

Atmel ATA6622/ATA6624/ATA6626 Atmel ATA6622C/ATA6624C/ATA6626C

LIN Bus Transceiver with 3.3V (5V) Regulator and Watchdog

DATASHEET

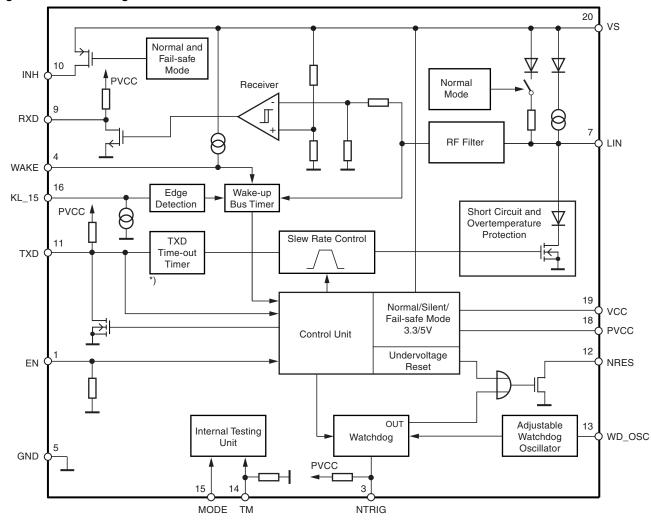
Features

- Master and slave operation possible
- Supply voltage up to 40V
- Operating voltage V_S = 5V to 27V
- Typically 10μA supply current during Sleep Mode
- Typically 57μA supply current in Silent Mode
- Linear low-drop voltage regulator, 85mA current capability:
- Normal, Fail-safe, and Silent Mode
- Atmel ATA6622 VCC = 3.3V ±2%
- Atmel ATA6624 VCC = 5.0V ±2%
- Atmel ATA6626 VCC = 5.0V ±2%, TXD time-out timer disabled
- In Sleep Mode VCC is switched off
- VCC- undervoltage detection (4ms reset time) and watchdog reset logical combined at open drain output NRES
- Negative trigger input for watchdog
- Boosting the voltage regulator possible with an external NPN transistor
- LIN physical layer according to LIN 2.0, 2.1 and SAEJ2602-2
- Wake-up capability via LIN-bus, wake pin, or KI_15 pin
- INH output to control an external voltage regulator or to switch off the master pull up resistor
- TXD time-out timer; Atmel ATA6626: TXD time-out timer Is disabled
- Bus pin is overtemperature and short circuit protected versus GND and battery
- Adjustable watchdog time via external resistor
- Advanced EMC and ESD performance
- Fulfills the OEM "Hardware Requirements for LIN in automotive Applications Rev.1.0"
- Interference and damage protection according to ISO7637
- Package: QFN 5mm x 5mm with 20 pins

1. Description

The Atmel® ATA6622 is a fully integrated LIN transceiver, which complies with the LIN 2.0, 2.1 and SAEJ2602-2 specifications. It has a low-drop voltage regulator for 3.3V/85mA output and a window watchdog. The Atmel ATA6624 has the same functionality as the Atmel ATA6622; however, it uses a 5V/85mA regulator. The Atmel ATA6626 has the same functionality as Atmel ATA6624 without a TXD time-out timer. The voltage regulator is able to source 85mA, but the output current can be boosted by using an external NPN transistor. This chip combination makes it possible to develop inexpensive, simple, yet powerful slave and master nodes for LIN-bus systems. Atmel ATA6622/ATA6624/ATA6626 are designed to handle the low-speed data communication in vehicles, e.g., in convenience electronics. Improved slope control at the LIN-driver ensures secure data communication up to 20kBaud. Sleep Mode and Silent Mode guarantee very low current consumption. The Atmel ATA6626 is able to switch the LIN unlimited to dominant level via TXD for low data rates.

Figure 1-1. Block Diagram



*) Not in ATA6626



2. Pin Configuration

Figure 2-1. Pinning QFN20

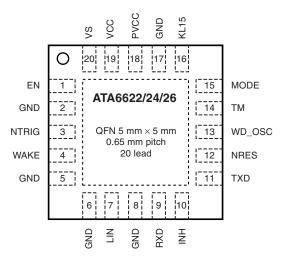


Table 2-1. Pin Description

Pin	Symbol	Function
1	EN	Enables the device in Normal Mode
2	GND	System ground (optional)
3	NTRIG	Low-level watchdog trigger input from microcontroller
4	WAKE	High-voltage input for local wake-up request; if not needed, connect directly to VS
5	GND	System ground (mandatory)
6	GND	System ground (optional)
7	LIN	LIN-bus line input/output
8	GND	System ground (optional)
9	RXD	Receive data output
10	INH	Battery related output for controlling an external voltage regulator
11	TXD	Transmit data input; active low output (strong pull down) after a local wake-up request
12	NRES	Output undervoltage and watchdog reset (open drain)
13	WD_OSC	External resistor for adjustable watchdog timing
14	TM	For factory testing only (tie to ground)
15	MODE	Low, watchdog is on; high, watchdog is off
16	KL_15	Ignition detection (edge sensitive)
17	GND	System ground (optional)
18	PVCC	3.3V/5V regulator sense input pin
19	VCC	3.3V/5V regulator output/driver pin
20	VS	Battery supply
Backside		Heat slug is connected to all GND pins



3. Functional Description

3.1 Physical Layer Compatibility

Since the LIN physical layer is independent from higher LIN layers (e.g., the LIN protocol layer), all nodes with a LIN physical layer according to revision 2.x can be mixed with LIN physical layer nodes, which, according to older versions (i.e., LIN 1.0, LIN 1.1, LIN 1.2, LIN 1.3), are without any restrictions.

3.2 Supply Pin (VS)

The LIN operating voltage is V_S = 5V to 27V. An undervoltage detection is implemented to disable data transmission if V_S falls below VS_{th} < 4V in order to avoid false bus messages. After switching on VS, the IC starts in Fail-safe Mode, and the voltage regulator is switched on.

The supply current is typically 10µA in Sleep Mode and 57µA in Silent Mode.

3.3 Ground Pin (GND)

The IC does not affect the LIN Bus in the event of GND disconnection. It is able to handle a ground shift up to 11.5% of VS. The mandatory system ground is pin 5.

3.4 Voltage Regulator Output Pin (VCC)

The internal 3.3V/5V voltage regulator is capable of driving loads up to 85mA. It is able to supply the microcontroller and other ICs on the PCB and is protected against overloads by means of current limitation and overtemperature shut-down. Furthermore, the output voltage is monitored and will cause a reset signal at the NRES output pin if it drops below a defined threshold V_{thun}. To boost up the maximum load current, an external NPN transistor may be used, with its base connected to the VCC pin and its emitter connected to PVCC.

3.5 Voltage Regulator Sense Pin (PVCC)

The PVCC is the sense input pin of the 3.3V/5V voltage regulator. For normal applications (i.e., when only using the internal output transistor), this pin is connected to the VCC pin. If an external boosting transistor is used, the PVCC pin must be connected to the output of this transistor, i.e., its emitter terminal.

3.6 Bus Pin (LIN)

A low-side driver with internal current limitation and thermal shutdown and an internal pull-up resistor compliant with the LIN 2.x specification are implemented. The allowed voltage range is between –27V and +40V. Reverse currents from the LIN bus to VS are suppressed, even in the event of GND shifts or battery disconnection. LIN receiver thresholds are compatible with the LIN protocol specification. The fall time from recessive to dominant bus state and the rise time from dominant to recessive bus state are slope controlled.

3.7 Input/Output Pin (TXD)

In Normal Mode the TXD pin is the microcontroller interface used to control the state of the LIN output. TXD must be pulled to ground in order to have a low LIN-bus. If TXD is high or unconnected (internal pull-up resistor), the LIN output transistor is turned off, and the bus is in recessive state. During Fail-safe Mode, this pin is used as output. It is current-limited to < 8mA. and is latched to low if the last wake-up event was from pin WAKE or KL_15.

3.8 TXD Dominant Time-out Function

The TXD input has an internal pull-up resistor. An internal timer prevents the bus line from being driven permanently in dominant state. If TXD is forced to low for longer than $t_{DOM} > 6ms$, the LIN-bus driver is switched to recessive state.

To reactivate the LIN bus driver, switch TXD to high (> 10µs).

The time-out function is disabled in the ATA6626. Switching to dominant level on the LIN bus occurs without any time limitations.



3.9 Output Pin (RXD)

This output pin reports the state of the LIN-bus to the microcontroller. LIN high (recessive state) is reported by a high level at RXD; LIN low (dominant state) is reported by a low level at RXD. The output has an internal pull-up resistor with typically $5k\Omega$ to VCC. The AC characteristics can be defined with an external load capacitor of 20pF.

The output is short-circuit protected. RXD is switched off in Unpowered Mode (i.e., $V_S = 0V$).

3.10 Enable Input Pin (EN)

The Enable Input pin controls the operation mode of the device. If EN is high, the circuit is in Normal Mode, with transmission paths from TXD to LIN and from LIN to RXD both active. The VCC voltage regulator operates with 3.3V/5V/85mA output capability.

If EN is switched to low while TXD is still high, the device is forced to Silent Mode. No data transmission is then possible, and the current consumption is reduced to I_{VS} typ. 57µA. The VCC regulator has its full functionality.

If EN is switched to low while TXD is low, the device is forced to Sleep Mode. No data transmission is possible, and the voltage regulator is switched off.

3.11 Wake Input Pin (WAKE)

The Wake Input pin is a high-voltage input used to wake up the device from Sleep Mode or Silent Mode. It is usually connected to an external switch in the application to generate a local wake-up. A pull-up current source, typically 10µA, is implemented. If a local wake-up is not needed in the application, connect the Wake pin directly to the VS pin.

3.12 Mode Input Pin (MODE)

Connect the MODE pin directly or via an external resistor to GND for normal watchdog operation. To debug the software of the connected microcontroller, connect MODE pin to 3.3V/5V and the watchdog is switched off.

3.13 TM Input Pin

The TM pin is used for final production measurements at Atmel[®]. In normal application, it has to be always connected to GND.

3.14 KL_15 Pin

The KL_15 pin is a high-voltage input used to wake up the device from Sleep or Silent Mode. It is an edge sensitive pin (low-to-high transition). It is usually connected to ignition to generate a local wake-up in the application when the ignition is switched on. Although KL_15 pin is at high voltage (V_{Batt}), it is possible to switch the IC into Sleep or Silent Mode. Connect the KL_15 pin directly to GND if you do not need it. A debounce timer with a typical Tdb_{Kl_15} of 160µs is implemented.

The input voltage threshold can be adjusted by varying the external resistor due to the input current I_{KL_15} . To protect this pin against voltage transients, a serial resistor of $47k\Omega$ and a ceramic capacitor of 100nF are recommended. With this RC combination you can increase the wake-up time Tw_{KL_15} and, therefore, the sensitivity against transients on the ignition Kl.15.

You can also increase the wake-up time using external capacitors with higher values.

3.15 INH Output Pin

The INH Output pin is used to switch an external voltage regulator on during Normal or Fail-safe Mode. The INH pin is switched off in Sleep or Silent Mode. It is possible to switch off the external $1k\Omega$ master resistor via the INH pin for master node applications. The INH pin is switched off during VCC undervoltage reset.

3.16 Reset Output Pin (NRES)

The Reset Output pin, an open drain output, switches to low during V_{CC} undervoltage or a watchdog failure.



3.17 WD_OSC Output Pin

The WD_OSC Output pin provides a typical voltage of 1.2V, which supplies an external resistor with values between $34k\Omega$ and $120k\Omega$ to adjust the watchdog oscillator time.

3.18 NTRIG Input Pin

The NTRIG Input pin is the trigger input for the window watchdog. A pull-up resistor is implemented. A negative edge triggers the watchdog. The trigger signal (low) must exceed a minimum time $t_{trigmin}$ to generate a watchdog trigger.

3.19 Wake-up Events from Sleep or Silent Mode

- LIN-bus
- WAKE pin
- EN pin
- KL_15



4. Modes of Operation

Figure 4-1. Modes of Operation

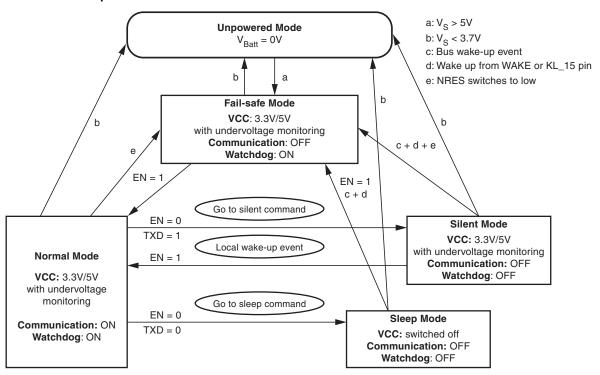


Table 4-1. Table of Modes

Mode of Operation	Transceiver	vcc	Watchdog	WD_OSC	INH	RXD	LIN
Fail-safe	Off	3.3V/5V	On	1.23V	On	High, except after wake-up	Recessive
Normal	On	3.3V/5V	On	1.23V	On	LIN depending	TXD depending
Silent	Off	3.3V/5V	Off	0V	Off	High	Recessive
Sleep	Off	0V	Off	0V	Off	0V	Recessive

4.1 Normal Mode

This is the normal transmitting and receiving mode of the LIN Interface in accordance with the LIN specification LIN 2.x. The voltage regulator is active and can source up to 85mA. The undervoltage detection is activated. The watchdog needs a trigger signal from NTRIG to avoid resets at NRES. If NRES is switched to low, the IC changes its state to Fail-safe Mode.



4.2 Silent Mode

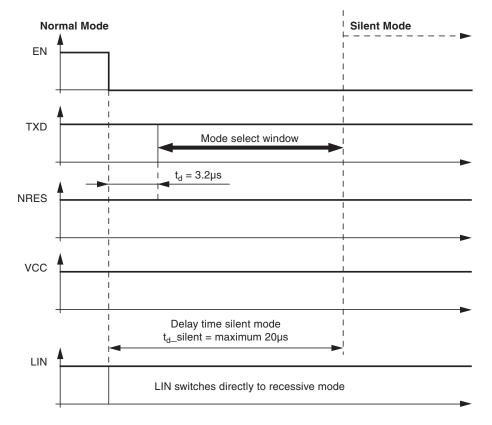
A falling edge at EN when TXD is high switches the IC into Silent Mode. The TXD Signal has to be logic high during the Mode Select window (see Figure 4-2). The transmission path is disabled in Silent Mode. The overall supply current from V_{Batt} is a combination of the I_{VSsi} = 57 μ A plus the VCC regulator output current I_{VCC} .

The internal slave termination between the LIN pin and the VS pin is disabled in Silent Mode, only a weak pull-up current (typically $10\mu A$) between the LIN pin and the VS pin is present. Silent Mode can be activated independently from the actual level on the LIN, WAKE, or KL_15 pins.

If an undervoltage condition occurs, NRES is switched to low, and the IC changes its state to Fail-safe Mode.

A voltage less than the LIN Pre_Wake detection VLINL at the LIN pin activates the internal LIN receiver and switches on the internal slave termination between the LIN pin and the V_S pin.

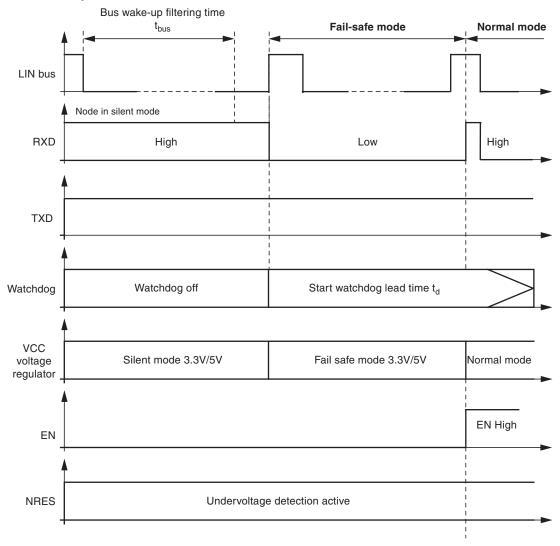
Figure 4-2. Switch to Silent Mode





A falling edge at the LIN pin followed by a dominant bus level maintained for a certain time period (> t_{bus}) and the following rising edge at the LIN pin (see Figure 4-3 on page 9) results in a remote wake-up request. The device switches from Silent Mode to Fail-safe Mode. The remote wake-up request is indicated by a low level at the RXD pin to interrupt the microcontroller (see Figure 4-3 on page 9). EN high can be used to switch directly to Normal Mode.

Figure 4-3. LIN Wake Up from Silent Mode





4.3 Sleep Mode

A falling edge at EN when TXD is low switches the IC into Sleep Mode. The TXD Signal has to be logic low during the Mode Select window (Figure 4-4 on page 10). In order to avoid any influence to the LIN-pin during switching into sleep mode it is possible to switch the EN up to 3.2 μ s earlier to LOW than the TXD. Therefore, the best and easiest way are two falling edges at TXD and EN at the same time. The transmission path is disabled in Sleep Mode. The supply current $I_{VSsleep}$ from V_{Batt} is typically 10μ A.

The VCC regulator is switched off. NRES and RXD are low. The internal slave termination between the LIN pin and VS pin is disabled, only a weak pull-up current (typically 10μ A) between the LIN pin and the VS pin is present. Sleep Mode can be activated independently from the current level on the LIN, WAKE, or KL_15 pin.

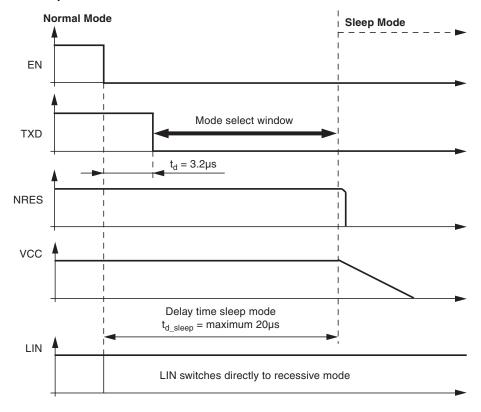
A voltage less than the LIN Pre_Wake detection VLINL at the LIN pin activates the internal LIN receiver and switches on the internal slave termination between the LIN pin and the V_S pin.

A falling edge at the LIN pin followed by a dominant bus level maintained for a certain time period (> t_{bus}) and a following rising edge at pin LIN results in a remote wake-up request. The device switches from Sleep Mode to Fail-safe Mode.

The VCC regulator is activated, and the remote wake-up request is indicated by a low level at the RXD pin to interrupt the microcontroller (see Figure 4-5 on page 11).

EN high can be used to switch directly from Sleep/Silent to Fail-safe Mode. If EN is still high after VCC ramp up and undervoltage reset time, the IC switches to the Normal Mode.

Figure 4-4. Switch to Sleep Mode





4.4 Fail-safe Mode

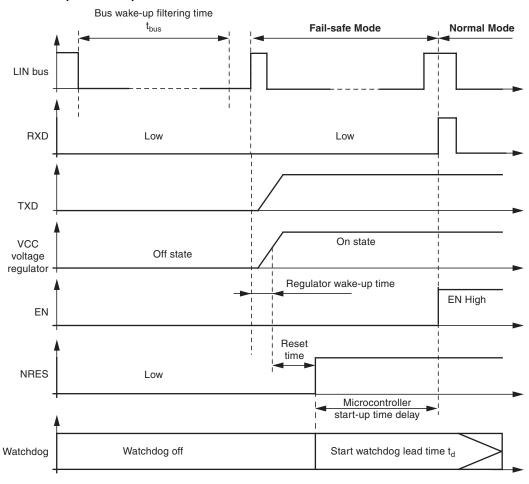
The device automatically switches to Fail-safe Mode at system power-up. The voltage regulator is switched on (see Figure 5-1 on page 13). The NRES output switches to low for t_{res} = 4ms and gives a reset to the microcontroller. LIN communication is switched off. The IC stays in this mode until EN is switched to high. The IC then changes to Normal Mode. A power down of V_{Batt} (V_{S} < 3.7V) during Silent or Sleep Mode switches the IC into Fail-safe Mode after power up. A low at NRES switches into Fail-safe Mode directly. During Fail-safe Mode the TXD pin is an output and signals the last wake-up source.

4.5 Unpowered Mode

If you connect battery voltage to the application circuit, the voltage at the VS pin increases according to the block capacitor (see Figure 5-1 on page 13). After VS is higher than the VS undervoltage threshold VS_{th} , the IC mode changes from Unpowered Mode to Fail-safe Mode. The VCC output voltage reaches its nominal value after t_{VCC} . This time, t_{VCC} , depends on the VCC capacitor and the load.

The NRES is low for the reset time delay t_{reset}. During this time, t_{reset}, no mode change is possible.

Figure 4-5. LIN Wake Up from Sleep Mode





5. Wake-up Scenarios from Silent or Sleep Mode

5.1 Remote Wake-up via Dominant Bus State

A voltage less than the LIN Pre_Wake detection V_{LINL} at the LIN pin activates the internal LIN receiver.

A falling edge at the LIN pin followed by a dominant bus level V_{BUSdom} maintained for a certain time period (> t_{BUS}) and a rising edge at pin LIN result in a remote wake-up request. The device switches from Silent or Sleep Mode to Fail-safe Mode. The VCC voltage regulator is/remains activated, the INH pin is switched to high, and the remote wake-up request is indicated by a low level at the RXD pin to generate an interrupt for the microcontroller. A low level at the LIN pin in the Normal Mode starts the bus wake-up filtering time, and if the IC is switched to Silent or Sleep Mode, it will receive a wake-up after a positive edge at the LIN pin.

5.2 Local Wake-up via Pin WAKE

A falling edge at the WAKE pin followed by a low level maintained for a certain time period ($> t_{WAKE}$) results in a local wake-up request. The device switches to Fail-safe Mode. The local wake-up request is indicated by a low level at the RXD pin to generate an interrupt in the microcontroller and a strong pull down at TXD. When the Wake pin is low, it is possible to switch to Silent or Sleep Mode via pin EN. In this case, the wake-up signal has to be switched to high $> 10 \mu$ s before the negative edge at WAKE starts a new local wake-up request.

5.3 Local Wake-up via Pin KL_15

A positive edge at pin KL_15 followed by a high voltage level for a certain time period (> t_{KL_15}) results in a local wake-up request. The device switches into the Fail-safe Mode. The extra long wake-up time ensures that no transients at KL_15 create a wake up. The local wake-up request is indicated by a low level at the RXD pin to generate an interrupt for the microcontroller and a strong pull down at TXD. During high-level voltage at pin KL_15, it is possible to switch to Silent or Sleep Mode via pin EN. In this case, the wake-up signal has to be switched to low > 250µs before the positive edge at KL_15 starts a new local wake-up request. With external RC combination, the time is even longer.

5.4 Wake-up Source Recognition

The device can distinguish between a local wake-up request (Wake or KL_15 pins) and a remote wake-up request (dominant LIN bus state). The wake-up source can be read on the TXD pin in Fail-safe Mode. A high level indicates a remote wake-up request (weak pull up at the TXD pin); a low level indicates a local wake-up request (strong pull down at the TXD pin). The wake-up request flag (signalled on the RXD pin), as well as the wake-up source flag (signalled on the TXD pin), is immediately reset if the microcontroller sets the EN pin to high (see Figure 4-2 on page 8 and Figure 4-3 on page 9) and the IC is in Normal Mode. The last wake-up source flag is stored and signalled in Fail-safe Mode at the TXD pin.

5.5 Fail-safe Features

- During a short-circuit at LIN to V_{Battery}, the output limits the output current to I_{BUS_lim}. Due to the power dissipation, the chip temperature exceeds T_{LINoff}, and the LIN output is switched off. The chip cools down and after a hysteresis of T_{hys}, switches the output on again. RXD stays on high because LIN is high. During LIN overtemperature switch-off, the VCC regulator works independently.
- During a short-circuit from LIN to GND the IC can be switched into Sleep or Silent Mode. If the short-circuit disappears, the IC starts with a remote wake-up.
- The reverse current is very low < 2μA at the LIN pin during loss of V_{Batt}. This is optimal behavior for bus systems where some slave nodes are supplied from battery or ignition.
- During a short circuit at VCC, the output limits the output current to I_{VCClim}. Because of undervoltage, NRES switches to low and sends a reset to the microcontroller. The IC switches into Fail-safe Mode. If the chip temperature exceeds the value T_{VCCoff}, the VCC output switches off. The chip cools down and after a hysteresis of T_{hys}, switches the output on again. Because of the Fail-safe Mode, the VCC voltage will switch on again although EN is switched off from the microcontroller. The microcontroller can start with its normal operation.
- EN pin provides a pull-down resistor to force the transceiver into recessive mode if EN is disconnected.
- RXD pin is set floating if V_{Batt} is disconnected.
- TXD pin provides a pull-up resistor to force the transceiver into recessive mode if TXD is disconnected.



- If TXD is short-circuited to GND, it is possible to switch to Sleep Mode via ENABLE after t_{dom} > 20ms (only for Atmel[®] ATA6622/ATA6624).
- If the WD_OSC pin has a short-circuit to GND and the NTRIG Signal has a period time > 27ms, the watchdog runs with an internal oscillator and guarantees a reset after the second NTRIG signal at the latest.
- If the resistor at WO_OSC pin is disconnected, the watchdog runs with an internal oscillator and guarantees a reseet
 after the second NTRIG signal at the latest.

5.6 Voltage Regulator

The voltage regulator needs an external capacitor for compensation and for smoothing the disturbances from the microcontroller. It is recommended to use an electrolythic capacitor with $C > 1.8\mu F$ and a ceramic capacitor with C = 100nF. The values of these capacitors can be varied by the customer, depending on the application.

The main power dissipation of the IC is created from the VCC output current I_{VCC} , which is needed for the application. In Figure 5-2 on page 13 the safe operating area of the Atmel ATA6624/ATA6626 is shown.

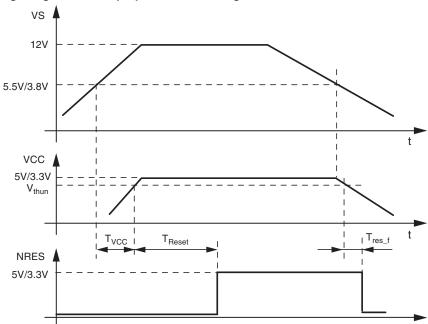
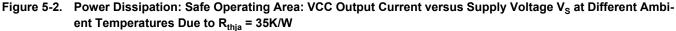
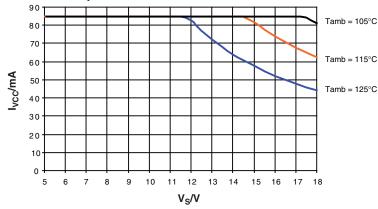


Figure 5-1. VCC Voltage Regulator: Ramp-up and Undervoltage Detection





For programming purposes of the microcontroller it is potentially necessary to supply the V_{CC} output via an external power supply while the V_S Pin of the system basis chip is disconnected. This will not affect the system basis chip.



6. Watchdog

The watchdog anticipates a trigger signal from the microcontroller at the NTRIG (negative edge) input within a time window of T_{wd} . The trigger signal must exceed a minimum time $t_{trigmin}$ > 200ns. If a triggering signal is not received, a reset signal will be generated at output NRES. The timing basis of the watchdog is provided by the internal oscillator. Its time period, T_{osc} , is adjustable via the external resistor R_{wd} osc $(34k\Omega to 120k\Omega)$.

During Silent or Sleep Mode the watchdog is switched off to reduce current consumption.

The minimum time for the first watchdog pulse is required after the undervoltage reset at NRES disappears. It is defined as lead time t_d . After wake up from Sleep or Silent Mode, the lead time t_d starts with the negative edge of the RXD output.

6.1 Typical Timing Sequence with $R_{WD OSC} = 51k\Omega$

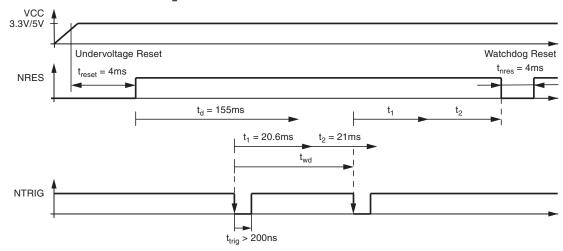
The trigger signal T_{wd} is adjustable between 20ms and 64ms using the external resistor R_{WD OSC}.

For example, with an external resistor of $R_{WD\ OSC}$ = 51k Ω ±1%, the typical parameters of the watchdog are as follows:

$$\begin{array}{l} t_{\rm osc} = 0.405 \times \ R_{\rm WD_OSC} - 0.0004 \times (R_{\rm WD_OSC})^2 \ (R_{\rm WD_OSC} \ \text{in } \ k\Omega; \ t_{\rm osc} \ \text{in } \mu \text{s}) \\ t_{\rm OSC} = 19.6 \mu \text{s} \ \text{due to } 51 \text{k}\Omega \\ t_{\rm d} = 7895 \times \ 19.6 \mu \text{s} = 155 \text{ms} \\ t_{1} = 1053 \times \ 19.6 \mu \text{s} = 20.6 \text{ms} \\ t_{2} = 1105 \times \ 19.6 \mu \text{s} = 21.6 \text{ms} \\ t_{\text{nres}} = \text{constant} = 4 \text{ms} \end{array}$$

After ramping up the battery voltage, the 3.3V/5V regulator is switched on. The reset output NRES stays low for the time t_{reset} (typically 4ms), then it switches to high, and the watchdog waits for the trigger sequence from the microcontroller. The lead time, t_d , follows the reset and is t_d = 155ms. In this time, the first watchdog pulse from the microcontroller is required. If the trigger pulse NTRIG occurs during this time, the time t_1 starts immediately. If no trigger signal occurs during the time t_d , a watchdog reset with t_{NRES} = 4ms will reset the microcontroller after t_d = 155ms. The times t_1 and t_2 have a fixed relationship between each other. A triggering signal from the microcontroller is anticipated within the time frame of t_2 = 21.6ms. To avoid false triggering from glitches, the trigger pulse must be longer than $t_{TRIG,min}$ > 200ns. This slope serves to restart the watchdog sequence. If the triggering signal fails in this open window t_2 , the NRES output will be drawn to ground. A triggering signal during the closed window t_1 immediately switches NRES to low.







Worst Case Calculation with R $_{\rm WD_OSC}$ = 51k Ω 6.2

The internal oscillator has a tolerance of 20%. This means that t₁ and t₂ can also vary by 20%. The worst case calculation for the watchdog period t_{wd} is calculated as follows.

The ideal watchdog time t_{wd} is between the maximum t_1 and the minimum t_1 plus the minimum t_2 .

$$\begin{array}{l} t_{1,\text{min}} = 0.8 \times t_1 = 16.5 \text{ms}, \, t_{1,\text{max}} = 1.2 \times t_1 = 24.8 \text{ms} \\ t_{2,\text{min}} = 0.8 \times t_2 = 17.3 \text{ms}, \, t_{2,\text{max}} = 1.2 \times t_2 = 26 \text{ms} \end{array}$$

$$t_{wdmax} = t_{1min} + t_{2min} = 16.5ms + 17.3ms = 33.8ms$$

 $t_{wdmin} = t_{1max} = 24.8ms$

 $t_{wd} = 29.3 \text{ms} \pm 4.5 \text{ms} (\pm 15\%)$

A microcontroller with an oscillator tolerance of ±15% is sufficient to supply the trigger inputs correctly.

Table 6-1. **Typical Watchdog Timings**

R _{WD_OSC} kΩ	Oscillator Period t _{osc} /µs	Lead Time t _d /ms	Closed Window t ₁ /ms	Open Window t ₂ /ms	Trigger Period from Microcontroller t _{wd} /ms	Reset Time t _{nres} /ms
34	13.3	105	14.0	14.7	19.9	4
51	19.61	154.8	20.64	21.67	29.32	4
91	33.54	264.80	35.32	37.06	50.14	4
120	42.84	338.22	45.11	47.34	64.05	4



7. Absolute Maximum Ratings

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Parameters	Symbol	Min.	Тур.	Max.	Unit
Supply voltage V _S	V _S	-0.3		+40	V
Pulse time \leq 500ms; T _a = 25°C Output current I _{VCC} \leq 85mA	V_S			+40	V
Pulse time ≤ 2min; T _a = 25°C Output current I _{VCC} ≤ 85mA	V_S			27	V
NAKE (with 33kΩ serial resistor) KL_15 (with 47kΩ/100nF) DC voltage Transient voltage due to ISO7637 (coupling 1nF)		-1 -150		+40 +100	V V
NH - DC voltage		-0.3		V _S + 0.3	V
LIN - DC voltage		-27		+40	V
Logic pins (RxD, TxD, EN, NRES, NTRIG, WD_OSC, MODE, TM)		-0.3		+5.5	V
Output current NRES	I _{NRES}			+2	mA
PVCC DC voltage VCC DC voltage		-0.3 -0.3		+5.5 +6.5	V V
ESD according to IBEE LIN EMC Test Spec. 1.0 following IEC 61000-4-2 - Pin VS, LIN, KL_15 (47kΩ/100nF) to GND - Pin WAKE (33kΩ serial resistor) to GND		±6 ±5			KV KV
ESD HBM following STM5.1 with 1.5kΩ100pF - Pin VS, LIN, KL_15, WAKE to GND		±6			KV
HBM ESD ANSI/ESD-STM5.1 JESD22-A114 AEC-Q100 (002)		±3			KV
CDM ESD STM 5.3.1		±750			V
Machine Model ESD AEC-Q100-RevF(003)		±200			V
Junction temperature	T_j	-40		+150	°C
Storage temperature	T_s	– 55		+150	°C

8. Thermal Characteristics

Parameters	Symbol	Min.	Тур.	Max.	Unit
Thermal resistance junction to heat slug	R_{thjc}			10	K/W
Thermal resistance junction to ambient, where heat slug is soldered to PCB	R _{thja}		35		K/W
Thermal shutdown of VCC regulator		150	165	170	°C
Thermal shutdown of LIN output		150	165	170	°C
Thermal shutdown hysteresis			10		°C



9. Electrical Characteristics

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Тур.	Max.	Unit	Type*
1	VS Pin								
1.1	Nominal DC voltage range		VS	V_S	5		27	V	Α
1.2	Supply current in Sleep	Sleep Mode $V_{LIN} > V_S - 0.5V$ $V_S < 14V (T_j = 25^{\circ}C)$	VS	I _{VSsleep}	3	10	14	μΑ	В
1.2	Mode	Sleep Mode $V_{LIN} > V_S - 0.5V$ $V_S < 14V (T_j = 125^{\circ}C)$	VS	l _{VSsleep}	5	11	16	μА	Α
1.3	Supply current in Silent	Bus recessive $V_S < 14V (T_j = 25^{\circ}C)$ Without load at VCC	VS	$I_{ m VSsi}$	47	57	67	μΑ	В
1.0	Mode	Bus recessive $V_S < 14V (T_j = 125^{\circ}C)$ Without load at VCC	VS	I _{VSsi}	56	66	76	μΑ	А
1.4	Supply current in Normal Mode	Bus recessive $V_S < 14V$ Without load at VCC	VS	I _{VSrec}	0.3		0.8	mA	Α
1.5	Supply current in Normal Mode	Bus dominant $V_S < 14V$ V_{CC} load current 50 mA	VS	I _{VSdom}	50		53	mA	A
1.6	Supply current in Fail- safe Mode	Bus recessive $V_S < 14V$ Without load at VCC	VS	I _{VSfail}	250		550	μΑ	Α
1.7	V _S undervoltage threshold		VS	V _{Sth}	3.7	4.4	5	V	Α
1.8	VS undervoltage threshold hysteresis		VS	V_{Sth_hys}		0.2		V	Α
2	RXD Output Pin								
2.1	Low-level output sink current	Normal Mode $V_{LIN} = 0V$ $V_{RXD} = 0.4V$	RXD	I _{RXD}	1.3	2.5	8	mA	Α
2.2	Low-level output voltage	$I_{RXD} = 1mA$	RXD	V_{RXDL}			0.4	V	Α
2.3	Internal resistor to V_{CC}		RXD	R_{RXD}	3	5	7	$k\Omega$	Α
3	TXD Input/Output Pin								
3.1	Low-level voltage input		TXD	V_{TXDL}	-0.3		+0.8	V	Α
3.2	High-level voltage input		TXD	V_{TXDH}	2		V _{CC} + 0.3V	٧	Α
3.3	Pull-up resistor	$V_{TXD} = 0V$	TXD	R_{TXD}	125	250	400	$k\Omega$	Α
3.4	High-level leakage current	V _{TXD} = VCC	TXD	I _{TXD}	-3		+3	μΑ	Α
3.5	Low-level output sink current at local wake-up request	Fail-safe Mode $V_{LIN} = V_{S}$ $V_{WAKE} = 0V$ $V_{TXD} = 0.4V$	TXD	I _{TXDwake}	2	2.5	8	mA	Α
4	EN Input Pin								
*\ T	1000/ tooted I	B = 100% correlation tested (Chair			Danisus			

^{*)} Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter



No.	Parameters	Test Conditions	Pin	Symbol	Min.	Тур.	Max.	Unit	Type*
4.1	Low-level voltage input		EN	V_{ENL}	-0.3		+0.8	V	Α
4.2	High-level voltage input		EN	V_{ENH}	2		V _{CC} + 0.3V	V	Α
4.3	Pull-down resistor	$V_{EN} = V_{CC}$	EN	R_{EN}	50	125	200	kΩ	Α
4.4	Low-level input current	$V_{EN} = 0V$	EN	I _{EN}	-3		+3	μΑ	Α
5	NTRIG Watchdog Input F	Pin							_
5.1	Low-level voltage input		NTRIG	V_{NTRIGL}	-0.3		+0.8	V	Α
5.2	High-level voltage input		NTRIG	V_{NTRIGH}	2		V _{CC} + 0.3V	V	Α
5.3	Pull-up resistor	$V_{NTRIG} = 0V$	NTRIG	R_{NTRIG}	125	250	400	kΩ	Α
5.4	High-level leakage current	$V_{NTRIG} = V_{CC}$	NTRIG	I _{NTRIG}	-3		+3	μΑ	Α
6	Mode Input Pin								
6.1	Low-level voltage input		MODE	V_{MODEL}	-0.3		+0.8	V	Α
6.2	High-level voltage input		MODE	V_{MODEH}	2		V _{CC} + 0.3V	V	Α
6.3	Leakage current	$V_{MODE} = V_{CC}$ or $V_{MODE} = 0V$	MODE	I _{MODE}	-3		+3	μΑ	Α
7	INH Output Pin								
7.1	High-level voltage	I _{INH} = -15mA	INH	V_{INHH}	$V_{S} - 0.75$		V_S	V	Α
7.2	Switch-on resistance between VS and INH		INH	R _{INH}		30	50	Ω	Α
7.3	Leakage current	Sleep Mode $V_{INH} = 0V/27V, V_S = 27V$	INH	I _{INHL}	-3		+3	μΑ	Α
8	Load 3 (Medium): 6.8nF,	d Conditions: ; Load 2 (Large): 10nF, 500Ω 660Ω Characterized on Sam he Timing Parameters for Pro	nples				9 at 10.4kE	Bit/s	
8.1	Driver recessive output voltage	Load1/Load2	LIN	V_{BUSrec}	$0.9 \times V_S$		V_S	V	Α
8.2	Driver dominant voltage	$V_{VS} = 7V$ $R_{load} = 500\Omega$	LIN	V_LoSUP			1.2	V	Α
8.3	Driver dominant voltage	$V_{VS} = 18V$ $R_{load} = 500\Omega$	LIN	$V_{ ext{HiSUP}}$			2	V	Α
8.4	Driver dominant voltage	$V_{VS} = 7.0V$ $R_{load} = 1000\Omega$	LIN	V_LoSUP_1k	0.6			V	Α
8.5	Driver dominant voltage	$V_{VS} = 18V$ $R_{load} = 1000\Omega$	LIN	V_HiSUP_1k	0.8			V	Α
8.6	Pull-up resistor to V _S	The serial diode is mandatory	LIN	R_{LIN}	20	30	60	kΩ	Α
8.7	Voltage drop at the serial diodes	In pull-up path with R_{slave} $I_{SerDiode}$ = 10mA	LIN	$V_{SerDiode}$	0.4		1.0	V	D
8.8	LIN current limitation V _{BUS} = V _{Batt_max}		LIN	I _{BUS_LIM}	40	120	200	mA	Α

^{*)} Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter



 $5 V < V_S < 27 V,$ -40°C < Tj < 150°C, unless otherwise specified. All values refer to GND pins

Parameters	Test Conditions	Pin	Symbol	Min.	Тур.	Max.	Unit	Type*
Input leakage current at the receiver including pull-up resistor as specified	Input leakage current Driver off V _{BUS} = 0V V _{Batt} = 12V	LIN	I _{BUS_PAS_dom}	– 1	-0.35		mA	Α
Leakage current LIN recessive	Driver off $8V < V_{Batt} < 18V$ $8V < V_{BUS} < 18V$ $V_{BUS} \ge V_{Batt}$	LIN	I _{BUS_PAS_rec}		10	20	μΑ	Α
Leakage current when control unit disconnected from ground. Loss of local ground must not affect communication in the residual network.	$GND_{Device} = V_S$ $V_{Batt} = 12V$ $0V < V_{BUS} < 18V$	LIN	I _{BUS_NO_gnd}	-10	+0.5	+10	μА	Α
Leakage current at a disconnected battery. Node has to sustain the current that can flow under this condition. Bus must remain operational under this condition.	V _{Batt} disconnected V _{SUP_Device} = GND 0V < V _{BUS} < 18V	LIN	I _{BUS_NO_bat}		0.1	2	μА	Α
Capacitance on pin LIN to GND		LIN	C_{LIN}			20	pF	D
LIN Bus Receiver								
Center of receiver threshold	$V_{BUS_CNT} = (V_{th_dom} + V_{th_rec})/2$	LIN	V _{BUS_CNT}	$\begin{array}{c} 0.475 \times \\ V_S \end{array}$	$^{0.5 imes}_{ extsf{S}}$	$\begin{array}{c} 0.525 \times \\ \mathrm{V_S} \end{array}$	V	Α
Receiver dominant state	V _{EN} = 5V	LIN	$V_{\rm BUSdom}$			$0.4 \times V_S$	V	Α
Receiver recessive state	V _{EN} = 5V	LIN	V_{BUSrec}	$0.6\timesV_{S}$			V	Α
Receiver input hysteresis	$V_{hys} = V_{th_rec} - V_{th_dom}$	LIN	V_{BUShys}	$\begin{array}{c} 0.028 \times \\ \text{V}_{\text{S}} \end{array}$	$0.1 \times V_S$	0.175 × V _S	V	Α
Pre_Wake detection LIN High-level input voltage		LIN	V_{LINH}	$V_S - 2V$		V _S + 0.3V	V	Α
Pre Wake detection LIN						V _S -		
Low-level input voltage	Activates the LIN receiver	LIN	V_{LINL}	–27		3.3V	V	Α
Low-level input voltage Internal Timers		LIN	V _{LINL}	<i>–</i> 27			V	А
Low-level input voltage		LIN	V _{LINL}	-27 30	90		ν μs	A
Low-level input voltage Internal Timers Dominant time for wake-					90	3.3V		
Low-level input voltage Internal Timers Dominant time for wake- up via LIN bus Time delay for mode change from Fail-safe into Normal Mode via EN	V _{LIN} = 0V	LIN	t _{bus}	30		3.3V 150	µs	Α
	Input leakage current at the receiver including pull-up resistor as specified Leakage current LIN recessive Leakage current when control unit disconnected from ground. Loss of local ground must not affect communication in the residual network. Leakage current at a disconnected battery. Node has to sustain the current that can flow under this condition. Bus must remain operational under this condition. Capacitance on pin LIN to GND LIN Bus Receiver Center of receiver threshold Receiver dominant state Receiver recessive state Receiver input hysteresis Pre_Wake detection LIN High-level input voltage	Input leakage current at the receiver including pull-up resistor as specified Leakage current LIN recessive Leakage current between control unit disconnected from ground. Loss of local ground must not affect communication in the residual network. Leakage current at a disconnected battery. Node has to sustain the current that can flow under this condition. Capacitance on pin LIN to GND LIN Bus Receiver Center of receiver threshold Lin bus leakage current at a disconnected only leave the current that can flow under this condition. Capacitance on pin LIN to GND LIN Bus Receiver Center of receiver threshold Capacitance on pin LIN to GND LIN Bus Receiver Center of receiver threshold Receiver dominant state V _{EN} = 5V Receiver input hysteresis Pre_Wake detection LIN High-level input voltage	Parameters Test Conditions Pin Input leakage current at the receiver including pull-up resistor as specified Input leakage current Driver off VBUS = 0V VBALT = 12V LIN Leakage current LIN recessive Driver off SV < VBALT < 18V VBUS < 18V	Input leakage current at the receiver including pull-up resistor as specified $V_{BUS} = 0V$ $V_{Batt} = 12V$ Leakage current LIN recessive $V_{But} < 18V$	Input leakage current at the receiver including pull-up resistor as specified $V_{Bus} = 0V \\ V_{Batt} = 12V $ $V_{Batt} = 12V $ $V_{Bus} = 0V \\ V_{Bus} = 12V $ $V_{Bus} = 1$	Input leakage current at the receiver including pull-up resistor as specified $\begin{array}{c} \text{Driver off} \\ V_{BUS} = 0V \\ V_{Batt} = 12V \\ \end{array}$ Driver off $\begin{array}{c} V_{BUS} = 0V \\ V_{Batt} = 12V \\ \end{array}$ Driver off $\begin{array}{c} V_{BUS} = 0V \\ V_{Batt} = 12V \\ \end{array}$ Driver off $\begin{array}{c} V_{BUS} = 0V \\ V_{Batt} = 12V \\ \end{array}$ Driver off $\begin{array}{c} V_{BUS} = 0V \\ V_{Batt} = 12V \\ \end{array}$ Driver off $\begin{array}{c} V_{BUS} = 0V \\ V_{BUS} = 18V \\ V_{BUS} = 18V \\ V_{BUS} = 18V \\ \end{array}$ LIN $\begin{array}{c} I_{BUS_PAS_dom} & -1 & -0.35 \\ \end{array}$ 10 $\begin{array}{c} I_{BUS_PAS_dom} & -1 & -0.35 \\ \end{array}$ Leakage current when control unit disconnected from ground. Loss of local ground must not affect communication in the residual network. Leakage current at a disconnected battery. Node has to sustain the current that can flow under this condition. Bus must remain operational under this condition. Capacitance on pin LIN to GND LIN Bus Receiver Center of receiver $\begin{array}{c} V_{BUS_CNT} = V_{V_{th_dom}} + V_{th_rec} / 2 \\ V_{th_dom} + V_{th_rec} / 2 \\ V_{th} = 5V \\ Receiver dominant state \\ Receiver recessive state \\ Receiver input hysteresis \\ Pre_Wake detection LIN High-level input voltage \\ Pre_Wake detection LIN High-level input voltage \\ Pre_Wake detection LIN High-level input voltage \\ $	Input leakage current at the receiver including pull-up resistor as value of the receiver including pull-up resistor as value of the value of the receiver including pull-up resistor as value of the v	Input leakage current at the receiver including pull-up resistor as specified $V_{SUS} = 0V$ $V_{Batt} = 12V$ $V_{Batt} = 12V$ $V_{Bus} = 12V$

^{*)} Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter



Parameters	Test Conditions	Pin	Symbol	Min.	Тур.	Max.	Unit	Type*
Time delay for mode change from Silent Mode into Normal Mode via EN	V _{EN} = 5V	EN	t _{s_n}	5	15	40	μs	Α
Duty cycle 1	$\begin{aligned} & \text{TH}_{\text{Rec(max)}} = 0.744 \times \text{V}_{\text{S}} \\ & \text{TH}_{\text{Dom(max)}} = 0.581 \times \text{V}_{\text{S}} \\ & \text{V}_{\text{S}} = 7.0 \text{V to 18V} \\ & t_{\text{Bit}} = 50 \mu \text{s} \\ & \text{D1} = t_{\text{bus_rec(min)}} / (2 \times t_{\text{Bit}}) \end{aligned}$	LIN	D1	0.396				Α
Duty cycle 2	$\begin{aligned} & \text{TH}_{\text{Rec}(\text{min})} = 0.422 \times \text{V}_{\text{S}} \\ & \text{TH}_{\text{Dom}(\text{min})} = 0.284 \times \text{V}_{\text{S}} \\ & \text{V}_{\text{S}} = 7.6 \text{V to } 18 \text{V} \\ & t_{\text{Bit}} = 50 \mu \text{s} \\ & \text{D2} = t_{\text{bus_rec}(\text{max})} / (2 \times t_{\text{Bit}}) \end{aligned}$	LIN	D2			0.581		А
Duty cycle 3	$\begin{aligned} & TH_{Rec(max)} = 0.778 \times V_S \\ & TH_{Dom(max)} = 0.616 \times V_S \\ & V_S = 7.0V \text{ to } 18V \\ & t_{Bit} = 96\mu s \\ & D3 = t_{bus_rec(min)}/(2 \times t_{Bit}) \end{aligned}$	LIN	D3	0.417				Α
Duty cycle 4	$\begin{aligned} & \text{TH}_{\text{Rec(min)}} = 0.389 \times \text{V}_{\text{S}} \\ & \text{TH}_{\text{Dom(min)}} = 0.251 \times \text{V}_{\text{S}} \\ & \text{V}_{\text{S}} = 7.6 \text{V to } 18 \text{V} \\ & t_{\text{Bit}} = 96 \mu \text{s} \\ & \text{D4} = t_{\text{bus_rec(max)}} / (2 \times t_{\text{Bit}}) \end{aligned}$	LIN	D4			0.590		Α
Slope time falling and rising edge at LIN	V _S = 7.0V to 18V	LIN	$t_{\text{SLOPE_fall}} \\ t_{\text{SLOPE_rise}}$	3.5		22.5	μs	Α
		l Layer						
Propagation delay of receiver (Figure 9-1 on page 23)	V_S = 7.0V to 18V t_{rx_pd} = max(t_{rx_pdr} , t_{rx_pdf})	RXD	t _{rx_pd}			6	μs	Α
Symmetry of receiver propagation delay rising edge minus falling edge	$V_S = 7.0V \text{ to } 18V$ $t_{rx_sym} = t_{rx_pdr} - t_{rx_pdf}$	RXD	t _{rx_sym}	-2		+2	μs	A
NRES Open Drain Output	t Pin							
Low-level output voltage	$V_S \ge 5.5V$ $I_{NRES} = 1mA$	NRES	V _{NRESL}			0.14	V	Α
Low-level output low	10k Ω to 5V V _{CC} = 0V	NRES	V_{NRESLL}			0.14	V	Α
Undervoltage reset time	$V_S \ge 5.5V$ $C_{NRES} = 20pF$	NRES	t _{reset}	2	4	6	ms	Α
Reset debounce time for falling edge	$V_S \ge 5.5V$ $C_{NRES} = 20pF$	NRES	t _{res_f}	1.5		10	μs	Α
Watchdog Oscillator								
Voltage at WD_OSC in Normal Mode	$I_{WD_OSC} = -200\mu A$ $V_{VS} \ge 4V$	WD_ OSC	V_{WD_OSC}	1.13	1.23	1.33	V	Α
Possible values of resistor		WD_ OSC	R _{osc}	34		120	kΩ	Α
	Time delay for mode change from Silent Mode into Normal Mode via EN Duty cycle 1 Duty cycle 2 Duty cycle 3 Duty cycle 4 Slope time falling and rising edge at LIN Receiver Electrical AC Palin Receiver (Figure 9-1 on page 23) Symmetry of receiver propagation delay rising edge minus falling edge NRES Open Drain Output Low-level output voltage Low-level output voltage Low-level output low Undervoltage reset time Reset debounce time for falling edge Watchdog Oscillator Voltage at WD_OSC in Normal Mode	Time delay for mode change from Silent Mode into Normal Mode via EN	Time delay for mode change from Silent Mode into Normal Mode via EN $V_{EN} = 5V \qquad EN$ $Duty cycle 1 \qquad TH_{Rec(max)} = 0.744 \times V_S \\ TH_{Dom(max)} = 0.581 \times V_S \\ V_S = 7.0V \text{ to } 18V \\ t_{Bit} = 50 \mu S \\ D1 = t_{bus_rec(min)}/(2 \times t_{Bit})$ $Duty cycle 2 \qquad TH_{Rec(min)} = 0.422 \times V_S \\ TH_{Dom(min)} = 0.284 \times V_S \\ V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 50 \mu S \\ D2 = t_{bus_rec(max)}/(2 \times t_{Bit})$ $Duty cycle 3 \qquad TH_{Rec(max)} = 0.778 \times V_S \\ TH_{Dom(min)} = 0.284 \times V_S \\ V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96 \mu S \\ D3 = t_{bus_rec(min)}/(2 \times t_{Bit})$ $Duty cycle 4 \qquad TH_{Rec(min)} = 0.389 \times V_S \\ TH_{Dom(min)} = 0.251 \times V_S \\ V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96 \mu S \\ D4 = t_{bus_rec(max)}/(2 \times t_{Bit})$ $Slope time falling and rising edge at LIN \\ Receiver Electrical AC Parameters of the LIN Physical Layer LIN Receiver, RXD Load Conditions: C_{RXD} = 20 pF Propagation delay of receiver (Figure 9-1 on page 23) Symmetry of receiver propagation delay rising edge minus falling edge V_S = 7.0V \text{ to } 18V \\ t_{rx_pd} = max(t_{rx_pdr}, t_{rx_pdr}) RXD V_S = 7.0V \text{ to } 18V \\ t_{rx_pd} = max(t_{rx_pdr}, t_{rx_pdr}) RXD V_S = 7.0V \text{ to } 18V \\ t_{rx_pd} = max(t_{rx_pdr}, t_{rx_pdr}) RXD V_S = 7.0V \text{ to } 18V \\ t_{rx_pd} = max(t_{rx_pdr}, t_{rx_pdr}) RXD V_S = 7.0V \text{ to } 18V \\ t_{rx_pd} = max(t_{rx_pdr}, t_{rx_pdr}) RXD V_S = 7.0V \text{ to } 18V \\ t_{rx_pd} = max(t_{rx_pdr}, t_{rx_pdr}) RXD V_S = 7.0V \text{ to } 18V \\ t_{rx_pd} = max(t_{rx_pdr}, t_{rx_pdr}) RXD V_{rx_sym} = t_{rx_pdr} - t_{rx_pdr} NRES NRES NRES Duty cycle 2 Delay Del$	Time delay for mode change from Silent Mode into Normal Mode via EN $V_{EN} = 5V \qquad EN \qquad t_{s_n}$ $TH_{Rec(max)} = 0.744 \times V_{S} \\ TH_{Dom(max)} = 0.581 \times V_{S} \\ V_{S} = 7.0V \text{ to } 18V \\ t_{Bit} = 50\mu_{S} \\ D1 = t_{bus_rec(min)}/(2 \times t_{Bit})$ $TH_{Rec(min)} = 0.422 \times V_{S} \\ TH_{Dom(min)} = 0.284 \times V_{S} \\ V_{S} = 7.6V \text{ to } 18V \\ t_{Bit} = 50\mu_{S} \\ D2 = t_{bus_rec(max)}/(2 \times t_{Bit})$ $TH_{Rec(max)} = 0.778 \times V_{S} \\ TH_{Dom(max)} = 0.616 \times V_{S} \\ V_{S} = 7.0V \text{ to } 18V \\ t_{Bit} = 96\mu_{S} \\ D3 = t_{bus_rec(min)}/(2 \times t_{Bit})$ $TH_{Rec(min)} = 0.389 \times V_{S} \\ TH_{Dom(min)} = 0.251 \times V_{S} \\ V_{S} = 7.0V \text{ to } 18V \\ t_{Bit} = 96\mu_{S} \\ D3 = t_{bus_rec(min)}/(2 \times t_{Bit})$ $TH_{Rec(min)} = 0.389 \times V_{S} \\ TH_{Dom(min)} = 0.251 \times V_{S} \\ V_{S} = 7.0V \text{ to } 18V \\ V_{S} = 7.0V to$	Time delay for mode change from Silent Mode into Normal Mode via EN $V_{EN} = 5V \qquad EN \qquad t_{g,n} \qquad 5$ $TH_{Rec(max)} = 0.744 \times V_{S} \\ TH_{Dom(max)} = 0.781 \times V_{S} \\ V_{S} = 7.0V \text{ to } 18V \\ t_{g,n} = 50 \text{ us} $ $Duty cycle 1 \qquad V_{S} = 7.6V \text{ to } 18V \\ t_{g,n} = 50 \text{ us} $ $Duty cycle 2 \qquad TH_{Rec(max)} = 0.422 \times V_{S} \\ TH_{Dom(max)} = 0.284 \times V_{S} \\ V_{S} = 7.6V \text{ to } 18V \\ t_{g,n} = 50 \text{ us} $ $D1 = t_{bus_{s}, rec(max)}/(2 \times t_{g,n})$ $TH_{Rec(max)} = 0.778 \times V_{S} \\ TH_{Dom(max)} = 0.278 \times V_{S} \\ TH_{Dom(max)} = 0.389 \times V_{S} \\ TH_{Dom(max)} = 0.389 \times V_{S} \\ TH_{Dom(max)} = 0.389 \times V_{S} \\ TH_{Dom(max)} = 0.251 \times V_{S} \\ V_{S} = 7.6V \text{ to } 18V \\ t_{g,n} = 96 \text{ us} $ $D4 = t_{bus_{s}, rec(max)}/(2 \times t_{g,n})$ $D4 = t_{bus_{s},$	Time delay for mode change from Silent Mode into Normal Mode via EN $TH_{Rec(max)} = 0.744 \times V_S \\ TH_{Dom(max)} = 0.581 \times V_S \\ V_S = 7.0V \text{ to } 18V \\ t_{int} = 50\mu S \\ D1 \text{ to } 18V \\ t_{int} = 50\mu S \\ D2 = t_{bus.rec(min)}/(2 \times t_{Bit})$ $D1 \text{ to } 190 \text{ to } 190 \text{ to } 190 \\ V_S = 7.6V \text{ to } 18V \\ t_{int} = 50\mu S \\ D2 = t_{bus.rec(min)}/(2 \times t_{Bit})$ $D2 \text{ to } 190 \text{ to } 190 \\ V_S = 7.6V \text{ to } 18V \\ t_{int} = 50\mu S \\ D2 = t_{bus.rec(min)}/(2 \times t_{Bit})$ $D2 \text{ to } 190 \text{ to } 190 \\ V_S = 7.6V \text{ to } 18V \\ t_{int} = 96\mu S \\ D3 = t_{bus.rec(min)}/(2 \times t_{Bit})$ $D1 \text{ to } 190 \text{ to } 190 \\ V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ D4 = t_{bus.rec(min)}/(2 \times t_{Bit})$ $D2 \text{ to } 190 \text{ to } 190 \text{ to } 190 \\ V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ D4 = t_{bus.rec(min)} = 0.289 \times V_S \\ V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ D4 = t_{bus.rec(min)}/(2 \times t_{Bit})$ $V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ D4 = t_{bus.rec(min)}/(2 \times t_{Bit})$ $V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ D3 = t_{bus.rec(min)}/(2 \times t_{Bit})$ $V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ D4 = t_{bus.rec(min)}/(2 \times t_{Bit})$ $V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ D3 = t_{bus.rec(min)}/(2 \times t_{Bit})$ $V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ D4 = t_{Bit}/(2 \times t_{Bit})$ $V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ D4 = t_{Bit}/(2 \times t_{Bit})$ $V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ D4 = t_{Bit}/(2 \times t_{Bit})$ $V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ D4 = t_{Bit}/(2 \times t_{Bit})$ $V_S = 7.6V \text{ to } 18V \\ t_{Bit} = 96\mu S \\ D4 = t_{Bit}/(2 \times t_{Bit})$ $V_S = 7.6V \text{ to } 18V \\ t_{Bit}/(2 \times t_{Bit})$ $V_S = 7.$	Time delay for mode change from Silent Mode into Normal Mode via EN	Time delay for mode change from Silent Mode into Normal Mode via EN THRec(max) = 0.744 × V _S THD _{Desc(max)} = 0.581 × V _S V _S = 7.0 V to 18V t _{st} = 50µS D1 = t _{bus.pec(min/(2 × t_{st}))} LIN D1 0.396 Duty cycle 1 THRec(max) = 0.422 × V_S THD_{Desc(max)} = 0.284 × V_S V_S = 7.0 V to 18V t_{st} = 50µS D2 = t_{bus.pec(min/(2 × t_{st}))} LIN D2 THRec(max) = 0.778 × V_S THD_{Desc(max)} = 0.284 × V_S V_S = 7.0 V to 18V t_{st} = 50µS D3 = t_{bus.pec(min/(2 × t_{st}))} LIN D3 0.417 THRec(max) = 0.78 × V_S THD_{Desc(max)} = 0.816 × V_S V_S = 7.0 V to 18V t_{st} = 96µS D3 = t_{bus.pec(min/(2 × t_{st}))} LIN D4 0.590 THRec(max) = 0.70 × V_S THD_{Desc(max)} = 0.251 × V_S V_S THD_{Desc(max)} = 0.251 × V_S V_S = 7.0 V to 18V t_{st} = 96µS D4 = t_{bus.pec(min/(2 × t_{st}))} LIN D4 0.590 Slope time falling and rising edge at LIN Receiver Electrical AC Parameters of the LIN Physical Layer LIN Receiver Electrical AC Parameters of the LIN Physical Layer LIN Receiver Electrical AC Parameters of the LIN Physical Layer LIN Receiver Richips 9-1 on page 23) Symmetry of receiver propagation delay of t_{x,y,p} = max(t_{x,y,p}, t_{x,p,pl}) RXD t_{x,p} = max(t_{x,y,p}, t_{x,p,pl}) RXD t_{x,p} = 0.24 ± 2 µS Symmetry of receiver propagation delay rising edge minus falling edge two propagation delay rising edge minus falling edge v_{x,y,pl} = t_{x,p,pl} - t_{x,p,pl} - t_{x,p,pl} - t_{x,pl} = max(t_{x,po} - t_{x,p,pl}) NRES Open Drain Output Pin Low-level output voltage V_S ≥ 5.5V NRES V_{NRES} 1.5 0.14 V Undervoltage reset time V_{x,pl} × 5.5 V V_{NRES} 2.20pF NRES 1.5 1.5 10 µS Reset debounce time for V_x ≥ 5.5V V_{NRES} 2.20pF NRES 1.5 1.5 10 µS Watchdog Oscillator Voltage at WD_OSC in Normal Mode VD_OSC 5.1 1.13 1.23 1.33 V}}}}}

^{*)} Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter



Parameters	Test Conditions	Pin	Symbol	Min.	Тур.	Max.	Unit	Type*
Oscillator period	$R_{OSC} = 34k\Omega$		t _{osc}	10.65	13.3	15.97	μs	Α
Oscillator period	$R_{OSC} = 51k\Omega$		tosc	15.68	19.6	23.52	μs	Α
Oscillator period	$R_{OSC} = 91k\Omega$		t_{OSC}	26.83	33.5	40.24	μs	Α
Oscillator period	R_{OSC} = 120k Ω		t _{osc}	34.2	42.8	51.4	μs	Α
Watchdog lead time after Reset			t_{d}		7895		cycles	Α
Watchdog closed window			t ₁		1053		cycles	Α
Watchdog open window			t ₂		1105		cycles	Α
Watchdog reset time NRES		NRES	t _{nres}	3.2	4	4.8	ms	Α
KL_15 Pin								
High-level input voltage $R_V = 47k\Omega$	Positive edge initializes a wake-up	KL_15	V_{KL_15H}	4		V _S + 0.3V	V	Α
Low-level input voltage $R_V = 47k\Omega$		KL_15	V _{KL_15L}	-1		+2	V	Α
KL_15 pull-down current	V _S < 27V V _{KL_15} = 27V	KL_15	I _{KL_15}		50	65	μΑ	Α
Internal debounce time	Without external capacitor	KL_15	Tdb _{KL_15}	80	160	250	μs	Α
KL_15 wake-up time	$R_V = 47k\Omega$, $C = 100nF$	KL_15	Tw_{KL_15}	0.4	2	4.5	ms	С
WAKE Pin								
High-level input voltage		WAKE	V_{WAKEH}	$V_S - 1V$		V _S + 0.3V	V	Α
Low-level input voltage	Initializes a wake-up signal	WAKE	\ /	– 1		V _S -		Α
		**/ " _	V _{WAKEL}			3.3V	V	
WAKE pull-up current	V _S < 27V V _{WAKE} = 0V	WAKE	V _{WAKEL}	-30	-10	3.3V	V μA	Α
WAKE pull-up current High-level leakage current	V _S < 27V				-10	3.3V +5		A
High-level leakage	$V_S < 27V$ $V_{WAKE} = 0V$ $V_S = 27V$	WAKE	I _{WAKE}	-30	-10 70		μA	
High-level leakage current Time of low pulse for	$V_S < 27V$ $V_{WAKE} = 0V$ $V_S = 27V$ $V_{WAKE} = 27V$ $V_{WAKE} = 0V$	WAKE	I _{WAKE}	-30 -5		+5	μΑ	Α
High-level leakage current Time of low pulse for wake-up via WAKE pin VCC Voltage Regulator A	$V_S < 27V$ $V_{WAKE} = 0V$ $V_S = 27V$ $V_{WAKE} = 27V$ $V_{WAKE} = 0V$	WAKE	I _{WAKE}	-30 -5		+5	μΑ	Α
High-level leakage current Time of low pulse for wake-up via WAKE pin	$V_S < 27V$ $V_{WAKE} = 0V$ $V_S = 27V$ $V_{WAKE} = 27V$ $V_{WAKE} = 0V$ ATA6622, PVCC = VCC $4V < V_S < 18V$	WAKE WAKE	I _{WAKE} I _{WAKEL} I _{WAKEL}	-30 -5 30		+5 150	μA μA μs	A A
High-level leakage current Time of low pulse for wake-up via WAKE pin VCC Voltage Regulator A	$V_{S} < 27V$ $V_{WAKE} = 0V$ $V_{S} = 27V$ $V_{WAKE} = 27V$ $V_{WAKE} = 0V$ ATA6622, PVCC = VCC $4V < V_{S} < 18V$ $(0mA to 50mA)$ $4.5V < V_{S} < 18V$	WAKE WAKE VCC	I _{WAKE} I _{WAKEL} I _{WAKEL} VCC _{nor}	-30 -5 30		+5 150 3.366	μΑ μΑ μs	A A
High-level leakage current Time of low pulse for wake-up via WAKE pin VCC Voltage Regulator A Output voltage VCC Output voltage VCC at	$V_{S} < 27V$ $V_{WAKE} = 0V$ $V_{S} = 27V$ $V_{WAKE} = 27V$ $V_{WAKE} = 0V$ ATA6622, PVCC = VCC $4V < V_{S} < 18V$ $(0mA to 50mA)$ $4.5V < V_{S} < 18V$ $(0mA to 85mA)$	WAKE WAKE WAKE VCC VCC	I _{WAKE} I _{WAKEL} I _{WAKEL} VCC _{nor} VCC _{nor}	-30 -5 30 3.234 3.234		+5 150 3.366 3.366	μA μA μs V	A A C
High-level leakage current Time of low pulse for wake-up via WAKE pin VCC Voltage Regulator A Output voltage VCC Output voltage VCC at low VS	$V_{S} < 27V$ $V_{WAKE} = 0V$ $V_{S} = 27V$ $V_{WAKE} = 27V$ $V_{WAKE} = 0V$ ATA6622, PVCC = VCC $4V < V_{S} < 18V$ $(0mA to 50mA)$ $4.5V < V_{S} < 18V$ $(0mA to 85mA)$ $3V < V_{S} < 4V$ $V_{S} > 3V$	WAKE WAKE WAKE VCC VCC VCC VCC VS,	I _{WAKEL} I _{WAKEL} VCC _{nor} VCC _{low}	-30 -5 30 3.234 3.234		+5 150 3.366 3.366 3.366	μA μA μs V V	A A C A
	Oscillator period Oscillator period Oscillator period Oscillator period Watchdog Timing Relativ Watchdog lead time after Reset Watchdog closed window Watchdog open window Watchdog reset time NRES KL_15 Pin High-level input voltage $R_V = 47k\Omega$ Low-level input voltage $R_V = 47k\Omega$ KL_15 pull-down current Internal debounce time KL_15 wake-up time WAKE Pin High-level input voltage	Oscillator period $R_{OSC} = 34k\Omega$ Oscillator period $R_{OSC} = 51k\Omega$ Oscillator period $R_{OSC} = 91k\Omega$ Oscillator period $R_{OSC} = 120k\Omega$ Watchdog Timing Relative to t_{OSC} Watchdog lead time after Reset Watchdog closed window Watchdog open window Watchdog reset time NRES KL_15 Pin High-level input voltage $R_V = 47k\Omega$ Positive edge initializes a wake-up Low-level input voltage $R_V = 47k\Omega$ KL_15 pull-down current $V_S < 27V$ $V_{KL_15} = 27V$ Internal debounce time Without external capacitor KL_15 wake-up time $R_V = 47k\Omega$ $C = 100nF$ WAKE Pin High-level input voltage	Oscillator period $R_{OSC} = 34k\Omega$ Oscillator period $R_{OSC} = 51k\Omega$ Oscillator period $R_{OSC} = 91k\Omega$ Oscillator period $R_{OSC} = 120k\Omega$ Watchdog Timing Relative to t_{OSC} Watchdog lead time after Reset Watchdog closed window Watchdog open window Watchdog reset time NRES KL_15 Pin High-level input voltage $R_V = 47k\Omega$ Low-level input voltage $R_V = 47k\Omega$ KL_15 pull-down current $V_S < 27V$ $V_{KL_15} = 27V$ Internal debounce time Without external capacitor KL_15 WAKE Pin High-level input voltage $R_V = 47k\Omega$ C = 100nF KL_15 WAKE Pin	$\begin{array}{cccccccccccccccccccccccccccccccccccc$	$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	Oscillator period $R_{OSC} = 34k\Omega$ t_{OSC}

^{*)} Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter



No.	Parameters	Test Conditions	Pin	Symbol	Min.	Тур.	Max.	Unit	Type*
17.6	Load regulation	5mA < I _{VCC} < 50mA	VCC	VCC_{load}		0.1	0.5	%	Α
17.7	Power supply ripple rejection	10Hz to 100kHz $C_{VCC} = 10\mu F$ $V_S = 14V, I_{VCC} = -15mA$	VCC		50			dB	D
17.8	Output current limitation	V _S > 4V	VCC	I _{VCClim}	-240	-160	-85	mA	Α
17.9	External load capacity	0.2Ω < ESR < 5Ω at 100kHz for phase margin ≥ 60° ESR < 0.2Ω at 100kHz for phase margin ≥ 30°	VCC	C_load	1.8	10		μF	D
17.10	VCC undervoltage threshold	Referred to VCC V _S > 4V	VCC	V_{thunN}	2.8		3.2	٧	А
17.11	Hysteresis of undervoltage threshold	Referred to VCC V _S > 4V	VCC	Vhys _{thun}		150		mV	Α
17.12	Ramp-up time $V_S > 4V$ to $V_{CC} = 3.3V$	C_{VCC} = 2.2 μ F I_{load} = -5mA at VCC	VCC	T _{VCC}		100	250	μs	Α
18	VCC Voltage Regulator A	TA6624/ATA6626, PVCC = V	CC						
18.1	Output voltage VCC	5.5V < V _S < 18V (0mA to 50mA)	VCC	VCC _{nor}	4.9		5.1	V	Α
10.1	Output voltage voo	6V < V _S < 18V (0mA to 85mA)	VCC	VCC _{nor}	4.9		5.1	V	С
18.2	Output voltage VCC at low VS	4V < V _S < 5.5V	VCC	VCC _{low}	$V_S - V_D$		5.1	٧	Α
18.3	Regulator drop voltage	$V_S > 4V$ $I_{VCC} = -20$ mA	VS, VCC	V_{D1}			250	mV	Α
18.4	Regulator drop voltage	$V_S > 4V$ $I_{VCC} = -50$ mA	VS, VCC	V_{D2}		400	600	mV	Α
18.5	Regulator drop voltage	$V_S > 3.3V$ $I_{VCC} = -15mA$	VS, VCC	V_{D3}			200	mV	Α
18.6	Line regulation	5.5V < V _S < 18V	VCC	VCC _{line}		0.1	0.2	%	Α
18.7	Load regulation	5mA < I _{VCC} < 50mA	VCC	VCC_{load}		0.1	0.5	%	Α
18.8	Power supply ripple rejection	10Hz to 100kHz $C_{VCC} = 10\mu F$ $V_S = 14V, I_{VCC} = -15mA$	VCC		50			dB	D
18.9	Output current limitation	V _S > 5.5V	VCC	I _{VCClim}	-240	-130	-85	mA	Α
18.10	External load capacity	0.2Ω < ESR < 5Ω at 100kHz for phase margin ≥ 60° ESR < 0.2Ω at 100kHz for phase margin ≥ 30°	VCC	C_{load}	1.8	10		μF	D
18.11	VCC undervoltage threshold	Referred to VCC V _S > 5.5V	VCC	V_{thunN}	4.2		4.8	V	Α
18.12	Hysteresis of undervoltage threshold	Referred to VCC V _S > 5.5V	VCC	Vhys _{thun}		250		mV	Α
18.13	Ramp-up time $V_S > 5.5V$ to $V_{CC} = 5V$	C_{VCC} = 2.2 μ F I_{load} = -5mA at VCC	VCC	t _{VCC}		130	300	μs	Α

^{*)} Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter



Figure 9-1. Definition of Bus Timing Characteristics

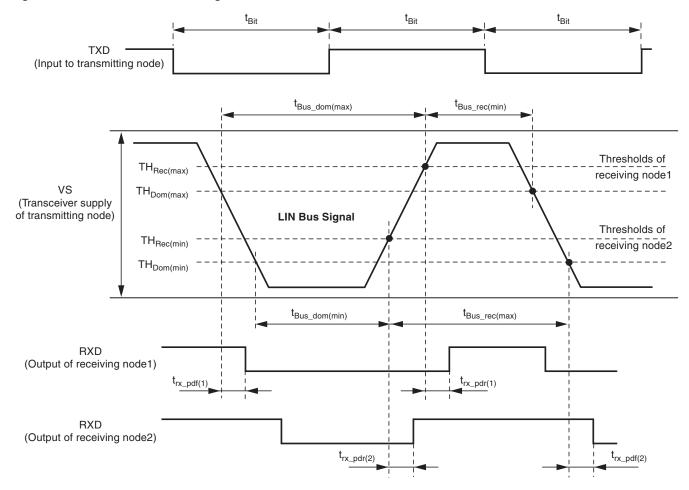
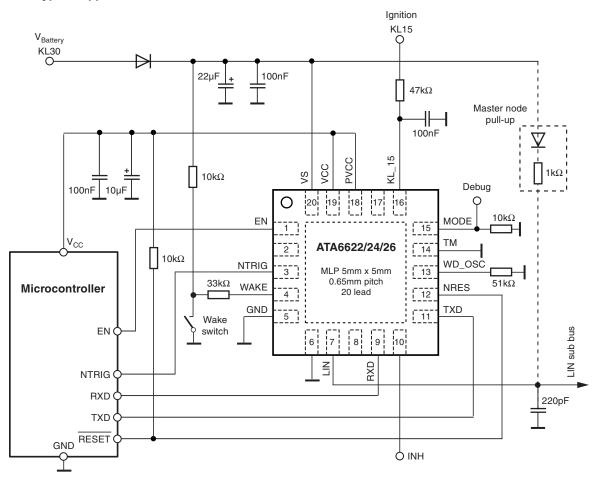




Figure 9-2. Typical Application Circuit





Ignition KL15 V_{Battery} KL30 0 100nF MJD31C 47kΩ Master node pull-up 2.2µF 100nF 3.3Ω KL_15 VCC S 10kΩ Debug 10µF 17 16 20 19 18 MODE $10k\Omega$ ΕN 15 V_{CC} TM ATA6622/24/26 14 $10 k \Omega$ WD_OSC **NTRIG** MLP 5mm x 5mm 0.65mm pitch 13 $33 \text{k}\Omega$ $51 k\Omega$ Microcontroller WAKE **NRES** 12 20 lead **GND** []] TXD Wake LIN sub bus ΕN switch 9 8 10 RXD Z **NTRIG** RXD 220pF TXD GND RESET

Figure 9-3. Application Circuit with External NPN-Transistor

*) Note that the output voltage PVCC is no longer short-ciruit protected when boosting the output current by an external NPN-transistor.



Q INH

VBAT O-+ C2 22μF/50V VCC PVCC C5 СЗ C1 100nF 10μF 100nF 20 19 18 17 16 ΕN MODE GND TM 14 VCC ▼ NTRIG WD_OSC ATA6622/24/26 LIN Sub Bus VCC WAKE NRES Microcontroller GND TXD 5 ΕN GND RXD Ξ NTRIG RXD TXD RESET GND 10kΩ

Figure 9-4. LIN Slave Application with Minimum External Devices

Note: No watchdog, INH output not used, no local wake-up



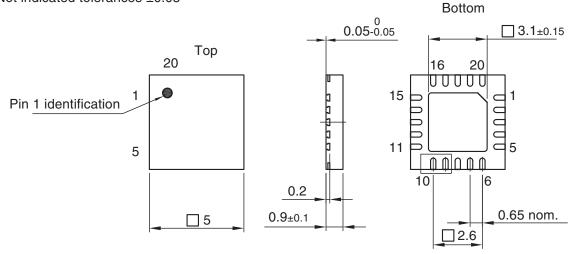
10. Ordering Information

Extended Type Number	Package	Remarks
ATA6622-PGPW	QFN20	3.3V LIN system-basis-chip, Pb-free, 1.5k, taped and reeled
ATA6624-PGPW	QFN20	5V LIN system-basis-chip, Pb-free, 1.5k, taped and reeled
ATA6622-PGQW	QFN20	3.3V LIN system-basis-chip, Pb-free, 6k, taped and reeled
ATA6624-PGQW	QFN20	5V LIN system-basis-chip, Pb-free, 6k, taped and reeled
ATA6626-PGQW	QFN20	5V LIN system-basis-chip, Pb-free, 6k, taped and reeled
ATA6622C-PGPW	QFN20	3.3V LIN system-basis-chip, Pb-free, 1.5k, taped and reeled
ATA6624C-PGPW	QFN20	5V LIN system-basis-chip, Pb-free, 1.5k, taped and reeled
ATA6622C-PGQW	QFN20	3.3V LIN system-basis-chip, Pb-free, 6k, taped and reeled
ATA6624C-PGQW	QFN20	5V LIN system-basis-chip, Pb-free, 6k, taped and reeled
ATA6626C-PGQW	QFN20	5V LIN system-basis-chip, Pb-free, 6k, taped and reeled

11. Package Information

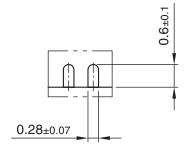
Package: VQFN_5 x 5_20L Exposed pad 3.1 x 3.1 Dimensions in mm

Not indicated tolerances ±0.05



Drawing-No.: 6.543-5129.01-4

Issue: 2; 09.02.07





technical drawings according to DIN specifications



12. Revision History

Please note that the following page numbers referred to in this section refer to the specific revision mentioned, not to this document.

document.	
Revision No.	History
4986K-AUTO-01/12	Table 2-1 "Pin Description" on page 3 changed
4986J-AUTO-03/11	Features on page 1 changed
	Section 1 "Description" on pages 1 to 2 changed
	Table 2-1 "Pin Description" on page 3 changed
	Section 3 "Functional Description" on pages 4 to 6 changed
	Section 4 "Modes of Operation" on pages 7 to 11 changed
	Section 5 "Wake-up Scenarios from Silent to Sleep Mode" on pages 12 to 14 changed
	Section 7 "Absolute Maximum Ratings" on page 17 changed
	Section 9 "Electrical Characteristics" on pages 18 to 26 changed
4986I-AUTO-07/10	 Section 6 "Watchdog" on pages 15 to 16 changed
4986H-AUTO-05/10	 New Part numbers ATA6622C, ATA6624C and ATA6626C added
	Features on page 1 changed
	Pin Description table: rows Pin 4 and Pin 15 changed
	• Text under headings 3.3, 3.9, 3.11, 5.5 and 6 changed
	• Figures 4-5, 6-1 and 9-3 changed
	Abs.Max.Rat.Table -> Values in row "ESD HBM following" changed
	• El.Char.Table -> rows changed: 7.1, 12.1, 12.2, 17.5, 17.6, 17.7, 17.8, 18.6, 18.7, 18.8, 18.9
	• El.Char.Table -> row 8.13 added
	• Figures 9-2 and 9-3 figure title changed
	Figure 9-4 on page 27 added
	Ord.Info.Table -> new part numbers added
4986G-AUTO-08/09	 complete datasheet: "LIN 2.0 specification" changed in "LIN 2.1 specification"
	• Figures changed: 1-1, 4-2, 4-3, 4-4, 4-5, 5-1, 9-2, 9-3
	• Sections changed: 3.1, 3.6, 3.8, 3.9, 3.10, 3.14, 4.1, 4.2, 4,3, 5.1, 5.2, 5.3, 5.5, 5.6
	Features and Description changed
	Table 4-1 changed
	Abs. Max. Ratings table changed
	Thermal Characteristics table inserted
	El. Characteristics table changed
4986F-AUTO-05/08	Section 3.15 "INH Output Pin" on page 6 changed
	Section 5.5 "Fail-safe Features" on page 13 changed
	• Section 6.1 "Typical Timing Sequence with R_{WD_OSC} = 51 k Ω " on page 15 changed
	Section 8 "Electrical Characteristics" numbers 1.6 to 1.8 on page 18 changed
4986E-AUTO-02/08	Figure 2-1 on page 3 renamed
	• Figure 6-1 "Timing Sequence with R_{WD_OSC} = 51 k Ω " on page 16 changed
	Figure 8-3 "Application Circuit with External NPN" on page 26 added
4986D-AUTO-10/07	Section 9 "Ordering Information" on page 26 changed





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