8 s |
| 0111b | latched |
| 1000 b | non-latched |
| 1001 b | temporary, $500 \mu \mathrm{~s}$ |
| 1010 b | temporary, $500 \mu \mathrm{~s}$ |
| 1011 b | temporary, 1 ms |
| 1100 b | temporary, 12.5 ms |
| 1101 b | temporary, 25 ms |
| 1110 b | temporary, 50 ms |
| 1111 b | latched |

An interrupt is generated if its activation condition is met. It can not be cleared as long as the activation condition is fulfilled. In the non-latched mode the interrupt status bit and the selected pin (the contribution to the 'or' condition for INT1 and/or INT2) are cleared as soon as the activation condition is no more valid. Exceptions to this behaviour are the new data, orientation, and flat interrupts, which are automatically reset after a fixed time.

In the latched mode an asserted interrupt status and the selected pin are cleared by writing ' 1 ' to bit (0x21) reset_int. If the activation condition still holds when it is cleared, the interrupt status is asserted again with the next change of the acceleration registers.

In the temporary mode an asserted interrupt and selected pin are cleared after a defined period of time. The behaviour of the different interrupt modes is shown graphically in figure 5:

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| :--- | :---: | :--- |



Figure 5: Interrupt modes
Several interrupt engines can use either unfiltered or filtered acceleration data as their input. For these interrupts, the source can be selected with the respective ( $0 \times 1 E$ ) int_src_... bits, in details these are $(0 \times 1 E)$ int_src_data, $(0 \times 1 E)$ int_src_tap, $(0 \times 1 E)$ int_src_slope, $(0 \times 1 E)$ int_src_high, and ( $0 \times 1 E$ ) int_src_low. Setting the respective bits to ' 0 ' ('1') selects filtered (unfiltered) data as input. For the other interrupts, orientation recognition and flat detection, such a selection is not possible. They always use filtered input data.

It is strongly recommended to set interrupt parameters prior to enabling the interrupt. Changing parameters of an already enabled interrupt may cause unwanted interrupt generation and generation of a false interrupt history. A safe way to change parameters of an enabled interrupt is to keep the following sequence: disable the desired interrupt, change parameters, wait for at least $600 \mu \mathrm{~s}$, enable the desired interrupt.

### 4.8.2 Mapping (inttype to INT Pin\#)

The mapping of interrupts to the interrupt pins \#05 or \#06 is done by registers ( $0 \times 19$ ) to ( $0 \times 1 B$ ). Setting (0x19) int1_"inttyp" to '1' ('0') maps (unmaps) "inttyp" to pin \#5 (INT1), correspondingly setting ( $0 \times 1 B$ ) int2_"inttyp" to '1' ('0') maps (unmaps) "inttyp" to pin \#6 (INT2).

Note: "inttyp" to be replaced with the precise notation, given in the memory map in chapter 5.
Example: For flat interrupt (int1_flat): Setting (0x19) int1_flat to '1' maps int1_flat to pin \#5 (INT1).

### 4.8.3 Electrical behaviour (INT pin\# to open-drive or push-pull)

Both interrupt pins can be configured to show desired electrical behaviour. The 'active' level of each pin is determined by the ( $0 \times 20$ ) int1_Ivl and ( $0 \times 20$ ) int2_Ivl bits.

If (0x20) int1_Ivl = '1' ('0') / (0x20) int2_Ivl = '1' ('0'), then pin \#05 (INT1) / pin \#06 (INT2) is active ' 1 ' (' 0 '). In addition to that, also the electric type of the interrupt pins can be selected. By

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| :--- | :---: | :--- |

setting bits ( $0 \times 20$ ) int1_od / ( $0 \times 20$ ) int2_od to ' 0 ', the interrupt pin output type gets push-pull, by setting the configuration bits to ' 1 ', the output type gets open-drive.

Remark: Due to their use for sub-mode selection in dedicated mode, the states of both INT pins are not defined during the first 2 ms after power-up.

### 4.8.4 New data interrupt

This interrupt serves for synchronous reading of acceleration data. It is generated after storing a new value of $z$-axis acceleration data in the data register. The interrupt is cleared automatically when the next cycle of data acquisition starts. The interrupt status is ' 0 ' for at least $50 \mu \mathrm{~s}$.

The interrupt mode of the new data interrupt is fixed to non-latched.
It is enabled (disabled) by writing ' 1 ' (' 0 ') to bit ( $0 \times 17$ ) data_en. The interrupt status is stored in bit ( $0 \times 0 A$ ) data_int.

### 4.8.5 Any-motion (slope) detection

Any-motion detection uses the slope between successive acceleration signals to detect changes in motion. An interrupt is generated when the slope (absolute value of acceleration difference) exceeds a preset threshold. It is cleared as soon as the slope falls below the threshold. The principle is made clear in figure 6.


Figure 6: Principle of any-motion detection

The threshold is set with the value of register ( $0 \times 28$ ) slope_th. 1 LSB of ( $0 \times 28$ ) slope_th corresponds to 1 LSB of acceleration data. Therefore, an increment of ( $0 \times 28$ ) slope_th is 3.91 mg in 2 g -range ( 7.81 mg in 4 g -range, 15.6 mg in 8 g -range and 31.3 mg in 16 g -range). And the maximum value is 996 mg in 2 g -range ( 1.99 g in 4 g -range, 3.98 g in 8 g -range and 7.97 g in 16 g range).

The time difference between the successive acceleration signals depends on the selected bandwidth and equates to $1 /\left(2^{*}\right.$ bandwidth) ( $\Delta t=1 /\left(2^{*} b w\right)$ ). In order to suppress failure signals, the interrupt is only generated (cleared) if a certain number $N$ of consecutive slope data points is larger (smaller) than the slope threshold given by ( $0 \times 28$ ) slope_th. This number is set by the ( $0 \times 27$ ) slope_dur bits. It is $N=(0 \times 27)$ slope_dur +1 for ( $0 \times 27$ ).

Example: (0x27) slope_dur $=00 \mathrm{~b}, \ldots, 11 \mathrm{~b}=1$ decimal,..., 4 decimal

### 4.8.5.1 Enabling (disabling) for each axis

Any-motion detection can be enabled (disabled) for each axis separately by writing ' 1 ' ( 0 ') to bits (0x16) slope_en_x, (0x16) slope_en_y, (0x16) slope_en_z. The criteria for any-motion detection are fulfilled and the slope interrupt is generated if the slope of any of the enabled axes exceeds the threshold ( $0 \times 28$ ) slope_th for [(0x27) slope_dur +1$]$ consecutive times. As soon as the slopes of all enabled axes fall or stay below this threshold for [(0x27) slope_dur +1 ] consecutive times the interrupt is cleared unless interrupt signal is latched.

### 4.8.5.2 Axis and sign information of any motion interrupt

The interrupt status is stored in bit ( $0 \times 09$ ) slope_int. The any-motion interrupt supplies additional information about the detected slope. The axis which triggered the interrupt is given by that one of bits ( $0 x 0 B$ ) slope_first_x, ( $0 x 0 B$ ) slope_first_y, ( $0 x 0 B$ ) slope_first_z that contains a ' 1 '. The sign of the triggering slope is held in bit ( $0 x 0 B$ ) slope_sign. If ( $0 x 0 B$ ) slope_sign $=0^{\prime} 0^{\prime}($ ' 1 '), the sign is positive (negative).

### 4.8.5.3 Serial interface and dedicated wake-up mode

When serial interface is active, any-motion detection logic is enabled if any of the axis specific (0x16) slope_en_... register bits are set. To disable the any-motion interrupt, clear all the axis specific ( $0 \times 1 \overline{6}$ ) slope_en_... bits.
In the dedicated wake-up mode (see chapter 4.2.2), all three axes are enabled for any-motion detection whether the individual axis enable bits are set or not.

### 4.8.6 Tap sensing

Tap sensing has a functional similarity with a common laptop touch-pad or clicking keys of a computer mouse. A tap event is detected if a pre-defined slope of the acceleration of at least one axis is exceeded. Two different tap events are distinguished: A 'single tap' is a single event within a certain time, followed by a certain quiet time. A 'double tap' consists of a first such event followed by a second event within a defined time frame.

Only one of the tap interrupts can be enabled at the same time. Single tap interrupt is enabled (disabled) by writing ' 1 ' (' 0 ') to bit ( $0 \times 16$ ) s_tap_en. Double tap interrupt is enabled (disabled) by writing ' 1 ' ( 0 ') to bit ( $0 \times 16$ ) $d_{-}$tap_en. If one tries to enable both interrupts by writing ' 1 ' to

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| :--- | :---: | :---: |

( $0 x 16$ ) s_tap_en and ( $0 x 16$ ) d_tap_en, then only ( $0 x 16$ ) d_tap_en keeps the value ' 1 ' and the double tap interrupt is enabled.
The status of the single tap interrupt is stored in bit ( $0 x 09$ ) s_tap_int, the status of the double tap interrupt is stored in bit (0x09) d_tap_int.

The slope threshold for detecting a tap event is set by bits ( $0 \times 2 B$ ) tap_th. The meaning of ( $0 x 2 B$ ) tap_th depends on the range setting. 1 LSB of ( $0 \times 2 B$ ) tap_th corresponds to a slope of 62.5 mg in 2 g -range, 125 mg in 4 g -range, 250 mg in 8 g -range, and 500 mg in 16 g -range.

In figure 7 the meaning of the different timing parameters is visualized:


Figure 7: Timing of tap detection

The parameters ( $0 x 2 A$ ) tap_shock and ( $0 x 2 A$ ) tap_quiet apply to both single tap and double tap detection, while ( $0 x 2 A$ ) tap_dur applies to double tap detection only. Within the duration of ( $0 \times 2 A$ ) tap_shock any slope exceeding ( $0 \times 2 B$ ) tap_th after the first event is ignored. Contrary to this, within the duration of ( $0 \times 2 A$ ) tap_quiet no slope exceeding ( $0 \times 2 B$ ) tap_th must occur, otherwise the first event will be cancelled.

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| :--- | :---: | :--- |

### 4.8.6.1 Single tap detection

A single tap is detected and the single tap interrupt is generated after the combined durations of ( $0 \times 2 A$ ) tap_shock and ( $0 \times 2 A$ ) tap_quiet, if the corresponding slope conditions are fulfilled. The interrupt is cleared after a delay of 12.5 ms .

### 4.8.6.2 Double tap detection

A double tap is detected and the double tap interrupt is generated if an event fulfilling the conditions for a single tap occurs within the set duration in ( $0 \times 2 A$ ) tap_dur after the completion of the first tap event. The interrupt is cleared after a delay of 12.5 ms .

### 4.8.6.3 Selecting the timing of tap detection

For each of parameters ( $0 \times 2 A$ ) tap_shock and ( $0 \times 2 A$ ) tap_quiet two values are selectable. By writing ' 0 ' (' 1 ') to bit ( $0 \times 2 A$ ) tap_shock the duration of ( $0 \times 2 A$ ) tap_shock is set to 50 ms ( 75 ms ). By writing ' 0 ' (' 1 ') to bit ( $0 \times 2 \mathrm{~A}$ ) tap_quiet the duration of ( $0 \times 2 \bar{A}$ ) tap_quiet is set to 30 ms ( 20 ms ).

The length of ( $0 \times 2 A$ ) tap_dur can be selected by setting the ( $0 \times 2 A$ ) tap_dur bits according to table 15:

Table 15: Selection of tap_dur

| (0x2A) <br> tap_dur | length of tap_dur |
| :---: | :---: |
| 000b | 50 ms |
| 001 b | 100 ms |
| 010 b | 150 ms |
| 011 b | 200 ms |
| 100 b | 250 ms |
| 101 b | 375 ms |
| 110 b | 500 ms |
| 111 b | 700 ms |

### 4.8.6.4 Axis and sign information of tap sensing

The sign of the slope of the first tap which triggered the interrupt is stored in bit ( $0 \times 0 B$ ) tap_sign ('0' means positive sign, ' 1 ' means negative sign). The value of this bit persists after clearing the interrupt.

The axis which triggered the interrupt is indicated by bits ( $0 \times 0 B$ ) tap_first_x, ( $0 x 0 B$ ) tap_first_y, and ( $0 x 0 B$ ) tap_first_z.
The bit corresponding to the triggering axis contains a ' 1 ' while the other bits hold a ' 0 '. These bits are cleared together with clearing the interrupt status.

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| :--- | :---: | :---: |

### 4.8.6.5 Tap sensing in low power mode

In low-power mode, a limited number of samples is processed after wake-up to decide whether an interrupt condition is fulfilled. The number of samples is selected by bits ( $0 \times 2 B$ ) tap_samp according to table 16.

Table 16: Meaning of ( $0 \times 2 B$ ) tap_samp

| (0x2B) <br> tap_samp | Number of <br> Samples |
| :---: | :---: |
| 00 b | 2 |
| 01 b | 4 |
| 10 b | 8 |
| 11 b | 16 |

### 4.8.7 Orientation recognition

The orientation recognition feature informs on an orientation change of the sensor with respect to the gravitational field vector 'g'. The measured acceleration vector components with respect to the gravitational field are defined as shown in figure 8.


Figure 8: Definition of vector components
Therefore, the magnitudes of the acceleration vectors are calculated as follows:

$$
\begin{aligned}
& \text { acc_x }=1 g \cdot \sin \theta \cdot \cos \varphi \\
& \text { acc_y }=-1 g \cdot \sin \theta \cdot \sin \varphi \\
& \text { acc_z }=1 g \cdot \cos \theta \\
& \rightarrow \text { acc_y } \mathrm{acc} x=-\tan \varphi
\end{aligned}
$$

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| :--- | :---: | :---: |

Depending on the magnitudes of the acceleration vectors the orientation of the device in the space is determined and stored in the three ( $0 \times 0 \mathrm{C}$ ) orient bits. These bits may not be reset in the sleep phase of low-power mode. There are three orientation calculation modes with different thresholds for switching between different orientations: symmetrical, high-asymmetrical, and low-asymmetrical. The mode is selected by setting the ( $0 \times 2 \mathrm{C}$ ) orient_mode bits as given in table 17.

Table 17: Orientation mode settings

| (0x2C) <br> orient_mode | Orientation Mode |
| :---: | :---: |
| 00b | symmetrical |
| 01 b | high-asymmetrical |
| 10 b | low-asymmetrical |
| 11 b | symmetrical |

For each orientation mode the ( $0 \times 0 \mathrm{C}$ ) orient bits have a different meaning as shown in table 18 to table 20:

Table 18: Meaning of the ( $0 \times 0 \mathrm{C}$ ) orient bits in symmetrical mode

| (0x0C) orient | Name | Angle | Condition |
| :---: | :---: | :---: | :---: |
| x00 | portrait upright | $315^{\circ}<\varphi<45^{\circ}$ | $\left\|a c c \_y\right\|<\left\|a c c \_x\right\|-' h y s t '$ and acc_x - 'hyst'" $\geq 0$ |
| x01 | portrait upside down | $135^{\circ}<\varphi<225^{\circ}$ | $\left\|a c c \_y\right\|<\left\|a c c \_x\right\|-' h y s t '$ and acc_x + 'hyst' < 0 |
| x10 | landscape left | $45^{\circ}<\varphi<135^{\circ}$ | $\left\|a c c \_y\right\| \geq\left\|a c c \_x\right\|+\text { 'hyst' }$ and acc_y<0 |
| x11 | landscape right | $225^{\circ}<\varphi<315^{\circ}$ | $\begin{gathered} \left\|a c c \_y\right\| \geq\left\|a c c \_x\right\|+\text { 'hyst' } \\ \text { and acc_y } \geq 0 \end{gathered}$ |

Table 19: Meaning of the ( $0 \times 0 \mathrm{C}$ ) orient bits in high-asymmetrical mode

| (0x0C) orient | Name | Angle | Condition |
| :---: | :---: | :---: | :---: |
| $\times 00$ | portrait upright | $297^{\circ}<\varphi<63^{\circ}$ | $\mid$ acc_y\| < $2 \cdot\left\|a c c \_x\right\|-' h y s t ' ~$ and acc_x - 'hyst' $\geq 0$ |
| x01 | portrait upside down | $117^{\circ}<\varphi<243^{\circ}$ | $\begin{gathered} \mid \text { acc_y }\|<2 \cdot\| a c c \_x \mid-' h y s t ' \\ \text { and acc_x }+ \text { 'hyst' }<0 \end{gathered}$ |
| x10 | landscape left | $63^{\circ}<\varphi<117^{\circ}$ | $\begin{gathered} \left\|a^{2} z_{y} y\right\| \geq 2 \cdot\left\|a c c \_x\right\|+\text { 'hyst' } \\ \text { and acc_y }<0 \end{gathered}$ |
| x11 | landscape right | $243^{\circ}<\varphi<297^{\circ}$ | $\begin{gathered} \left\|a c c \_y\right\| \geq 2 \cdot\left\|a c c \_x\right\|+\text { 'hyst' } \\ \text { and acc_y } \geq 0 \end{gathered}$ |


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| :--- | :---: | :---: |

Table 20: Meaning of the ( $0 \times 0 C$ ) orient bits in low-asymmetrical mode

| (0x0C) <br> orient | Name | Angle | Condition |
| :---: | :---: | :---: | :---: |
| $x 00$ | portrait upright | $333^{\circ}<\varphi<27^{\circ}$ | $\mid$ acc_y <br> and $<0.5 \cdot \mid a c c \_x-x-h y s t ' ~$ 'hyst' |$|$

In the preceding tables, the parameter 'hyst' stands for a hysteresis, which can be selected by setting the ( $0 x 0 \mathrm{C}$ ) orient_hyst bits. 1 LSB of ( $0 \times 0 \mathrm{C}$ ) orient_hyst always corresponds to 62.5 mg , in 2 g -range, 125 mg in 4 g -range, 250 mg in 8 g -range and 500 mg in 16 g -range.. It is important to note that by using a hysteresis $\neq 0$ the actual switching angles become different from the angles given in the tables since there is an overlap between the different orientations.

The most significant bit of the ( $0 \times 0 \mathrm{C}$ ) orient bits (which is displayed as an ' $x$ ' in the above given tables) contains information about the direction of the $z$-axis. It is set to ' 0 ' (' 1 ') if acc_z $\geq 0$ (acc_z < 0).

Figure 9 shows the typical switching conditions between the four different orientations for the symmetrical mode (i.e. without hysteresis):


Figure 9: Typical orientation switching conditions w/o hysteresis

The orientation interrupt is enabled (disabled) by writing ' 1 ' (' 0 ') to bit ( $0 x 16$ ) orient_en. The interrupt is generated if the value of ( $0 \times 0 \mathrm{C}$ ) orient has changed. It is automatically cleared after

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| :--- | :---: | :--- |

one stable period of the $(0 \times 0 C)$ orient value. The interrupt status is stored in the ( $0 \times 09$ ) orient_int bit.

If temporary or latched interrupt mode is used, after the generation of the interrupt the changed ( $0 \times 0 \mathrm{C}$ ) orient value is kept fixed as long as the interrupt persists (e. g. until the latch time expires or the interrupt is reset). After clearing the interrupt, the ( $0 x 0 \mathrm{C}$ ) orient is only updated with the next following value change (i.e. with the next occurring interrupt). In order to ensure the continuous availability of up-to-date orientation data it is therefore optimal to use the nonlatched interrupt. It is strongly advised against using latched interrupt mode or temporary interrupt mode with latch times above 50 ms for orient recognition.

### 4.8.7.1 Orientation blocking

The change of the ( $0 \times 0 \mathrm{C}$ ) orient value and - as a consequence - the generation of the interrupt can be blocked according to conditions selected by setting the value of the ( $0 \times 2 \mathrm{C}$ ) orient_blocking bits as described by table 21.

Table 21: Blocking conditions for orientation recognition

| $(0 \times 2 C)$ <br> orient blocking | Conditions |
| :---: | :---: |
| 00b | no blocking |
| 01b | theta blocking |
| 10b | theta blocking or orceleration slope in any axis $>0.2 \mathrm{~g}$ |
| 11b | ```value of orient is not stable for at least }100\textrm{ms or theta blocking or acceleration slope in any axis > 0.4 g``` |

The theta blocking is defined by the following inequality:

$$
|\tan \theta|<\frac{\sqrt{\text { blocking_theta }}}{8} .
$$

The parameter blocking_theta of the above given equation stands for the contents of the ( $0 \times 2 \mathrm{D}$ ) orient_theta bits. Hereby it is possible to define a blocking angle between $0^{\circ}$ and $44.8^{\circ}$. The internal blocking algorithm saturates the acceleration values before further processing. As a consequence, the blocking angles are strictly valid only for a device at rest; they can be different if the device is moved.

Example:
To get a maximum blocking angle of $19^{\circ}$ the parameter blocking_theta is determined in the following way: $\left(8^{*} \tan \left(19^{\circ}\right)\right)^{2}=7.588$, therefore, blocking_value $=8 \mathrm{dec}=001000 \mathrm{~b}$ has to be chosen.

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| :--- | :---: | :---: |

In order to avoid unwanted generation of the orientation interrupt in a nearly flat position ( $z \sim 0$, sign change due to small movements or noise), a hysteresis of 0.2 g is implemented for the z axis, i. e. a after a sign change the interrupt is only generated after $|z|>0.2 \mathrm{~g}$.

### 4.8.8 Flat detection

The flat detection feature gives information about the orientation of the devices' $z$-axis relative to the $g$-vector, i. e. it recognizes whether the device is in a flat position or not.

The condition for the device to be in the flat position is

$$
|\tan \theta|<\frac{\sqrt{\text { parameter_theta }}}{8} .
$$

Like blocking_theta, used with orientation recognition, the parameter_theta stands for a userdefined setting. In this case the content of the ( $0 \times 2 E$ ) flat_theta bits. The possible flat angles also range from $0^{\circ}$ to $44.8^{\circ}$. To ensure proper operation, parameter_theta has to be less than or equal to blocking_theta.

The flat interrupt is enabled (disabled) by writing ' 1 ' ('0') to bit ( $0 \times 16$ ) flat_en. The flat interrupt is generated if the flat value has changed and the new value is stable for at least the time given by the ( $0 \times 2 \mathrm{~F}$ ) flat_hold_time bits. The flat value is stored in the ( $0 \times 0 \mathrm{C}$ ) flat bit if the interrupt is enabled. This value is ' 1 ' if the device is in the flat position, it is ' 0 ' otherwise. The content of the ( $0 \times 0 \mathrm{C}$ ) flat bit is changed only if the interrupt is generated. The interrupt is automatically cleared after one sample period. Its status is stored in the (0x09) flat_int bit.

If temporary or latched interrupt mode is used, after the generation of the interrupt the changed ( $0 \times 0 \mathrm{C}$ ) flat value is kept fixed as long as the interrupt persists (e. g. until the latch time expires or the interrupt is reset). After clearing the interrupt, the ( $0 \times 0 \mathrm{C}$ ) flat value is only updated with the next following value change (i.e. with the next occurring interrupt).

The meaning of the (0x2F) flat_hold_time bits can be seen from table 22.

Table 22: Meaning of flat_hold_time

| (0x2F) <br> flat_hold_time | Time |
| :---: | :---: |
| 00 b | 0 |
| 01 b | 512 ms |
| 10 b | 1024 ms |
| 11 b | 2048 ms |


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| :--- | :---: | :--- |

### 4.8.9 Low-g interrupt

This interrupt is based on the comparison of acceleration data against a low-g threshold, which is most useful for free-fall detection.

The interrupt is enabled (disabled) by writing ' 1 ' ('0') to the ( $0 x 17$ ) low_en bit. There are two modes available, 'single' mode and 'sum' mode. In 'single' mode, the acceleration of each axis is compared with the threshold; in 'sum' mode, the sum of absolute values of all accelerations $\left|a c c_{-} x\right|+\left|a c c_{-} y\right|+\left|a c c \_z\right|$ is compared with the threshold. The mode is selected by the contents of the (0x24) low_mode bit: '0' means 'single' mode, '1' means 'sum' mode.

The low-g threshold is set through the (0x23) low_th register. 1 LSB of (0x23) low_th always corresponds to an acceleration of 7.81 mg (i.e. increment is independent from g-range setting).

A hysteresis can be selected by setting the (0x24) low_hy bits. 1 LSB of (0x24) low_hy always corresponds to an acceleration difference of 125 mg in any g-range (as well, increment is independent from g-range setting).

The low-g interrupt is generated if the absolute values of the acceleration of all axes ('and' relation, in case of single mode) or their sum (in case of sum mode) are lower than the threshold for at least the time defined by the (0x22) low_dur register. The interrupt is reset if the absolute value of the acceleration of at least one axis ('or' relation, in case of single mode) or the sum of absolute values (in case of sum mode) is higher than the threshold plus the hysteresis for at least one data acquisition. In bit (0x09) low_int the interrupt status is stored.

The relation between the content of ( $0 \times 22$ ) low_dur and the actual delay of the interrupt generation is: delay [ms] = [(0x22) low_dur +1$] \cdot 2 \mathrm{~ms}$. Therefore, possible delay times range from 2 ms to 512 ms .

### 4.8.10 High-g interrupt

This interrupt is based on the comparison of acceleration data against a high-g threshold for the detection of shock or other high-acceleration events.

The high-g interrupt is enabled (disabled) per axis by writing ' 1 ' (' 0 ') to bits ( $0 \times 17$ ) high_en_x, ( $0 \times 17$ ) high_en_y, and ( $0 \times 17$ ) high_en_z, respectively. The high-g threshold is set through the (0x26) high_th register. The meaning of an LSB of (0x26) high_th depends on the selected grange: it corresponds to 7.81 mg in 2 g -range, 15.63 mg in 4 g -range, 31.25 mg in 8 g -range, and 62.5 mg in 16 g -range (i.e. increment depends from g-range setting).

A hysteresis can be selected by setting the (0x24) high_hy bits. Analogously to (0x26) high_th, the meaning of an LSB of (0x24) high_hy is g-range dependent: it corresponds to an acceleration difference of 125 mg in 2 g -range, 250 mg in 4 g -range, 500 mg in 8 g -range, and 1000 mg in 16g-range (as well, increment depends from g-range setting).

The high-g interrupt is generated if the absolute value of the acceleration of at least one of the enabled axes ('or' relation) is higher than the threshold for at least the time defined by the (0x25) high_dur register. The interrupt is reset if the absolute value of the acceleration of all enabled axes ('and' relation) is lower than the threshold minus the hysteresis for at least the

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| :--- | :---: | :--- |

time defined by the (0x25) high_dur register. In bit (0x09) high_int the interrupt status is stored. The relation between the content of ( $0 \times 25$ ) high_dur and the actual delay of the interrupt generation is delay $[\mathrm{ms}]=[(0 \times 22)$ low_dur +1$] \cdot 2 \mathrm{~ms}$. Therefore, possible delay times range from 2 ms to 512 ms .

### 4.8.10.1 Axis and sign information of high-g interrupt

The axis which triggered the interrupt is indicated by bits ( $0 x 0 \mathrm{C}$ ) high_first_x, ( $0 x 0 \mathrm{C}$ ) high_first_y, and ( $0 \times 0 C$ ) high_first_z. The bit corresponding to the triggering axis contains a ' 1 ' while the other bits hold a ' 0 '. These bits are cleared together with clearing the interrupt status. The sign of the triggering acceleration is stored in bit ( $0 \times 0 \mathrm{C}$ ) high_sign. If ( $0 \times 0 \mathrm{C}$ ) high_sign = ' 0 ' (' 1 '), the sign is positive (negative).

## 5. Register description

### 5.1 General remarks

The entire communication with the device is performed by reading from and writing to registers (exception: dedicated mode, see chapter 4.2.2). Registers have a width of 8 bits; they are mapped to a common space of 64 addresses from ( $0 \times 00$ ) up to ( $0 \times 3 F$ ). Within the used range there are several registers which are either completely or partially marked as 'reserved'. Any reserved bit is ignored when it is written and no specific value is guaranteed when read. It is recommended not to use registers at all which are completely marked as 'reserved'. Furthermore it is recommended to mask out (logical and with zero) reserved bits of registers which are partially marked as reserved.

Registers with addresses from ( $0 \times 00$ ) up to ( $0 \times 0 E$ ) are read-only. Any attempt to write to these registers is ignored. There are bits within some registers that connected with an action to be done and, therefore, are intended for write-only access, e. g. (0x21) reset_int or the entire (0x14) softreset register. Such bits always give ' 0 ' when read.

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| :--- | :---: | :---: |

### 5.2 Register map



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| :--- | :---: | :--- |

### 5.3 Chip ID

Register (0x00) Chip ID contains the chip identification number.
Table 23: Chip identification number, register (0x00)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 |

Register (0x01) is reserved

### 5.4 Acceleration data

Register (0x02) contains the LSB part of $x$-axis acceleration data and the new data flag for the $x$-axis.

Table 24: LSB part of $x$-axis acceleration, register ( $0 \times 02$ )

| (0x02) Bit | Name | Description |
| :--- | :--- | :--- |
| Bit 7 | acc_ $x_{-} l s b<1>$ | Bit 1 of $x$-axis acceleration data |
| Bit 6 | acc_ _lsb <0> | Bit 0 of $x$-axis acceleration data $=\mathbf{x ~ L S B ~}$ |
| Bit 5 | - | (fixed to 0) |
| Bit 4 | - | (fixed to 0) |
| Bit 3 | - | (fixed to 0) |
| Bit 2 | - | (fixed to 0) |
| Bit 1 | - | (fixed to 0) |
| Bit 0 | new_data_ $x$ | New data flag of $x$-axis |

Register (0x03) contains the MSB part of $x$-axis acceleration data.
Table 25: MSB part of $x$-axis acceleration, register ( $0 \times 03$ )

| (0x03) Bit | Name | Description |
| :--- | :--- | :--- |
| Bit 7 | acc_ $x-m s b<9>$ | Bit 9 of $x$-axis acceleration data $=\mathbf{x}$ MSB |
| Bit 6 | acc_ $x \_m s b<8>$ | Bit 8 of $x$-axis acceleration data |
| Bit 5 | acc_ $x \_m s b<7>$ | Bit 7 of $x$-axis acceleration data |
| Bit 4 | acc_ $x-m s b<6>$ | Bit 6 of $x$-axis acceleration data |
| Bit 3 | acc_ $x_{-} m s b<5>$ | Bit 5 of $x$-axis acceleration data |
| Bit 2 | acc_ $m s b<4>$ | Bit 4 of $x$-axis acceleration data |
| Bit 1 | acc_ $x \_m s b<3>$ | Bit 3 of $x$-axis acceleration data |
| Bit 0 | acc_ $x \_m s b<2>$ | Bit 2 of $x$-axis acceleration data |


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| :--- | :---: | :---: |

Register (0x04) contains the LSB part of $y$-axis acceleration data and the new data flag for the $y$-axis.

Table 26: LSB part of $y$-axis acceleration, register ( $0 \times 04$ )

| (0x04) Bit | Name | Description |
| :--- | :--- | :--- |
| Bit 7 | acc_y_Isb <1> | Bit 1 of $y$-axis acceleration data |
| Bit 6 | acc_y_Isb <0> | Bit 0 of $y$-axis acceleration data $=\mathbf{y}$ LSB |
| Bit 5 | - | (fixed to 0) |
| Bit 4 | - | (fixed to 0) |
| Bit 3 | - | (fixed to 0) |
| Bit 2 | - | (fixed to 0) |
| Bit 1 | - | (fixed to 0) |
| Bit 0 | new_data_y | New data flag of $y$-axis |

Register (0x05) contains the MSB part of acceleration data for the $y$-axis.
Table 27: MSB part of $y$-axis acceleration, register ( $0 \times 05$ )

| (0x05) Bit | Name | Description |
| :---: | :---: | :---: |
| Bit 7 | acc_y_msb <9> | Bit 9 of y -axis acceleration data $=\mathbf{y}$ MSB |
| Bit 6 | $a c c \_y=m s b<8>$ | Bit 8 of $y$-axis acceleration data |
| Bit 5 | $a c c \_y=m s b<7>$ | Bit 7 of y -axis acceleration data |
| Bit 4 | $a c c \_y \_m s b<6>$ | Bit 6 of $y$-axis acceleration data |
| Bit 3 | $a c c \_y \_m s b<5>$ | Bit 5 of $y$-axis acceleration data |
| Bit 2 | $a c c \_y \_m s b<4>$ | Bit 4 of y -axis acceleration data |
| Bit 1 | $a c c \_y \_m s b<3>$ | Bit 3 of y -axis acceleration data |
| Bit 0 | $a c c \_y \_m s b<2>$ | Bit 2 of y -axis acceleration data |

Register (0x06) contains the LSB part of acceleration data and the new data flag for the $z$-axis.
Table 28: LSB part of $y$-axis acceleration, register ( $0 \times 06$ )

| (0x06) Bit | Name | Description |
| :--- | :--- | :--- |
| Bit 7 | acc_z_Isb <1> | Bit 1 of z-axis acceleration data |
| Bit 6 | acc_z_Isb <0> | Bit 0 of z-axis acceleration data = z LSB |
| Bit 5 | - | (fixed to 0) |
| Bit 4 | - | (fixed to 0) |
| Bit 3 | - | (fixed to 0) |
| Bit 2 | - | (fixed to 0) |
| Bit 1 | - | (fixed to 0) |
| Bit 0 | new_data_z | New data flag of z-axis |


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| :--- | :---: | :---: |

Register ( $0 \times 07$ ) contains the MSB part of acceleration data for the z -axis.

Table 29: MSB part of $z$-axis acceleration, register ( $0 \times 07$ )

| (0x07) Bit | Name | Description |
| :---: | :---: | :---: |
| Bit 7 | acc_z_msb <9> | Bit 9 of $\mathbf{z}$-axis acceleration data $=\mathbf{z}$ MSB |
| Bit 6 | acc_z_msb <8> | Bit 8 of $z$-axis acceleration data |
| Bit 5 | acc_z_msb < 7 > | Bit 7 of $z$-axis acceleration data |
| Bit 4 | acc_z_msb <6> | Bit 6 of $z$-axis acceleration data |
| Bit 3 | acc_z_msb <5> | Bit 5 of $z$-axis acceleration data |
| Bit 2 | acc_z_msb <4> | Bit 4 of $z$-axis acceleration data |
| Bit 1 | acc_z_msb <3> | Bit 3 of $z$-axis acceleration data |
| Bit 0 | acc_z_msb <2> | Bit 2 of $z$-axis acceleration data |

### 5.5 Temperature data

Register (0x08) temp contains temperature data in two's complement representation. Center temperature $=24^{\circ} \mathrm{C} \rightarrow$ i.e. ( $0 \times 08$ ) temp $=00000000 \mathrm{~b}$ 1 LSB increment of temperature sensor is $0.5^{\circ} \mathrm{C}\left(0.9^{\circ} \mathrm{F}\right)$.

Table 30: Temperature data, register (0x08)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Temp $<7>$ | Temp <6> | Temp <5> | Temp <4> | Temp <3> | Temp <2> | Temp <1> | Temp <0> |

### 5.6 Status registers

Register (0x09) contains the states of several interrupts.

Table 31: Interrupt status, register ( $0 \times 09$ )

| (0x09) Bit | Name | Description |
| :--- | :--- | :--- |
| Bit 7 | flat_int | Flat interrupt status |
| Bit 6 | orient_int | Orientation interrupt status |
| Bit 5 | s_tap_int | Single tap interrupt status |
| Bit 4 | d_tap_int | Double tap interrupt status |
| Bit 3 | -_reserved - | reserved |
| Bit 2 | slope_int | Slope interrupt status |
| Bit 1 | high_int | High-g interrupt status |
| Bit 0 | low_int | Low-g interrupt status |


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| :--- | :---: | :--- |

Register ( $0 \times 0 A$ ) contains the status of the new data interrupt.
Table 32: New data status, register ( $0 \times 0 \mathrm{~A}$ )

| (0x0A) Bit | Name | Description |
| :--- | :--- | :--- |
| Bit 7 | data_int | New data interrupt status |
| Bit 6 | -reserved - | reserved |
| Bit 5 | -reserved - | reserved |
| Bit 4 | -reserved - | reserved |
| Bit 3 | -reserved - | reserved |
| Bit 2 | -reserved - | reserved |
| Bit 1 | -reserved - | reserved |
| Bit 0 | -reserved - | reserved |

Register ( $0 \times 0 B$ ) contains the sign and triggering axis information for the tap and slope interrupts. Here tap interrupt comprises both single and double tap interrupt.

Table 33: Tap and slope interrupts status, register ( $0 \times 0 B$ )

| (0x0B) Bit | Name | Description |
| :---: | :---: | :---: |
| Bit 7 | tap_sign | Sign of $1^{\text {st }}$ tap that triggered the interrupt (' 0 ' $=$ positive, '1'=negative) |
| Bit 6 | tap_first_z | '1' indicates that z -axis is triggering axis of tap interrupt |
| Bit 5 | tap_first_y | '1' indicates that y -axis is triggering axis of tap interrupt |
| Bit 4 | tap_first_ $x$ | '1' indicates that $x$-axis is triggering axis of tap interrupt |
| Bit 3 | slope_sign | Sign of slope that triggered the interrupt (' 0 ' = positive, ' 1 ' =negative) |
| Bit 2 | slope_first_z | ' 1 ' indicates that z -axis is triggering axis of slope interrupt |
| Bit 1 | slope_first_y | '1' indicates that y -axis is triggering axis of slope interrupt |
| Bit 0 | slope_first_x | ' 1 ' indicates that x -axis is triggering axis of slope interrupt |

Register ( $0 \times 0 \mathrm{C}$ ) contains the flat and orientation status, and the sign and triggering axis information for the high-g interrupt. Registers ( $0 \times 0 \mathrm{D}$ ) and ( $0 \times 0 \mathrm{E}$ ) are reserved.

Table 34: Flat and orientation Status, register ( $0 \times 0 \mathrm{O}$ )

| (0x0C) Bit | Name | Description |
| :---: | :---: | :---: |
| Bit 7 | flat | flat detection ('1' if flat condition is fulfilled, '0' otherwise) |
| Bit 6 | orient <2> | orientation value of $z$-axis ('0' if upward looking, '1' if downward looking) |
| Bit 5 | orient <1> | orientation value of $x-y$ plane ('00'=portrait upright, |
| Bit 4 | orient <0> | '01'= portrait upside-down, '10'=landscape left, '11'=landscape right) |
| Bit 3 | high_sign | Sign of slope that triggered the interrupt ( ${ }^{\prime} 0$ ' $=$ positive, ' 1 '=negative) |
| Bit 2 | high_first_z | '1' indicates that z-axis is triggering axis of high-g interrupt |
| Bit 1 | high_first_y | '1' indicates that y -axis is triggering axis of high-g interrupt |
| Bit 0 | high_first_x | '1' indicates that x -axis is triggering axis of high-g interrupt |

Registers ( $\mathbf{O X O D}$ ) and ( $\mathbf{0 x O E}$ ) are reserved.

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## 5.7 g-range selection

Register ( $0 \times 0 F$ ) contains the selection of the g-range. Proper settings for ( $0 \times 0 F$ ) range are '0011b' (selects $\pm 2 \mathrm{~g}$ range), '0101b' (selects $\pm 4 \mathrm{~g}$ range), '1000b' (selects $\pm 8 \mathrm{~g}$ range), '1100b' (selects $\pm 16 \mathrm{~g}$ range). All other settings are irregular; if such a setting is used, $\pm 2 \mathrm{~g}$ range is selected. Default value of (0x0F) range (after reset) is '0011b'.

Table 35: g-range, register (0x0F)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :---: | :---: |
| reserved | reserved | reserved | reserved | range <br> $\langle 3>$ | range <br> $<2>$ | range <br> $<1>$ | range <br> $<0>$ |

### 5.8 Bandwidths

Register ( $\mathbf{0 x 1 0}$ ) contains the selection of the bandwidth for filtered acceleration data. Settings for (0x10) bw are '00xxxb' (bandwidth $=7.81 \mathrm{~Hz}$ ), '01000b' (bandwidth $=7.81 \mathrm{~Hz}$ ), '01001b' (bandwidth $=15.63 \mathrm{~Hz}$ ), '01010b' (bandwidth $=31.25 \mathrm{~Hz}$ ), '01011b' (bandwidth $=62.5 \mathrm{~Hz}$ ), '01100b' (bandwidth $=125 \mathrm{~Hz}$ ), '01101b' (bandwidth $=250 \mathrm{~Hz}$ ), '01110b' (bandwidth $=500$ Hz ), '01111b' (bandwidth $=1000 \mathrm{~Hz}$ ), '1xxxxb' (bandwidth $=1000 \mathrm{~Hz}$ ). Default value of ( $0 \times 10$ ) bw (after reset) is ' $11111 b^{\prime}$ '. It is recommended to actively use the range from ' 01000 b ' to '01111b' only in order to be compatible with future products.

Table 36: Bandwidths, register ( $0 \times 10$ )

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| reserved | reserved | reserved | $b w<4>$ | $b w<3>$ | $b w<2>$ | $b w<1>$ | $b w<0>$ |

### 5.9 Power modes

Register ( $0 \times 11$ ) contains the configuration of the power modes. ( $0 \times 11$ ) suspend $=$ ' 1 ' ('0') sets (resets) suspend mode; default value of ( $0 \times 11$ ) suspend is ' 0 '.
(0x11) lowpower_en = '1' ('0') sets (resets) low-power mode, default value of (0x11) lowpower_en is ' 0 '.

The settings for ( $0 x 11$ ) sleep_dur are '0000b' to '0101b' (sleep phase duration $=0.5 \mathrm{~ms}$ ), '0110b' (sleep phase duration $=1 \mathrm{~ms}$ ), '0111b' (sleep phase duration $=2 \mathrm{~ms}$ ), '1000b' (sleep phase duration $=4 \mathrm{~ms}$ ), '1001b' (sleep phase duration $=6 \mathrm{~ms}$ ), '1010b' (sleep phase duration $=10 \mathrm{~ms}$ ), '1011b' (sleep phase duration $=25 \mathrm{~ms}$ ), ' 1100 b ' (sleep phase duration $=50 \mathrm{~ms}$ ), ' 1101 b ' (sleep phase duration $=100 \mathrm{~ms}$ ), ' 1110 b ' (sleep phase duration $=500 \mathrm{~ms}$ ), ' 1111 b ' (sleep phase duration $=1 \mathrm{~s}$ ). Default value of ( $0 \times 11$ ) sleep_dur is ' 0000 b '.

Table 37: Power modes, register (0x11)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| suspend | lowpower <br> _en | reserved | sleep_ <br> dur<3> | sleep_ <br> dur<2> | sleep_ <br> dur $<1>$ | sleep_- <br> dur $<0>$ | reserved |


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### 5.10 Special control settings

## Register (0x12) is reserved.

Register (0x13) contains settings for the configuration of the acceleration data acquisition and the data output format.
( $0 \times 13$ ) data_high_bw = '0' ('1') selects filtered (unfiltered) acceleration data to be written into the data registers $(0 \times 02)$ to ( $0 \times 07$ ). Default value of ( $0 \times 13$ ) data_high_bw is ' 0 '.
( $0 \times 13$ ) shadow_dis $={ }^{\prime} 0^{\prime}$ ('1') enables (disables) the shadowing procedure. Shadowing means that the MSB register is updated by reading the corresponding LSB register. Default value of ( $0 \times 13$ ) shadow_dis is ' 0 '.

Table 38: Acceleration data acquisition \& data output format, register (0x13)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| data_high <br> _bw | shadow <br> _dis | reserved | reserved | reserved | reserved | reserved | reserved |

Register (0x14) is the softreset register. A user-triggered reset (softreset) of the sensor is performed after writing ' $0 \times B 6$ ' to the softreset register. After that reset all registers return to their default values. Reading (0x14) softreset returns $0 \times 00$.

Register (0x15) is reserved.

### 5.11 Interrupt settings

Registers ( $0 \times 16$ ) and ( $0 \times 17$ ) contain the enable bits for the interrupts. Default value of each enable bit is ' 0 '.

Table 39: Interrupt setting, register (0x16)

| (0x16) Bit | Name | Description |
| :--- | :--- | :--- |
| Bit 7 | flat_en | '1' ('0') enables (disables) flat interrupt |
| Bit 6 | orient_en | '1' ('0') enables (disables) orientation interrupt |
| Bit 5 | s_tap_en | '1' ('0') enables (disables) single tap interrupt |
| Bit 4 | d_tap_en | '1' ('0') enables (disables) double tap interrupt |
| Bit 3 | -reserved - | reserved |
| Bit 2 | slope_en_z | '1' ('0') enables (disables) slope interrupt for z-axis |
| Bit 1 | slope_en_y | '1' ('0') enables (disables) slope interrupt for y-axis |
| Bit 0 | slope_en_ $x$ | '1'('0') enables (disables) slope interrupt for x-axis |


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| :--- | :---: | :---: |

Table 40: Interrupt setting, register (0x17)

| (0x17) Bit | Name | Description |
| :---: | :---: | :---: |
| Bit 7 | - reserved - | reserved |
| Bit 6 | - reserved - | reserved |
| Bit 5 | - reserved - | reserved |
| Bit 4 | data_en | '1' ('0') enables (disables) new data interrupt |
| Bit 3 | low_en | '1' ('0') enables (disables) low-g interrupt |
| Bit 2 | high_en_z | '1' ('0') enables (disables) high-g interrupt for z-axis |
| Bit 1 | high_en_y | '1' ('0') enables (disables) high-g interrupt for y -axis |
| Bit 0 | high_en_x | '1' ('0') enables (disables) high-g interrupt for x-axis |

Register (0x18) is reserved.

Registers ( $0 \times 19$ ) to ( $0 \times 1 B$ ) contain the mapping of interrupts onto the interrupt pins. Default value of each mapping bit is ' 0 '.

Table 41: Interrupt mapping, register (0x19)

| (0x19) Bit | Name | Description |
| :--- | :--- | :--- |
| Bit 7 | int1_flat | '1' ('0') maps (unmaps) flat interrupt to INT1 pin |
| Bit 6 | int1_orient | '1 ' ('0') maps (unmaps) orientation interrupt to INT1 pin $^{\text {Bit 5 }}$ |
| int1_s_tap | '1' ('0') maps (unmaps) single tap interrupt to INT1 pin |  |
| Bit 4 | int1_d_tap | '1' ('0') maps (unmaps) double tap interrupt to INT1 pin |
| Bit 3 | -reserved - | reserved |
| Bit 2 | int1_slope | '1' ('0') maps (unmaps) slope interrupt to INT1 pin |
| Bit 1 | int1_high | '1' ('0') maps (unmaps) high-g interrupt to INT1 pin |
| Bit 0 | int1_low | '1' ('0') maps (unmaps) low-g interrupt to INT1 pin |

Table 42: Interrupt mapping, register ( $0 \times 1 A$ )

| (0x1A) Bit | Name | Description |
| :--- | :--- | :--- |
| Bit 7 | int2_data | '1' ('0') maps (unmaps) new data interrupt to INT2 pin |
| Bit 6 | - reserved - | reserved |
| Bit 5 | -reserved - | reserved |
| Bit 4 | -reserved - | reserved |
| Bit 3 | -reserved - | reserved |
| Bit 2 | -reserved - | reserved |
| Bit 1 | -reserved - | reserved |
| Bit 0 | int1_data | '1' ('0') maps (unmaps) new data interrupt to INT1 pin |


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| :--- | :---: | :---: |

## Table 43: Interrupt mapping, register ( $0 \times 1 B$ )

| (0x1B) Bit | Name | Description |
| :---: | :---: | :---: |
| Bit 7 | int2_flat | '1' ('0') maps (unmaps) flat interrupt to INT2 pin |
| Bit 6 | int2_orient | '1' ('0') maps (unmaps) orientation interrupt to INT2 pin |
| Bit 5 | int2_s_tap | '1' ('0') maps (unmaps) single tap interrupt to INT2 pin |
| Bit 4 | int2_d_tap | '1' ('0') maps (unmaps) double tap interrupt to INT2 pin |
| Bit 3 | - reserved - | reserved |
| Bit 2 | int2_slope | '1' ('0') maps (unmaps) slope interrupt to INT2 pin |
| Bit 1 | int2_high | '1' ('0') maps (unmaps) high-g interrupt to INT2 pin |
| Bit 0 | int2_low | '1' ('0') maps (unmaps) low-g interrupt to INT2 pin |

## Registers ( $0 \times 1 C$ ) and ( $0 \times 1 D$ ) are reserved.

Register (0x1E) contains the data source definition for those interrupts with selectable data source. Default value of each data source selection bit is ' 0 '.

Table 44: Interrupt data source definition, register ( $0 \times 1 \mathrm{E}$ )

| (0x1E) Bit | Name | Description |
| :---: | :---: | :---: |
| Bit 7 | - reserved - | reserved |
| Bit 6 | - reserved - | reserved |
| Bit 5 | int_src_data | '1' ('0') selects unfiltered (filtered) data for the new data interrupt |
| Bit 4 | int_src_tap | '1' ('0') selects unfiltered (filtered) data for the single tap and double tap interrupts |
| Bit 3 | - reserved - | reserved |
| Bit 2 | int_src_slope | '1' ('0') selects unfiltered (filtered) data for the slope interrupt |
| Bit 1 | int_src_high | '1' ('0') selects unfiltered (filtered) data for the high-g interrupt |
| Bit 0 | int_src_low | '1' ('0') selects unfiltered (filtered) data for the low-g interrupt |

## Register (0x1F) is reserved.

Register ( $\mathbf{0 x 2 0}$ ) contains the behavioural configuration (electrical behaviour) of the interrupt pins. Default value of ( $0 \times 20$ ) int1_od and ( $0 \times 20$ ) int2_od is '0'. Default value of ( $0 \times 20$ ) int1_Ivl and ( $0 \times 20$ ) int2_Ivl is ' 1 '.

Table 45: Electrical behaviour of interrupt pin, register ( $0 \times 20$ )

| (0x20) Bit | Name | Description |
| :---: | :---: | :---: |
| Bit 7 | - reserved - | reserved |
| Bit 6 | - reserved - | reserved |
| Bit 5 | - reserved - | reserved |
| Bit 4 | - reserved - | reserved |
| Bit 3 | int2_od | '0' selects push-pull, '1' selects open drive for INT2 pin |
| Bit 2 | int2_\|v| | 0' ('1') selects active level ' 0 ' ('1') for INT2 pin |
| Bit 1 | int1_od | '0' selects push-pull, '1' selects open drive for INT1 |
| Bit 0 | int1_\|v| | '0' ('1') selects active level '0' ('1') for INT1 pin |


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Register ( $\mathbf{0 x 2 1}$ ) contains the interrupt reset bit and the interrupt mode selection. Writing ' 1 ' to (0x21) reset_int resets any latched interrupt.

The settings for (0x21) latch_int are '0000b' (non-latched), '0001b' (temporary, 250 ms ), '0010b' (temporary, 500 ms ), '0011b' (temporary, 1 s ), '0100b' (temporary, 2 s ), '0101b' (temporary, 4 s ), '0110b' (temporary, 8 s), '0111b' (latched), '1000b' (non-latched), '1001b' (temporary, $500 \mu \mathrm{~s}$ ), '1010b' (temporary, $500 \mu \mathrm{~s}$ ), '1011b' (temporary, 1 ms ), '1100b' (temporary, 12.5 ms ), '1101b' (temporary, 25 ms ), '1110b' (temporary, 50 ms ), '1111b' (latched).
Default value of ( $0 \times 21$ ) latch_int is '0000b'.

Table 46: Interrupt reset bit and interrupt mode selection, register (0x21)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| reset_int | reserved | reserved | reserved | latch_ <br> int $<3>$ | latch_ <br> int $<2>$ | latch_ <br> int $<1>$ | latch_ <br> int $<0>$ |

Register (0x22) contains the delay time definition for the low-g interrupt. The physical delay time can be computed from the content of ( $0 \times 22$ ) low_dur according to:
delay $[\mathrm{ms}]=[(0 \times 22)$ low_dur +1$] \cdot 2 \mathrm{~ms}$.
Possible delay times range from 2 ms to 512 ms . Default value of ( $0 \times 22$ ) low_dur is $0 \times 09$, corresponding to a delay of 20 ms .

Table 47: Delay time definition for the low-g interrupt, register (0x22)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| low dur<7> | $\begin{aligned} & \text { low } \\ & \text { dur }<\overline{6}> \end{aligned}$ | $\begin{gathered} \text { low } \\ \text { dur }<5> \end{gathered}$ | $\begin{aligned} & \text { low } \\ & \text { dur }<\overline{4}> \end{aligned}$ | $\begin{gathered} \text { low } \\ \text { dur }<\overline{3}> \end{gathered}$ | $\begin{aligned} & \text { low } \\ & \text { dur }<2 \text { 2 } \end{aligned}$ | $\begin{gathered} \text { low } \\ \text { dur }<1> \end{gathered}$ | $\begin{gathered} \text { low } \\ \text { dur }<\overline{0}> \end{gathered}$ |

Register (0x23) contains the threshold definition for the low-g interrupt. An LSB of (0x23) low_th corresponds to an actual acceleration of 7.81 mg . Therefore, the threshold ranges from 0 g to 1.992 g . Default value of ( $0 \times 23$ ) low_th is $0 \times 30$, corresponding to an acceleration of 375 mg .

Table 48: Threshold definition for the low-g interrupt, register (0x23)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| low | low | low | low | low | low | low | low |
| th $\langle\overline{7}\rangle$ | th $\langle\overline{6}>$ | th $\langle\overline{5}>$ | th $\langle\overline{4}>$ | th $\langle\overline{3}>$ | th $\langle\overline{2}>$ | th $\langle\overline{1}\rangle$ | th $\langle\overline{0}>$ |

Register ( $0 \times 24$ ) contains the low-g interrupt mode selection, the low-g interrupt hysteresis setting, and the high-g interrupt hysteresis setting. Setting ( $0 \times 24$ ) low_mode to ' 0 ' (' 1 ') selects 'single' mode ('sum' mode). Default value is ' 0 ' ('single' mode).

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| :--- | :---: | :--- |

(0x24) low_hy sets the hysteresis of the low-g interrupt. An LSB of (0x24) low_hy corresponds to an acceleration difference of 125 mg . Default value of ( $0 \times 24$ ) low_hy is ' 01 b '.
(0x24) high_hy sets the hysteresis of the high-g interrupt. The meaning of an LSB of (0x24) high_hy depends on the selected g-range. It corresponds to an acceleration difference of 125 mg in 2 g -range, 250 mg in 4 g -range, 500 mg in 8 g -range, and 1000 mg in 16 g -range.
Default value of ( $0 \times 24$ ) high_hy is '10b'.

Table 49: Threshold definition for the low-g interrupt, register (0x24)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| high_ <br> $h y<1>$ | high_ <br> hy<0> | reserved | reserved | reserved | low <br> mode | low <br> hy< | low <br> $h y<\overline{0}>$ |

Register (0x25) contains the delay time definition for the high-g interrupt. The physical delay time can be computed from the content of (0x25) high_dur according to delay $[\mathrm{ms}]=[(0 \times 25)$ high dur +1$] \cdot 2 \mathrm{~ms}$. Possible delay times range from 2 ms to 512 ms . Default value of (0x25) high_dur is 0x0F, corresponding to a delay of 32 ms .

Table 50: Delay time definition for the high-g interrupt, register (0x25)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| high | high | high | high | high | high | high | high |
| dur $<\overline{7}>$ | dur $<\overline{6}>$ | dur $<\overline{5}>$ | dur $<\overline{4}>$ | dur $<\overline{3}>$ | dur $<\overline{2}>$ | dur $<\overline{1}>$ | dur $<\overline{0}>$ |

Register (0x26) contains the threshold definition for the high-g interrupt. The meaning of an LSB of (0x26) high_th depends on the selected g-range. It corresponds to 7.81 mg in 2 g -range, 15.63 mg in 4 g -range, 31.25 mg in 8 g -range, and 62.5 mg in 16 g -range.

Default value of ( $0 \times 26$ ) high_th is $0 \times C 0$.

Table 51: Threshold definition for the high-g interrupt, register ( $0 \times 26$ )

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| high_ | high_ | high_ | high | high | high | high | high |
| th<7> | th $\langle 6>$ | th $\langle 5>$ | th $\langle 4>$ | th $\langle 3>$ | th $\langle 2>$ | th $\langle 1>$ | th $\langle 0\rangle$ |

Register ( $0 \times 27$ ) contains the definition of the number of samples to be evaluated for the slope interrupt (any-motion detection). The number of samples is $N=(0 \times 27)$ slope_dur +1 .
Default value of ( $0 \times 27$ ) slope_dur is ' 00 b '.

Table 52: Samples number definition for the slope interrupt, register (0x27)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| reserved | reserved | reserved | reserved | reserved | reserved | slope <br> dur<1> | slope- <br> dur $<0>$ |


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| :--- | :---: | :--- |

Register ( $0 \times 28$ ) contains the threshold definition for the slope interrupt. An LSB of (0x28) slope_th corresponds to an LSB of acceleration data. Its meaning therefore depends on the selected g-range. Default value of ( $0 \times 28$ ) slope_th is $0 \times 14$.

Table 53: Slope threshold for the slope interrupt, register (0x28)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| slope | slope | slope | slope | slope | slope | slope | slope |
| $t h<7\rangle$ | $t h<6>$ | $t h<5>$ | $t h<4>$ | $t h<3>$ | $t h<2\rangle$ | $t h<1\rangle$ | $t h<0\rangle$ |

Register (0x29) is reserved.

Register ( $0 \times 2 A$ ) contains the timing definitions for the single tap and double tap interrupts.
( $0 \times 2 \mathrm{~A}$ ) tap_quiet $=$ ' 0 ' ('1') selects a quiet duration of $30 \mathrm{~ms}(20 \mathrm{~ms}$ ). The default value of ( $0 \times 2 A$ ) tap_quiet is ' 0 '.
( $0 \times 2 A$ ) tap_shock = '0' ('1') selects a shock duration of $50 \mathrm{~ms}(75 \mathrm{~ms})$. The default value of ( $0 \times 2 A$ ) tap_shock is ' 0 '.
( $0 \times 2 A$ ) tap_dur selects the length of the time window for the second shock event (for double tap detection). The settings for ( $0 x 2 A$ ) tap_dur are '000b' ( 50 ms ), '001b' ( 100 ms ), '010b' (150 $\mathrm{ms})$, '011b' (200 ms), '100b' (250 ms), '101b' (375 ms), '110b' (500 ms), '111b' ( 700 ms ). The default value of ( $0 \times 2 A$ ) tap_dur is '100b'.

Table 54: Tap Quiet duration and tap shock duration, register (0x2A)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| tap <br> quiet | tap <br> shock | reserved | reserved | reserved | $\operatorname{tap}$ <br> dur $<2>$ | $\operatorname{tap}$ <br> dur $<\overline{1}>$ | $\operatorname{tap}$ <br> dur $<0>$ |

Register ( $0 \times 2 B$ ) contains the definition of the number of samples to be processed after wakeup in low-power mode and the threshold definition for the single and double tap interrupts. ( $0 \times 2 B$ ) tap_samp selects the number of samples that are processed after wake-up in the lowpower mode. The settings for ( $0 \times 2 B$ ) tap_samp are '00b' ( 2 samples), ' 01 b ' ( 4 samples), ' 10 b ' (8 samples), and '11b' (16 samples). Default value of (0x2B) tap_samp is '00b'.

The meaning of an LSB of ( $0 \times 2 B$ ) tap_th depends on the selected g-range. It corresponds to an acceleration difference of 62.5 mg in 2 g -range, 125 mg in 4 g -range, 250 mg in 8 g -range, and 500 mg in 16 g -range. Default value of ( $0 \times 2 B$ ) tap_th is $0 \times 0 \mathrm{~A}$.

Table 55: Samples number after wake-up and threshold tap interrupt, register (0x2B)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{gathered} \operatorname{tap}_{-} \\ \operatorname{samp}<1> \end{gathered}$ | $\begin{gathered} \operatorname{tap}_{-} \\ \operatorname{samp<0>} \end{gathered}$ | reserved | $\begin{aligned} & \operatorname{tap}_{-} \\ & \operatorname{th}<\overline{4}> \end{aligned}$ | $\begin{gathered} \operatorname{tap}_{t h<\overline{3}>} \end{gathered}$ | $\begin{gathered} \operatorname{tap} \\ \text { th< } \\ \hline \end{gathered}$ | $\begin{gathered} \operatorname{tap} \\ \operatorname{th}<\overline{1}> \end{gathered}$ | $\begin{aligned} & \operatorname{tap}_{-} \\ & \operatorname{th}<\overline{0}> \end{aligned}$ |


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Register ( $0 \times 2 \mathrm{C}$ ) contains the definition of hysteresis, blocking, and mode for the orientation interrupt. ( $0 \times 2 \mathrm{C}$ ) orient_hyst sets the hysteresis of the orientation interrupt; 1 LSB always corresponds to 62.5 mg , in any g-range (i.e. increment is independent from g-range setting). Default value of $(0 \times 2 C)$ orient_hyst is '001b'.
(0x2C) orient_blocking selects the kind of blocking that is used for the generation of the orientation interrupt. The settings for ( $0 \times 2 \mathrm{C}$ ) orient_blocking are '00b' (no blocking), '01b' (theta blocking), '10b' (theta blocking or slope in any axis > 0.2 g ), and '11b' (orient value not stable for at least 100 ms or theta blocking or slope in any axis $>0.4 \mathrm{~g}$ ). Default value of ( $0 \times 2 \mathrm{C}$ ) orient_blocking is ' 10 b '.
( $0 \times 2 \mathrm{C}$ ) orient_mode sets the thresholds for switching between the different orientations. The settings for ( $0 \times 2 \mathrm{C}$ ) orient_mode are '00b' (symmetrical), '01b' (high-asymmetrical), '10b' (lowasymmetrical), ' $11 b^{\prime}$ (symmetrical). Default value of ( $0 \times 2 \mathrm{C}$ ) orient_mode is '00b'.

Table 56: Hysteresis, Blocking for Orientation Interrupt, Register (0x2C)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| reserved | orient hyst<2> | orient hyst<1> | orient hyst<0> | orient_ blocking<1> | orient blocking<0> | $\begin{gathered} \text { orient } \\ \text { mode<1> } \end{gathered}$ | $\begin{gathered} \text { orient_ } \\ \text { mode<0> } \end{gathered}$ |

Register ( $0 \times 2 \mathrm{D}$ ) contains the definition of the theta blocking angle for the orientation interrupt. ( $0 \times 2 \mathrm{D}$ ) orient_theta defines a blocking angle between $0^{\circ}$ and $44.8^{\circ}$ as described in section "4.8.1.7 Orientation blocking". Default value of ( $0 \times 2 \mathrm{D}$ ) orient_theta is $0 \times 08$.

Table 57: Theta blocking angle, register ( $0 \times 2 \mathrm{D}$ )

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| reserved | reserved | orient <br> theta $\langle\overline{5}>$ | orient <br> theta $\langle\overline{4}>$ | orient <br> theta $\langle\overline{3}>$ | orient <br> theta $\langle\overline{2}>$ | orient <br> theta $\langle\overline{1}>$ | orient <br> theta $\langle\overline{0}>$ |

Register ( $0 \times 2 E$ ) contains the definition of the flat threshold angle for the flat interrupt. ( $0 \times 2 E$ ) flat_theta defines a blocking angle between $0^{\circ}$ and $44.8^{\circ}$ as described in section"4.8.8 Flat detection". Default value of ( $0 \times 2 E$ ) flat theta is $0 \times 08$.

Table 58: Flat threshold angle, register ( $0 \times 2 \mathrm{E}$ )

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| reserved | reserved | flat <br> theta<5> | flat <br> theta<4> | flat <br> theta<3> | flat <br> theta<2> | flat <br> theta<1> | flat <br> theta<0> |

Register ( $\mathbf{0 x 2 F}$ ) contains the definition of the flat hold time. ( $0 \times 2 F$ ) flat_hold_time defines the time a new flat value has to be at least stable for before the interrupt is generated. The settings for (0x2F) flat_hold_time are '00b' (0), '01b' (512 ms), '10b' (1024 ms), '11b' (2048 ms).
Default value of (0x2F) flat_hold_time is '01b'.

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Table 59: Flat threshold angle, register (0x2F)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| reserved | reserved | flat_hold_- <br> time<1> | flat_hold_ <br> time<0> | reserved | reserved | reserved | reserved |

Register (0x30) and (0x31) are reserved.

### 5.12 Self-test

Register (0x32) contains the settings for the activation of the sensor self-test.
(0x32) self_test_sign sets the sign of the electrostatic excitation. The settings for (0x32) self_test_sign are ' 0 ' (positive sign) and ' 1 ' (negative sign). Default value of (0x32) self_test_sign is ' 0 '.
(0x32) self_test_axis defines the axis which shall be excited. Only one axis can be excited at the same time. The settings for (0x32) self_test_axis are '00b' (no self-test), '01' (x-axis), '10' ( $y$-axis), and ' 11 ' ( $z$-axis). Default value of (0x32) self_test_axis is '00b'.

Table 60: Sensor self-test, register (0x32)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| reserved | reserved | reserved | reserved | reserved | self_test <br> _sign | self_test <br> _axis $<1>$ | self_test <br> _axis $<0>$ |

### 5.13 Non-volatile memory control (EEPROM control)

Register (0x33) contains the control settings for the non-volatile memory (EEPROM). (0x33) nvm_load is used to perform a user-defined image update. Writing ' 1 ' ( $0 \times 33$ ) nvm_load starts the update procedure. The value ' 1 ' is kept as long as the update procedure runs, afterwards it is reset to ' 0 '.
(0x33) nvm_rdy contains the status of writing the EEPROM. (0x33) nvm_rdy is '0' as long as writing the EEPROM endures, it is ' 1 ' if currently no write access is performed and, therefore, a new write access can be initiated.

Writing '1'to (0x33) nvm_prog_trig triggers writing the EEPROM. The EEPROM can only be written if it was unlocked before.

Writing ' 1 'to ( $0 \times 33$ ) nvm_prog_mode unlocks the EEPROM.

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Table 61: EEPROM control settings, register ( $0 \times 33$ )

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| reserved | reserved | reserved | reserved | nvm_load | nvm_rdy | nvm_prog <br> _trig | nvm_prog <br> _mode |

### 5.14 Interface configuration

Register (0x34) contains the settings for the digital interfaces. Writing '1'to (0x34) i2c_wdt_en enables the watchdog at the SDI pin (= SDA for $I^{2} \mathrm{C}$ ) if $I^{2} \mathrm{C}$ is selected. Default value of (0x34) i2c_wdt_en is ' 0 '.
(0x34) i2c_wdt_sel selects the $1^{2} \mathrm{C}$ data pad watchdog timer period. The settings for (0x34) $i 2 c \_w d t \_s e l$ are ' 0 ' ( 1 ms ) and ' 1 ' ( 50 ms ). Default value of ( $0 \times 34$ ) i2c_wdt_sel is ' 0 '.
(0x34) spi3 selects the SPI mode. The settings for (0x34) spi3 are ' 0 ' ( 4 -wire SPI) and '1' (3wire SPI). Default value of ( $0 \times 34$ ) spi3 is ' 0 '.

Table 62: EEPROM control settings, register (0x34)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| reserved | reserved | reserved | reserved | reserved | i2c_wdt <br> _en | i2c_wdt <br> _sel | spi3 |

Register (0x35) is reserved.

### 5.15 Offset compensation

Register (0x36) contains settings for the offset compensation in general, for fast offset compensation, and for slow offset compensation. Writing '1'to (0x36) offset_reset sets all offset compensation registers ( $0 \times 38$ to $0 \times 3 D$ ) to zero.

Default value of ( $0 \times 36$ ) offset_reset is ' 0 '.
(0x36) cal_trigger starts the fast compensation process for the specified axis. The settings for (0x36) cal_trigger are '00b' (no axis selected), '01b' (x-axis), '10b' (y-axis), '11b' (z-axis). A non-zero value is kept until the fast compensation procedure is finished. Default value of (0x36) cal_trigger is '00b'.
(0x36) cal_rdy indicates the state of the fast compensation. (0x36) cal_rdy is ' 0 ' when ( $0 \times 36$ ) cal_trigger has a nonzero value, otherwise ( $0 \times 36$ ) cal_rdy is ' 1 '.

Writing ' 1 ' ('0') to ( $0 \times 36$ ) hp_z_en enables (disables) slow offset compensation for the $z$-axis. Writing '1' ('0') to ( $0 \times 36$ ) hp_y_en enables (disables) slow offset compensation for the $y$-axis.

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Writing '1' ('0') to ( $0 \times 36$ ) hp_x_en enables (disables) slow offset compensation for the $x$-axis. Default value for each of ( $0 \times 36$ ) hp_x_en, ( $0 \times 36$ ) $h p_{-} y_{-} e n$, and ( $0 \times 36$ ) hp_x_en is ' 0 ', respectively.

Table 63: Offset compensation, fast offset compensation, register (0x36)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| offset <br> reset | cal <br> trigger<1> | cal <br> trigger <br> r | cal_rdy | reserved | hp_z_en | hp_y_en | $h p_{-} x_{-} e n$ |

Register (0x37) contains settings for the offset compensation in general, and for slow offset compensation. ( $0 \times 37$ ) offset_target_z sets the target value for the offset compensation of the $z$ axis.
(0x37) offset_target_y sets the target value for the offset compensation of the $y$-axis.
(0x37) offset_target_x sets the target value for the offset compensation of the $x$-axis.
The settings for ( $0 \times 37$ ) offset_target_x, (0x37) offset_target_y, and ( $0 \times 37$ ) offset_target_z are '00b' ( 0 g ), '01b' ( +1 g ), ' ${ }^{-10 b}$ ' $(-1 \mathrm{~g})$, and ' $11 \mathrm{~b}^{\prime}$ ' $(0 \mathrm{~g})$. Default value of each of ${ }^{-}(0 \times 37)$ offset_target_x, ( $0 \times 37$ ) offset_target_y, and ( $0 \times 37$ ) offset_target_z is '00b', respectively.
(0x37) cut_off defines the number of samples for comparison by the slow offset compensation. The settings for ( $0 x 37$ ) cut_off are ' 0 ' ( 8 samples) and ' 1 ' ( 16 samples). The default value of (0x37) cut_off is ' 0 '.

Table 64: Offset compensation, slow offset compensation, register (0x37)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| reserved | offset_tar | offset_tar | offset_tar | offset_tar | offset_tar | offset_tar | cut_off |
|  | get_z<1> | get_z<0> | get_y<1> | get_y<0> | get_ $x<1>$ | get_ $x<0>$ |  |

Register ( $0 x 38$ ) contains the compensation value for filtered data for the $x$-axis. The contents of each of the registers ( $0 \times 38$ ) to ( $0 \times 3 \mathrm{D}$ ) is added to the corresponding acceleration data; it can be set either automatically by one of the implemented compensation algorithms or manually. These registers are image registers of registers in the EEPROM; the content of the EEPROM is copied to them after every reset.

Table 65: Filtered data compensation for the x -axis, register ( $0 \times 38$ )

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| offset_ | offset |  |  |  |  |  |  |
| filt_ $x<\overline{7}>$ | offset_ |  |  |  |  |  |  |
| filt_ $x<\overline{6}>$ | offset_ |  |  |  |  |  |  |
| filt_ $x<\overline{5}>$ | offset_ <br> filt_ $x<\overline{4}>$ | offset__ <br> filt_ $x<\overline{3}>$ <br> filt_ $x<\overline{2}>$ | offset__ <br> filt_ $x<\overline{1}>$ | offset_ <br> filt_ $x<\overline{0}>$ |  |  |  |


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Register (0x39) contains the compensation value for filtered data for the $y$-axis.
Table 66: Filtered data compensation for the $y$-axis, register ( $0 \times 39$ )

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| offset filt_y<7> | offset <br> filt $y<\overline{6}>$ | offset filt_ $y<5>$ | offset filt_ $y<4>$ | offset <br> filt_ $y<3>$ | offset_ <br> filt_ $y<2$ > | offset filt_y<1> | offset filt_y<0> |

Register ( $0 \times 3 A$ ) contains the compensation value for filtered data for the $z$-axis.
Table 67: Filtered data compensation for the $\mathbf{z}$-axis, register ( $0 \times 3 A$ )

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| offset | offset_ | offset_ | offset_ | offset_ | offset_ | offset_ | offset_ |
| filt_ $z<\overline{7}>$ | filt_ $z<\overline{6}>$ | filt_ $z<\overline{5}>$ | filt_ $z<\overline{4}>$ | filt_ $z<\overline{3}>$ | filt_ $z<\overline{2}>$ | filt_ $z<\overline{1}>$ | filt_ $z<\overline{0}>$ |

Register ( $0 \times 3 B$ ) contains the compensation value for unfiltered data for the $x$-axis.
Table 68: Unfiltered data compensation for the $x$-axis, register ( $0 \times 3 B$ )

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| offset_ | offset | offset_ | offset_ | offset_ | offset_ | offset_ | offset_ |
| unfilt_x | unfilt_x | unfilt_x | unfilt_x | unfilt_x | unfilt_x | unfilt_x | unfilt_x |
| <7> | <6> | <5> | <4> | <3> | <2> | <1> | <0> |

Register ( $0 \times 3 C$ ) contains the compensation value for unfiltered data for the $y$-axis.
Table 69: Unfiltered data compensation for the x -axis, register ( $0 \times 3 \mathrm{C}$ )

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| offset_ | offset_ | offset_ | offset_ | offset_ | offset_ | offset_ | offset_ |
| unfilt_y | unfilt_y | unfilt_y | unfilt_y | unfilt_y | unfilt_y | unfilt_y | unfilt_y |
| <7> | <6> | <5> | <4> | <3> | <2> | <1> | <0> |

Register ( $0 \times 3 D$ ) contains the compensation value for unfiltered data for the $z$-axis.

Table 70: Unfiltered data compensation for the $y$-axis, register (0x3D)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| offset_ | offset_ | offset_ | offset_ | offset_ | offset_ | offset_ | offset_ |
| unfil_z | unfilt_z | unfilt_z | unfilt_z | unfilt_z | unfilt_z | unfilt_z | unfilt_z |
| <7> | <6> | <5> | <4> | <3> | <2> | <1> | <0> |


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Registers ( $0 \times 3 E$ ) and ( $0 \times 3 F$ ) are image registers of registers in the EEPROM. They are not linked to any sensor-specific functionality.

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## 6. Digital interfaces

The BMA250 supports two serial digital interface protocols for communication as a slave with a host device (when operating in general mode): SPI and $I^{2} \mathrm{C}$. The active interface is selected by the state of the Pin\#11 (PS) 'protocol select' pin: '0' ('1') selects SPI ( $\left.I^{2} \mathrm{C}\right)$. For details see section 4.2 Operational modes.

By default, SPI operates in the standard 4-wire configuration. It can be re-configured by software to work in 3 -wire mode instead of standard 4 -wire mode.

Both interfaces share the same pins. The mapping for each interface is given in the following table:

Table 71: Mapping of the interface pins

| Pin\# | Name | use w/ <br> SPI | use w/ <br> $\mathbf{I}^{2} C$ | Description |
| :---: | :--- | :--- | :--- | :--- |
| 1 | SDO | SDO | address | SPI: Data Output (4-wire mode) <br> $I^{2} \mathrm{C}: ~ U s e d ~ t o ~ s e t ~ L S B ~ o f ~$ <br> $I^{2} \mathrm{C}$ <br> address |
| 2 | SDx | SDI | SDA | SPI: Data Input (4-wire mode) Data Input/ Output (3-wire mode) <br> $I^{2} \mathrm{C}:$ Serial Data |
| 10 | CSB | CSB | unused | Chip Select (enable) |
| 12 | SCx | SCK | SCL | SPI: Serial Clock <br> $I^{2} \mathrm{C}: ~ S e r i a l ~ C l o c k ~$ |

The following table shows the electrical specifications of the interface pins:

Table 72: Electrical specification of the interface pins

| Parameter | Symbol | Condition | Min | Typ | Max | Units |
| :--- | :---: | :--- | :---: | :---: | :---: | :---: |
| PS Impedance <br> for Tri-state Detection | $\mathrm{R}_{\mathrm{TS}}$ |  | 1 |  |  | $\mathrm{M} \Omega$ |
|  | $\mathrm{C}_{\mathrm{TS}}$ |  |  |  | 10 | pF |
| PS Impedance <br> for Non-Tri-state | $\mathrm{R}_{\text {NTS }}$ |  |  | 5 | $\mathrm{k} \Omega$ |  |
| Pull-up Resistance | $\mathrm{R}_{\text {up }}$ | Internal Pull-up <br> Resistance to VDDIO | 70 | 120 | 190 | $\mathrm{k} \Omega$ |
| Pull-down Resistance | $\mathrm{R}_{\text {down }}$ | Internal Pull-down <br> Resistance to GND | 12 | 20 | 32 | $\mathrm{k} \Omega$ |
| Input Capacitance | $\mathrm{C}_{\text {in }}$ |  |  | 5 | 10 | pF |
| In $^{2} \mathrm{C}$ Bus Load <br> Capacitance (max. <br> drive capability) | $\mathrm{C}_{\text {I2C_Load }}$ |  |  |  | 400 | pF |


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| :--- | :---: | :--- |

### 6.1 Serial peripheral interface (SPI)

The timing specification for SPI of the BMA250 is given in the following table:

Table 73: SPI timing

| Parameter | Symbol | Condition | Min | Max | Units |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Clock Frequency | $\mathrm{f}_{\text {SPI }}$ | Max. Load on SDI or <br> SDO $=25 \mathrm{pF}$ |  | 10 | MHz |
| SCK Low Pulse | $\mathrm{t}_{\text {SCKL }}$ |  | 20 |  | ns |
| SCK High Pulse | $\mathrm{t}_{\text {SCKH }}$ |  | 20 |  | ns |
| SDI Setup Time | $\mathrm{t}_{\text {SDI setup }}$ |  | 20 |  | ns |
| SDI Hold Time | $\mathrm{t}_{\text {SDI_hold }}$ |  | 20 |  | ns |
| SDO Output Delay | $\mathrm{t}_{\text {SDO_OD }}$ |  | Load $=25 \mathrm{pF}$ <br>  | Load $=250 \mathrm{pF}$, <br> $\mathrm{V}_{\text {DDIO }}=2.4 \mathrm{~V}$ |  |
| 30 | ns |  |  |  |  |
| CSB Setup Time | $\mathrm{t}_{\text {CSB setup }}$ |  | 40 | ns |  |
| CSB Hold Time | $\mathrm{t}_{\text {CSB_hold }}$ |  | 20 |  | ns |

The following figure shows the definition of the SPI timings given in table 73:


Figure 10: SPI timing diagram

The SPI interface of the BMA250 is compatible with two modes, '00' and ' 11 '. The automatic selection between $\left[\mathrm{CPOL}={ }^{\prime} 0\right.$ ' and $\mathrm{CPHA}=$ ' 0 '] and $[\mathrm{CPOL}=' 1$ ' and $\mathrm{CPHA}=$ ' 1 '] is done based on the value of SCK after a falling edge of CSB.

Two configurations of the SPI interface are supported by the BMA250: 4 -wire and 3 -wire. The same protocol is used by both configurations. The device operates in 4 -wire configuration by default. It can be switched to 3 -wire configuration by writing ' 1 ' to ( $0 \times 34$ ) spi3. Pin SDI is used as the common data pin in 3 -wire configuration.

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| :--- | :---: | :---: |

For single byte read as well as write operations, 16 -bit protocols are used. The BMA250 also supports multiple-byte read operations.

In SPI 4-wire configuration CSB (chip select low active), SCK (serial clock), SDI (serial data input), and SDO (serial data output) pins are used. The communication starts when the CSB is pulled low by the SPI master and stops when CSB is pulled high. SCK is also controlled by SPI master. SDI and SDO are driven at the falling edge of SCK and should be captured at the rising edge of SCK.

The basic write operation waveform for 4 -wire configuration is depicted in figure 11. During the entire write cycle SDO remains in high- impedance state.


Figure 11: 4-wire basic SPI write sequence (mode '11')

The basic read operation waveform for 4-wire configuration is depicted in figure 12:


Figure 12: 4 -wire basic SPI read sequence (mode '11')

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| :--- | :--- | :--- |

The data bits are used as follows:
Bit0: Read/Write bit. When 0 , the data SDI is written into the chip. When 1 , the data SDO from the chip is read.

Bit1-7: Address AD(6:0).
Bit8-15: when in write mode, these are the data SDI, which will be written into the address. When in read mode, these are the data SDO, which are read from the address.

Multiple read operations are possible by keeping CSB low and continuing the data transfer. Only the first register address has to be written. Addresses are automatically incremented after each read access as long as CSB stays active low.

The principle of multiple read is shown in figure 13:

|  | Control byte |  |  |  |  |  |  |  | Data byte |  |  |  |  |  |  |  | Data byte |  |  |  |  |  |  |  | Data byte |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Start | RW | Register adress (02h) |  |  |  |  |  |  | Data register - adress 02h |  |  |  |  |  |  |  | Data register - adress 03h |  |  |  |  |  |  |  | Data register - adress 04h |  |  |  |  |  |  |  | Stop |
| CSB | 1 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | X | X | X | X | X | X | X | X | X | X | X | X | X | X | X |  | X |  |  |  |  |  |  |  | CSB |
| = |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | X |  | X | X | X | X | X | X | X | $=$1 |
| 0 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

Figure 13: SPI multiple read
In SPI 3-wire configuration CSB (chip select low active), SCK (serial clock), and SDI (serial data input and output) pins are used. The communication starts when the CSB is pulled low by the SPI master and stops when CSB is pulled high. SCK is also controlled by SPI master. SDI is driven (when used as input of the device) at the falling edge of SCK and should be captured (when used as the output of the device) at the rising edge of SCK.

The protocol as such is the same in 3 -wire configuration as it is in 4 -wire configuration. The basic operation waveform (read or write access) for 3 -wire configuration is depicted in figure 14:


Figure 14: 3-wire basic SPI read or write sequence (mode '11')

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| :--- | :---: | :---: |

### 6.2 Inter-Integrated Circuit ( ${ }^{2} \mathrm{C}$ )

The $I^{2} C$ bus uses SCL (= SCx pin, serial clock) and SDA (= SDx pin, serial data input and output) signal lines. Both lines are connected to $\mathrm{V}_{\text {DDIO }}$ externally via pull-up resistors so that they are pulled high when the bus is free.

The $I^{2} \mathrm{C}$ interface of the BMA250 is compatible with the $\mathrm{I}^{2} \mathrm{C}$ Specification UM10204 Rev. 03 (19 June 2007), available at http://www.nxp.com. The BMA250 supports $I^{2} \mathrm{C}$ standard mode and fast mode, only 7 -bit address mode is supported. For $\mathrm{V}_{\text {DDIO }}=1.2 \mathrm{~V}$ to 1.8 V the guaranteed voltage output levels are slightly relaxed as described in the Parameter Specification (table 1).

The default $I^{2} \mathrm{C}$ address of the device is 0011000 b ( $0 \times 18$ ). It is used if the SDO pin is pulled to 'GND'. The alternative address 0011001 b ( $0 \times 19$ ) is selected by pulling the SDO pin to ' $\mathrm{V}_{\text {DDIO }}$ '.

The timing specification for $I^{2} \mathrm{C}$ of the BMA250 is given in table 74:

Table 74: ${ }^{2} \mathrm{C}$ timings

| Parameter | Symbol | Condition | Min | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Clock Frequency | $\mathrm{f}_{\text {SCL }}$ |  |  | 400 | kHz |
| SCL Low Period | tow |  | 1.3 |  | $\mu \mathrm{S}$ |
| SCL High Period | $\mathrm{t}_{\text {HIGH }}$ |  | 0.6 |  |  |
| SDA Setup Time | $\mathrm{t}_{\text {SUDAT }}$ |  | 0.1 |  |  |
| SDA Hold Time | $\mathrm{t}_{\text {HDDAT }}$ |  | 0.0 |  |  |
| Setup Time for a repeated Start Condition | $\mathrm{t}_{\text {SUSTA }}$ |  | 0.6 |  |  |
| Hold Time for a Start Condition | $\mathrm{t}_{\text {HDSTA }}$ |  | 0.6 |  |  |
| Setup Time for a Stop Condition | $\mathrm{t}_{\text {SUSTO }}$ |  | 0.6 |  |  |
| Time before a new Transmission can start | $\mathrm{t}_{\text {BUF }}$ |  | 1.3 |  |  |


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Figure 15 shows the definition of the $1^{2} \mathrm{C}$ timings given in table 74 :


Figure 15: $I^{2} \mathrm{C}$ timing diagram
The $\mathrm{I}^{2} \mathrm{C}$ protocol works as follows:
START: Data transmission on the bus begins with a high to low transition on the SDA line while SCL is held high (start condition (S) indicated by ${ }^{2} \mathrm{C}$ bus master). Once the START signal is transferred by the master, the bus is considered busy.

STOP: Each data transfer should be terminated by a Stop signal (P) generated by master. The STOP condition is a low to HIGH transition on SDA line while SCL is held high.

ACK: Each byte of data transferred must be acknowledged. It is indicated by an acknowledge bit sent by the receiver. The transmitter must release the SDA line (no pull down) during the acknowledge pulse while the receiver must then pull the SDA line low so that it remains stable low during the high period of the acknowledge clock cycle.

In the following diagrams these abbreviations are used:

| S | Start |
| :--- | :--- |
| P | Stop |
| ACKS | Acknowledge by slave |
| ACKM | Acknowledge by master |
| NACKM | Not acknowledge by master |
| RW | Read / Write |

A START immediately followed by a STOP (without SCK toggling from logic " 1 " to logic " 0 ") is not supported. If such a combination occurs, the STOP is not recognized by the device.

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| :--- | :--- | :--- |

## $1^{2} \mathrm{C}$ write access:

$1^{2} \mathrm{C}$ write access can be used to write a data byte in one sequence.
The sequence begins with start condition generated by the master, followed by 7 bits slave address and a write bit ( $\mathrm{RW}=0$ ). The slave sends an acknowledge bit $(\mathrm{ACK}=0)$ and releases the bus. Then the master sends the one byte register address. The slave again acknowledges the transmission and waits for the 8 bits of data which shall be written to the specified register address. After the slave acknowledges the data byte, the master generates a stop signal and terminates the writing protocol.

Example of an $I^{2} \mathrm{C}$ write access:


Figure 16: $I^{2} \mathrm{C}$ write

## $1^{2} \mathrm{C}$ read access:

$1^{2} \mathrm{C}$ read access also can be used to read one or multiple data bytes in one sequence.
A read sequence consists of a one-byte $I^{2} \mathrm{C}$ write phase followed by the $I^{2} \mathrm{C}$ read phase. The two parts of the transmission must be separated by a repeated start condition ( Sr ). The $I^{2} \mathrm{C}$ write phase addresses the slave and sends the register address to be read. After slave acknowledges the transmission, the master generates again a start condition and sends the slave address together with a read bit $(R W=1)$. Then the master releases the bus and waits for the data bytes to be read out from slave. After each data byte the master has to generate an acknowledge bit $(A C K=0)$ to enable further data transfer. A NACKM (ACK = 1) from the master stops the data being transferred from the slave. The slave releases the bus so that the master can generate a STOP condition and terminate the transmission.
The register address is automatically incremented and, therefore, more than one byte can be sequentially read out. Once a new data read transmission starts, the start address will be set to the register address specified in the latest $I^{2} \mathrm{C}$ write command. By default the start address is set at $0 \times 00$. In this way repetitive multi-bytes reads from the same starting address are possible.

In order to prevent the $I^{2} \mathrm{C}$ slave of the device to lock-up the $I^{2} \mathrm{C}$ bus, a watchdog timer (WDT) is implemented. The WDT observes internal $I^{2} \mathrm{C}$ signals and resets the $I^{2} \mathrm{C}$ interface if the bus is locked-up by the BMA250. The activity and the timer period of the WDT can be configured through the bits ( $0 \times 34$ ) i2c_wdt_en and (0x34) i2c_wdt_sel.

Writing ' 1 ' ('0') to ( $0 x 34$ ) i2c_wdt_en activates (de-activates) the WDT. Writing ' 0 ' ('1') to (0x34) i2c_wdt_se selects a timer period of $1 \mathrm{~ms}(50 \mathrm{~ms})$.

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| :--- | :---: | :--- |

Example of an $I^{2} \mathrm{C}$ read access:


Figure 17: ${ }^{12} \mathrm{C}$ multiple read

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| :--- | :---: | :---: |

## 7. Pin-out and connection diagram

### 7.1 Pin-out



Figure 18: Pin-out top view


Figure 19 Pin-out bottom view

Table 75: Pin description

| Pin\# | Name | I/O Type | Description | Connect to |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | in SPI 4W | In SPI 3W | in ${ }^{2} \mathrm{C}$ |
| 1 | SDO | Digital out | Serial data output in SPI Address select in $\mathrm{I}^{2} \mathrm{C}$ mode see chapter 6.2 | SDO | DNC (float) | GND for default addr. |
| 2 | SDx | Digital I/O | SDA serial data $1 / \mathrm{O}$ in $\mathrm{I}^{2} \mathrm{C}$ SDI serial data input in SPI 4W SDA serial data I/O in SPI 3W | SDI | SDA | SDA |
| 3 | VDDIO | Supply | Digital I/O supply voltage (1.2V ... 3.6V) | $\mathrm{V}_{\text {DIIO }}$ | $\mathrm{V}_{\text {DIIO }}$ | $\mathrm{V}_{\text {DDIO }}$ |
| 4 | NC | -- |  | GND | GND | GND |
| 5 | INT1 | Digital out | Interrupt output 1 | INT1 | INT1 | INT1 |
| 6 | INT2 | Digital out | Interrupt output 2 | INT2 | INT2 | INT2 |
| 7 | VDD | Supply | Power supply for analog \& digital domain (1.62V ... 3.6V) | $V_{\text {D }}$ | $\mathrm{V}_{\mathrm{DD}}$ | $\mathrm{V}_{\mathrm{DD}}$ |
| 8 | GNDIO | Ground | Ground for l/O | GND | GND | GND |
| 9 | GND | Ground | Ground for digital \& analog | GND | GND | GND |
| 10 | CSB | Digital in | Chip select for SPI mode | CSB | CSB | DNC (float) |
| 11 | PS | Digital in | Protocol select (GND = SPI, $V_{\text {DDIO }}=I^{2}$ C, float $=\mu \mathrm{C}$-less). Pin must not float unless dedicated mode is used, see chapter 4.2.2 | GND | GND | $\mathrm{V}_{\text {DDIO }}$ |
| 12 | SCx | Digital in | SCK for SPI serial clock SCL for $\mathrm{I}^{2} \mathrm{C}$ serial clock | SCK | SCK | SCL |


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| :--- | :---: | :--- |

### 7.2 Connection diagram 4-wire SPI



Figure 20: 4-wire SPI connection

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| :--- | :---: | :--- |

### 7.3 Connection diagram 3-wire SPI



Figure 21: 3-wire SPI connection

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| :--- | :---: | :--- |

### 7.4 Connection diagram $\mathrm{I}^{2} \mathrm{C}$



Figure 22: $\mathrm{I}^{2} \mathrm{C}$ connection

Note: the recommended value for $\mathrm{C}_{1}, \mathrm{C}_{2}$ is 100 nF .

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| :--- | :---: | :---: |

## 8. Package

### 8.1 Outline dimensions

The sensor housing is a standard LGA package. It is compliant with JEDEC Standard MO-229 Type VGGD-3. Its dimensions are the following.


Figure 23: Package outline dimensions

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| :--- | :---: | :---: |

### 8.2 Sensing axes orientation

If the sensor is accelerated in the indicated directions, the corresponding channel will deliver a positive acceleration signal (dynamic acceleration). If the sensor is at rest and the force of gravity is acting along the indicated directions, the output of the corresponding channel will be negative (static acceleration).

Example: If the sensor is at rest or at uniform motion in a gravity field according to the figure given below, the output signals are:

- $\quad \pm 0 \mathrm{~g}$ for the X channel
- $\pm 0 \mathrm{~g}$ for the Y channel
- $\quad+1 \mathrm{~g}$ for the Z channel


Figure 24: Orientation of sensing axis

The following table lists all corresponding output signals on $\mathrm{X}, \mathrm{Y}$, and Z while the sensor is at rest or at uniform motion in a gravity field under assumption of a $\pm 2 \mathrm{~g}$ range setting and a top down gravity vector as shown above.

Table 76: Output signals depending on sensor orientation

| Sensor Orientation (gravity vector $\downarrow$ ) |  | $\square$ | $\square$ | $\square$ | upright | 1Чठె! |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Output Signal X | 0g / OLSB | $1 \mathrm{~g} / 256 \mathrm{LSB}$ | 0g / OLSB | -1g/-256LSB | Og / OLSB | 0g / OLSB |
| Output Signal Y | -1g/-256LSB | 0g / OLSB | +1g / 256LSB | 0g / OLSB | 0g / OLSB | 0g / OLSB |
| Output Signal Z | 0g / OLSB | 0g / OLSB | 0g / OLSB | 0g / OLSB | 1g/256LSB | -1g/-256LSB |


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| :--- | :---: | :--- |

### 8.3 Landing pattern recommendation

For the design of the landing patterns, we recommend the following dimensioning:


Figure 25: Landing patterns relative to the device pins, dimensions are in mm

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| :--- | :---: | :---: |

### 8.4 Marking

### 8.4.1 Mass production samples

Table 77: Marking of mass production samples

| Labeling | Name | Symbol | Remark |
| :---: | :---: | :---: | :---: |
|  | Lot counter | CCC | 3 alphanumeric digits, variable to generate mass production trace-code |
|  | Product number | T | 1 alphanumeric digit, fixed to identify product type, $\mathrm{T}=$ " 8 " |
|  | Sub-con ID | L | 1 alphanumeric digit, variable to identify sub-con ( $L=$ " $A$ " or $L=$ "U" or $L=$ " $P$ ") |
|  | Pin 1 identifier | - | -- |

### 8.4.2 Engineering samples

Table 78: Marking of engineering samples

| Labeling | Name | Symbol | Remark |
| :---: | :---: | :---: | :---: |
| $\begin{array}{r} \text { XXN } \\ \text { C } \end{array}$ | Eng. sample ID | N | 1 alphanumeric digit, fixed to identify engineering sample, $\mathrm{N}=$ "e" |
|  | Sample ID | XX | 2 alphanumeric digits, variable <br> to generate trace-code |
|  | Counter ID | CC | 2 alphanumeric digits, variable <br> to generate trace-code |
|  | Pin 1 identifier | - | -- |


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| :--- | :---: | :---: |

### 8.5 Soldering guidelines

The moisture sensitivity level of the BMA250 sensors corresponds to JEDEC Level 1, see also

- IPC/JEDEC J-STD-020C "Joint Industry Standard: Moisture/Reflow Sensitivity Classification for non-hermetic Solid State Surface Mount Devices"
- IPC/JEDEC J-STD-033A "Joint Industry Standard: Handling, Packing, Shipping and Use of Moisture/Reflow Sensitive Surface Mount Devices".

The sensor fulfils the lead-free soldering requirements of the above-mentioned IPC/JEDEC standard, i.e. reflow soldering with a peak temperature up to $260^{\circ} \mathrm{C}$.

| Profile Feature |  | Pb -Free Assembly |
| :---: | :---: | :---: |
| Average Ramp-Up Rate $\left(T s_{\max } \text { to } T p\right)$ |  | $3^{\circ} \mathrm{C} /$ second max. |
| $\begin{gathered} \text { Preheat } \\ \text { - Temperature } \operatorname{Min}\left(\mathrm{Ts}_{\min }\right) \\ \text { - Temperature } \operatorname{Max}\left(\mathrm{Ts}_{\max }\right) \\ \text { - Time }\left(\mathrm{t}_{\min } \text { to } \mathrm{t} \mathrm{~s}_{\max }\right) \\ \hline \end{gathered}$ |  | $\begin{gathered} 150^{\circ} \mathrm{C} \\ 200^{\circ} \mathrm{C} \\ 60-180 \text { seconds } \end{gathered}$ |
| Time maintained above: <br> - Temperature ( $\mathrm{T}_{\mathrm{L}}$ ) <br> - Time ( $\mathrm{t}_{\mathrm{L}}$ ) |  | $\begin{gathered} 217^{\circ} \mathrm{C} \\ 60-150 \text { seconds } \end{gathered}$ |
| Peak/Classification Temperature (Tp) |  | $260{ }^{\circ} \mathrm{C}$ |
| Time within $5{ }^{\circ} \mathrm{C}$ of actual Peak Temperature (tp) |  | 20-40 seconds |
| Ramp-Down Rate |  | $6^{\circ} \mathrm{C} /$ second max. |
| Time $25^{\circ} \mathrm{C}$ to Peak Temperature |  | 8 minutes max. |

Note 1: All temperatures refer to topside of the package, measured on the package body surface.


Figure 26: Soldering profile

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| :--- | :---: | :---: |

### 8.6 Handling instructions

Micromechanical sensors are designed to sense acceleration with high accuracy even at low amplitudes and contain highly sensitive structures inside the sensor element. The MEMS sensor can tolerate mechanical shocks up to several thousand g's. However, these limits might be exceeded in conditions with extreme shock loads such as e.g. hammer blow on or next to the sensor, dropping of the sensor onto hard surfaces etc.

We recommend to avoid g-forces beyond the specified limits during transport, handling and mounting of the sensors in a defined and qualified installation process.

This device has built-in protections against high electrostatic discharges or electric fields (e.g. 2 kV HBM); however, anti-static precautions should be taken as for any other CMOS component. Unless otherwise specified, proper operation can only occur when all terminal voltages are kept within the supply voltage range. Unused inputs must always be tied to a defined logic voltage level.

### 8.7 Tape and reel specification

The BMA250 is shipped in a standard cardboard box.
The box dimension for 1 reel is: $L \times W \times H=35 \mathrm{~cm} \times 35 \mathrm{~cm} \times 6 \mathrm{~cm}$ BMA250 quantity: 10,000 pcs per reel, please handle with care.

| $A_{0}$ | 2,20 | $+/-0,05$ |
| :--- | :---: | :---: |
| $B_{0}$ | 2,20 | $+/-0,05$ |
| $K_{0}$ | 1,15 | $+/-0,1$ |
| $F$ | 5,50 | $+/-0,1$ |
| $P_{1}$ | 4,00 | $+/-0,1$ |
| $W$ | 12,00 | $+/-0,3$ |


(1) Measured from axtrinelhe of spradiet hole
(II) Cupulative talerance of 18 sprocket
hales $15 \pm \|$,
(1). heasured tran centrellne of sprocket
(IV) Dther mentere af pocket
ALL DNENSIDS ©N MHLIMETRES UNLESS GTHERWTSE STATER

Figure 27: Tape and reel dimensions in mm

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| :--- | :---: | :--- |

### 8.7.1 Orientation within the reel



Figure 28: Orientation of the BMA250 devices relative to the tape

### 8.8 Environmental safety

The BMA250 sensor meets the requirements of the EC restriction of hazardous substances (RoHS) directive, see also:

Directive 2002/95/EC of the European Parliament and of the Council of 27 January 2003 on the restriction of the use of certain hazardous substances in electrical and electronic equipment.

### 8.8.1 Halogen content

Results of chemical analysis indicate that the BMA250 contains less than 900ppm (by weight) of Fluorine, Chlorine, lodine and Bromine (i.e. < 50 ppm per each substance). Therefore the BMA250 can be regarded as halogen-free. For more details on the analysis results please contact your Bosch Sensortec representative.

### 8.8.2 Internal package structure

Within the scope of Bosch Sensortec's ambition to improve its products and secure the mass product supply, Bosch Sensortec qualifies additional sources (e.g. $2^{\text {nd }}$ source) for the LGA package of the BMA250.

While Bosch Sensortec took care that all of the technical packages parameters are described above are $100 \%$ identical for all sources, there can be differences in the chemical content and the internal structural between the different package sources.

However, as secured by the extensive product qualification process of Bosch Sensortec, this has no impact to the usage or to the quality of the BMA250 product.

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| :--- | :---: | :---: |

## 9. Legal disclaimer

### 9.1 Engineering samples

Engineering Samples are marked with "e". Samples may vary from the valid technical specifications of the product series contained in this data sheet. They are therefore not intended or fit for resale to third parties or for use in end products. Their sole purpose is internal client testing. The testing of an engineering sample may in no way replace the testing of a product series. Bosch Sensortec assumes no liability for the use of engineering samples. The Purchaser shall indemnify Bosch Sensortec from all claims arising from the use of engineering samples.

### 9.2 Product use

Bosch Sensortec products are developed for the consumer goods industry. They may only be used within the parameters of this product data sheet. They are not fit for use in life-sustaining or security sensitive systems. Security sensitive systems are those for which a malfunction is expected to lead to bodily harm or significant property damage. In addition, they are not fit for use in products which interact with motor vehicle systems.

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The purchaser must monitor the market for the purchased products, particularly with regard to product safety, and inform Bosch Sensortec without delay of all security relevant incidents.

### 9.3 Application examples and hints

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## 10. Document history and modification

| Revision | Chapter | Description of modification/changes | Date |
| :--- | :--- | :--- | :--- |
| 0.8 |  | Document release | 17 December 2010 |
| 0.9 | 1 | Update table 1 | 26 January 2011 |
|  | 4.2 .2 | Added missing table numbers |  |
|  | 4.3 | Update table 7 |  |
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