

<b>36-55 V</b> Continuous Input	<b>55 V</b> Transient Input	<b>9.6 V</b> Output	<b>240 W</b> Max Power	<b>2000 V dc</b> Isolation	<b>Quarter-brick</b> DC-DC Converter
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The BusQor® BQ55090QTA27 bus converter is a next-generation, board-mountable, isolated, fixed switching frequencydc/dc converter that uses synchronous rectification to achieve extremely high conversion efficiency. The power dissipated by the converter is so low that a heatsink is not required, which saves cost, weight, height, and application effort. The BusQor series provides an isolated step down voltage from 48V to 12 or 9.6V intermediate bus with no regulation in a standard "quarter-brick" module. BusQor converters are ideal for creating the mid-bus voltage required to drive point-of-load (non-isolated) converters in intermediate bus architectures.

### Operational Features

- High efficiency, 96 % at full rated load current at full load
- Delivers full power with minimal derating - no heatsink required
- Operating input voltage range: 36-55 V dc
- Fixed frequency switching provides predictable EMI
- No minimum load requirement

### Mechanical Features

- Industry standard quarter-brick pin-out configuration (two or four output pin option)
- Size: 1.45" x 2.3" (36.8x58.4mm)
- Total height: 0.445" (11.3mm)
- Total weight: 1.5 oz (42 g)

### Protection Features

- Input under-voltage lockout and over-voltage shutdown protects against abnormal input voltages
- Output current limit and short circuit protection (auto recovery)
- Output over-voltage protection
- Thermal shutdown

**BusQor®**  
Bus  
Converter



BQ55090QTA27  
Module

### Control Features

- On/Off control referenced to input side
- Inherent current share (by droop method) for high current and parallel applications.

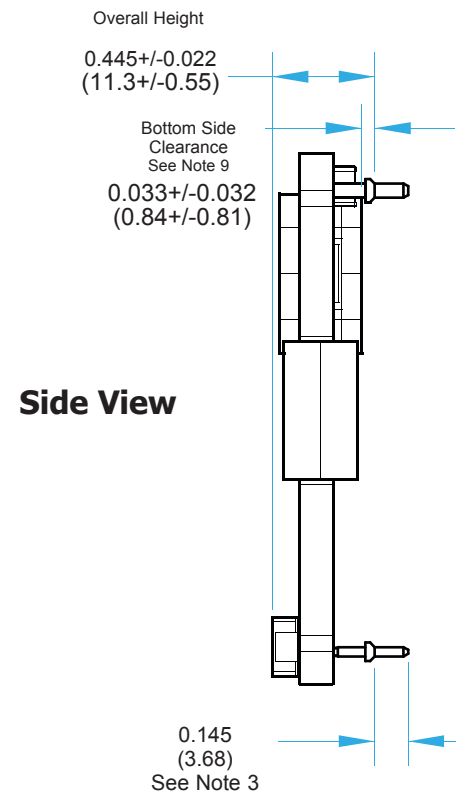
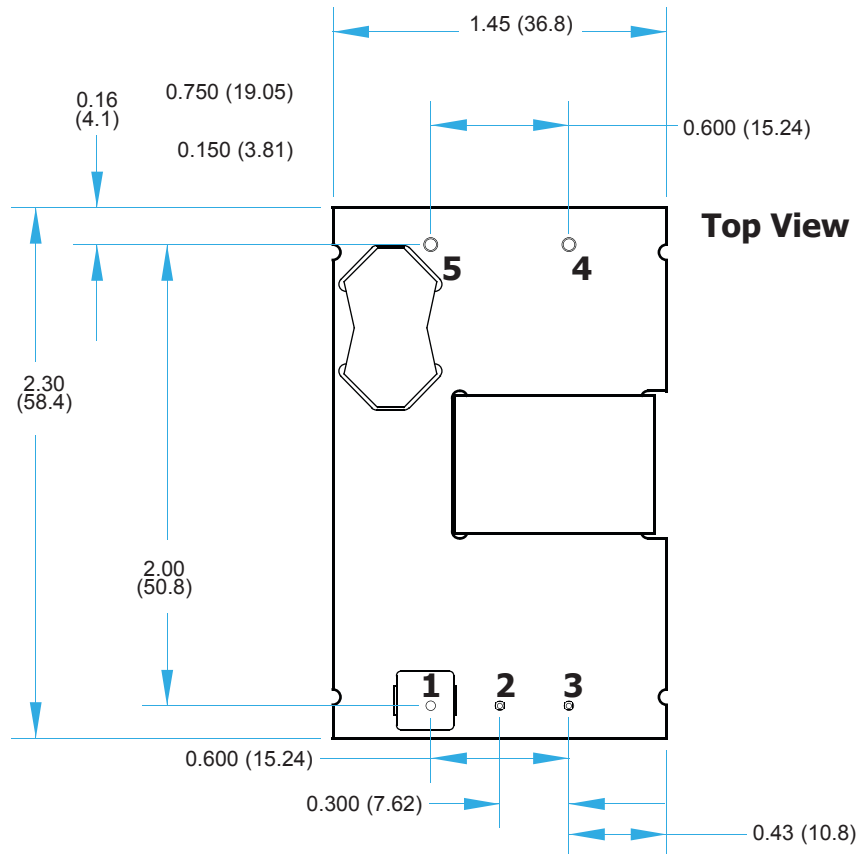
### Safety Features

- 2000 V input-to-output isolation
- UL 60950-1:2003
- CAN/CSA-C22.2 No. 60950-1:2003
- EN60950-1:2001
- Needle Flame Test (IEC 695-2-2)
- IEC 61000-4-2
- RoHS compliant (see page 14)

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### MECHANICAL DIAGRAM



### NOTES

- Pins 1-3 are 0.040" (1.02mm) diameter with 0.080" (2.03 mm) diameter standoff shoulders.
- Pins 4-5 are 0.062" (1.57 mm) diameter with 0.100" (2.54 mm) diameter standoff shoulders.
- Other pin lengths available. Recommended pin length is 0.03" (0.76mm) greater than the PCB thickness.
- All Pins: Material - Copper Alloy  
Finish: Tin over Nickel plate
- Undimensioned components are shown for visual reference only.
- All dimensions in inches (mm)  
Tolerances: x.xx +/-0.02 in. (x.x +/-0.5mm)  
x.xxx +/-0.010 in. (x.xx +/-0.25mm)
- Weight: 1.5 oz (42 g) typical
- Workmanship: Meets or exceeds IPC-A-610C Class II
- UL/TUV standards require a clearance of 0.04" (1.02mm) around primary areas of the module.
- The flanged pins are designed to permit surface mount soldering (allowing to avoid the wave soldering process) through the use of the flanged pin-in-paste technique.

### PIN DESIGNATIONS

Pin	Name	Function
1	Vin(+)	Positive input voltage
2	ON/OFF	Logic control input to turn converter on and off.
3	Vin(-)	Negative input
4	Vout(-)	Negative output
5	Vout(+)	Positive output



## Technical Specification

**Input:** 36-55 V  
**Output:** 9.6 V  
**Current:** 27A  
**Package:** Quarter-brick

### BQ55 FAMILY ELECTRICAL CHARACTERISTICS

Ta = 25 °C, airflow rate = 300 LFM, Vin = 48 V dc unless otherwise noted; full operating temperature range is -40 °C to +100 °C ambient temperature with appropriate power derating. Specifications subject to change without notice.

Parameter	Min.	Typ.	Max.	Units	Notes & Conditions
<b>ABSOLUTE MAXIMUM RATINGS</b>					
Input Voltage					
Non-Operating	0		60	V	Continuous
Operating	36	48	55	V	Continuous
Operating Transient Protection			N/A	V	100 ms transient, square wave
Isolation Voltage					
Input to Output			2000	V	
Operating Temperature	-40		100	°C	
Storage Temperature	-55		125	°C	
Voltage at ON/OFF input pin	-2		18	V	
<b>INPUT CHARACTERISTICS</b>					
Operating Input Voltage Range	36	48	55	V	
Input Under-Voltage Lockout					
Turn-On Voltage Threshold		34.3		V	
Turn-Off Voltage Threshold		32		V	
Lockout Voltage Hysteresis		2.4		V	
Input Over-Voltage Shutdown				V	
Turn-On Voltage Threshold		59.2		V	
Turn-Off Voltage Threshold		56.9		V	
Maximum Input Current			8.5	A	
No-Load Input Current			110	mA	
Disabled Input Current		15.0	30.0	mA	
Input Reflected-Ripple Current		10	20	mA	RMS through 10µH inductor
Input Terminal-Ripple Current		140		mA	RMS, full load
Recommended Input Fuse (see Note 1)			20	A	Fast blow external fuse recommended
Recommended External Input Capacitance		47		µF	Typical ESR 0.1-0.2 Ω
Input Filter Component Values (C/L/C)		1\6.6		µH\µF	Internal values; see Figure E
<b>OUTPUT CHARACTERISTICS</b>					
Output Voltage Set Point		9.6		V	
Output Voltage Regulation					
Over Line		40\3.8		%\V	
Over Load		5.2\500		%\mV	
Over Temperature		2\200		%\mV	
Total Output Voltage Range	6.550		11.000	V	Over sample, line, load, temperature & life
Output Voltage Ripple and Noise					20 MHz bandwidth; see Note 2
Peak-to-Peak		50	90	mV	Full load
RMS		20		mV	Full load
Operating Output Current Range	0		36	A	at 7.2 Vout, 36Vin; Subject to thermal derating
Output DC Current-Limit Inception		48		A	36Vin, Output Voltage 10% Low
Output DC Current-Limit Shutdown Voltage		5		V	
Back-Drive Current Limit while Disabled		10		mA	Negative current drawn from output
Maximum Output Capacitance			3,000	µF	9.6Vout at 27 A Resistive Load
<b>DYNAMIC CHARACTERISTICS</b>					
Output Voltage during Load Current Transient					
Step Change in Output Current (0.1 A/µs)		200		mV	50% to 75% to 50% Iout max
Settling Time		100		µs	To within 1% Vout nom
Turn-On Transient					
Turn-On Time (with 3mF output capacitance)		2	3	ms	Full load, Vout=90% nom.
Start-Up Inhibit Time		200		ms	-40 °C to +125 °C; Figure F
Output Voltage Overshoot		0		%	3 mF load capacitance



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Parameter	Min.	Typ.	Max.	Units	Notes & Conditions
<b>ISOLATION CHARACTERISTICS</b>					
Isolation Voltage (dielectric strength)		2000		V	See Absolute Maximum Ratings
Isolation Resistance		30		MΩ	
Isolation Capacitance (input to output)		470		pF	See Note 3
<b>TEMPERATURE LIMITS FOR POWER DERATING CURVES</b>					
Semiconductor Junction Temperature			125	°C	Package rated to 150 °C
Board Temperature			125	°C	UL rated max operating temp 130 °C
Transformer Temperature			125	°C	
Maximum Baseplate Temperature, Tb		N/A		°C	
<b>EFFICIENCY</b>					
100% Load		96		%	
50% Load		96		%	
<b>FEATURE CHARACTERISTICS</b>					
Switching Frequency	131	155	178	kHz	
ON/OFF Control (Option P)					
Off-State Voltage	-1		0.8	V	
On-State Voltage	2.4		18	V	
ON/OFF Control (Option N)					Application notes Figures A & B
Pull-Up Voltage	2.4		18	V	
Pull-Up Resistance	-1		0.8	V	
Output Over-Voltage Protection		7.5		V	Over full temp range; no load
Over-Temperature Shutdown OTP Trip Point	140		150	°C	Average PCB Temperature
Over-Temperature Shutdown Restart Hysteresis		10		°C	
<b>RELIABILITY CHARACTERISTICS</b>					
Calculated MTBF (Telcordia) TR-NWT-000332		1		10 <sup>6</sup> Hrs.	80% load, 200LFM, 40 °C Ta
Calculated MTBF (MIL-217) MIL-HDBK-217F		TBD		10 <sup>6</sup> Hrs.	80% load, 200LFM, 40 °C Ta
Field Demonstrated MTBF				10 <sup>6</sup> Hrs.	See our website for details

Note 1: UL's product certification tests were carried out using 20A fast blow fuse. Fuse interruption characteristics have to be taken into account while designing input traces. User should ensure that Input trace is capable of withstanding fault currents

Note 2: For applications requiring reduced output voltage ripple and noise, consult SynQor applications support (e-mail: support@synqor.com)

Note 3: Isolation capacitance can be added external to the module.



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### BQ55090QTA27 Standards and Qualifications

Parameter	Notes & Conditions
<b>STANDARDS COMPLIANCE</b>	
UL 60950-1:2003	
CAN/CSA-C22.2 No. 60950-1:2003	
EN60950-1:2001	
Needle Flame Test (IEC 695-2-2)	Test on entire assembly; board & plastic components UL94V-0 compliant
IEC 61000-4-2	ESD test, 8 kV - NP, 15 kV air - NP (Normal Performance)

Note: An external input fuse must always be used to meet these safety requirements. Contact SynQor for official safety certificates on new releases or download from the SynQor website.

Parameter	# Units	Test Conditions
<b>QUALIFICATION TESTING</b>		
Life Test	32	95% rated Vin and load, units at derating point, 1000 hours
Vibration	5	10-55 Hz sweep, 0.060" total excursion, 1 min./sweep, 120 sweeps for 3 axis
Mechanical Shock	5	100g minimum, 2 drops in x and y axis, 1 drop in z axis
Temperature Cycling	10	-40 °C to 100 °C, unit temp. ramp 15 °C/min., 500 cycles
Power/Thermal Cycling	5	Toperating = min to max, Vin = min to max, full load, 100 cycles
Design Marginality	5	Tmin-10 °C to Tmax+10 °C, 5 °C steps, Vin = min to max, 0-105% load
Humidity	5	85 °C, 85% RH, 1000 hours, continuous Vin applied except 5 min/day
Solderability	15 pins	MIL-STD-883, method 2003



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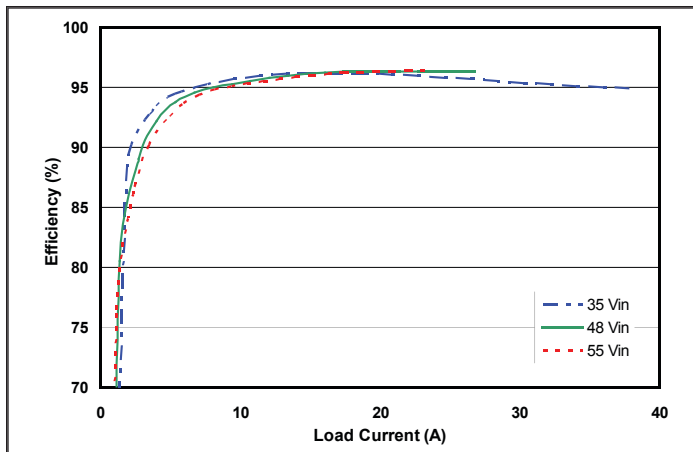


Figure 1: Efficiency at nominal output voltage vs. load current for minimum, nominal, and maximum input voltage at 25°C.

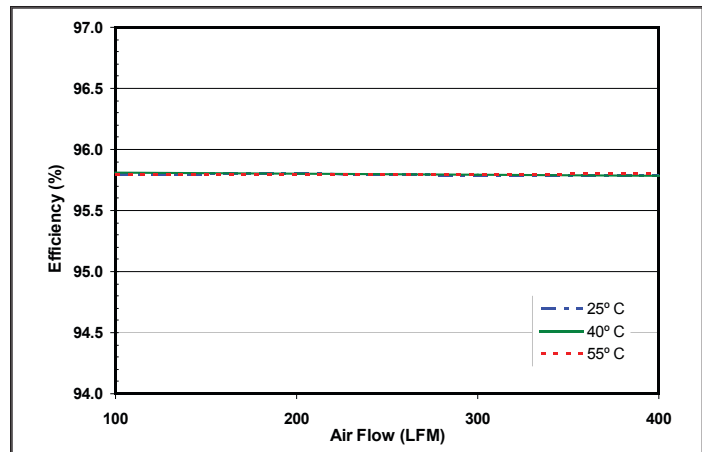


Figure 2: Efficiency at nominal output voltage and 60% rated power vs. airflow rate for ambient air temperatures of 25°C, 40°C, and 55°C (nominal input voltage).

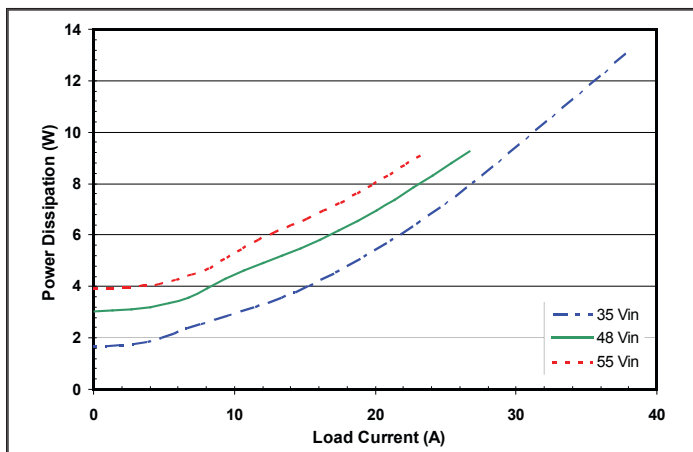


Figure 3: Power dissipation at nominal output voltage vs. load current for minimum, nominal, and maximum input voltage at 25°C.

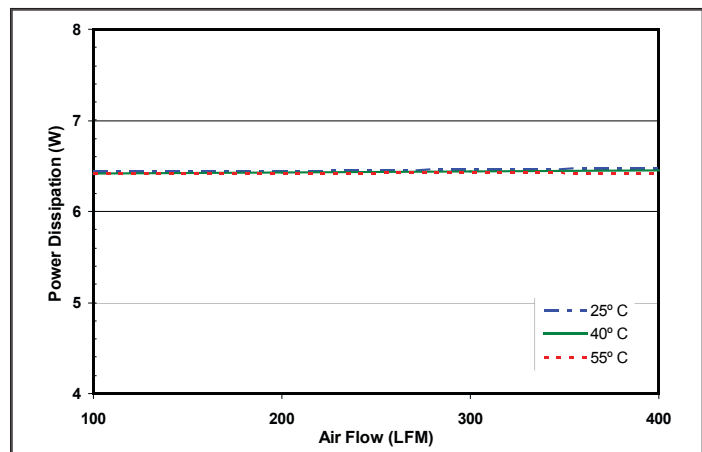


Figure 4: Power dissipation at nominal output voltage and 60% rated power vs. airflow rate for ambient air temperatures of 25°C, 40°C, and 55°C (nominal input voltage).

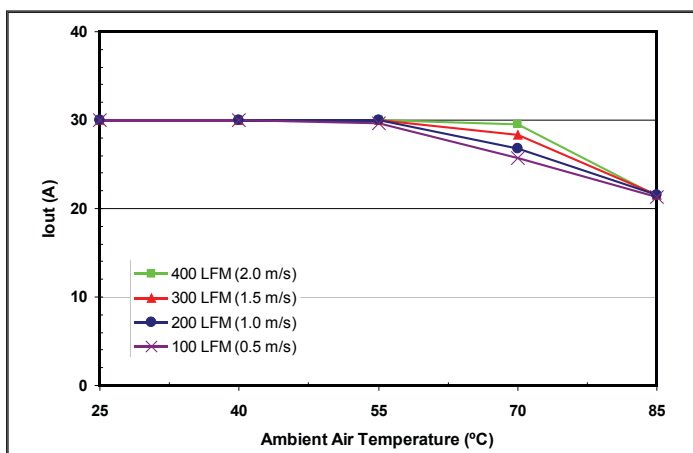


Figure 5: Maximum output power-derating curve vs. ambient air temperature for airflow rates of 100 LFM through 400 LFM with air flowing from pin 1 to pin 3 (nominal input voltage).

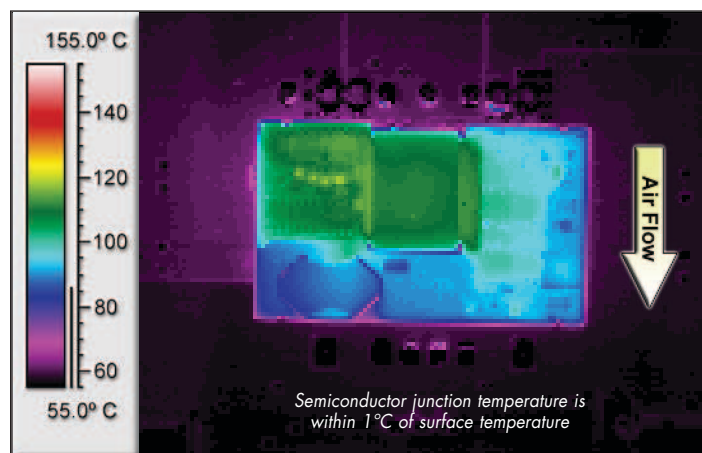


Figure 6: Thermal plot of converter at 30 A load current (268 W) with 55°C air flowing at the rate of 200 LFM. Air is flowing across the converter from pin 1 to pin 3 (nominal input voltage).





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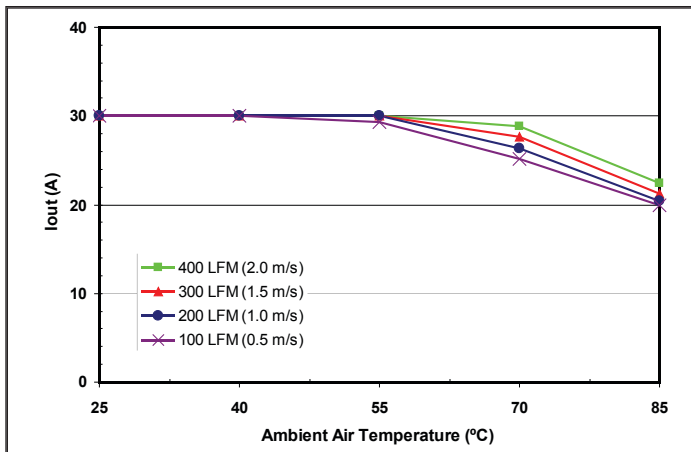


Figure 7: Maximum output power derating curves vs. ambient air temperature for airflow rates of 100 LFM through 400 LFM with air flowing across the converter from output to input (nominal input voltage).

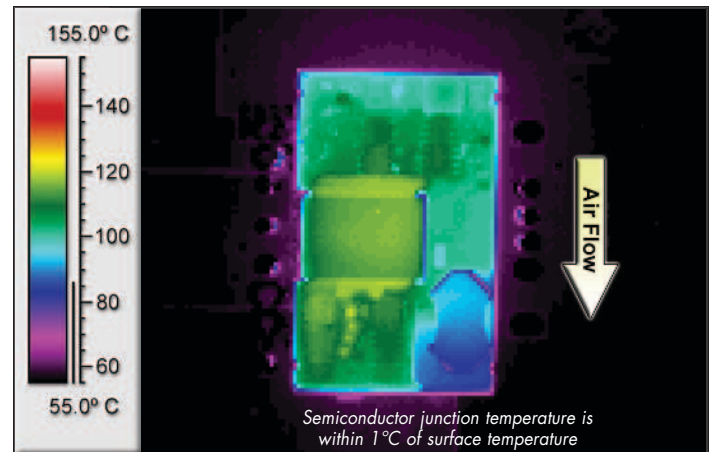


Figure 8: Thermal plot of converter at 30 A load current (268 W) with 55°C air flowing at the rate of 200 LFM. Air is flowing across the converter from output to input (nominal input voltage).

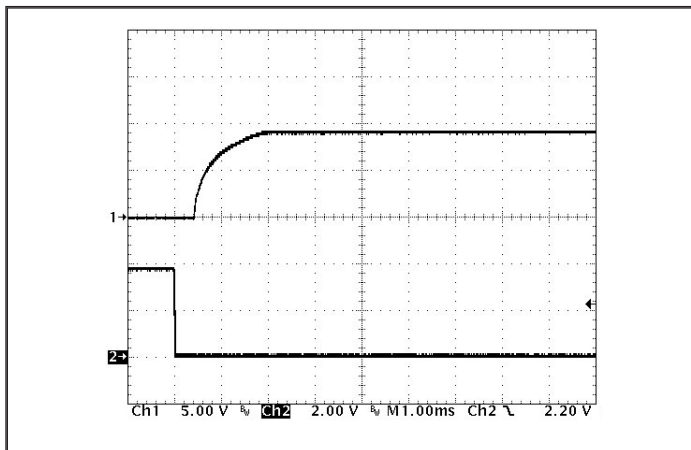


Figure 9: Turn-on transient at full load (resistive load) (1 ms/div). Load cap: 3 mF output capacitance. Input voltage pre-applied. Ch 1:  $V_{out}$  (5 V/div), Ch 2: ON/OFF input (2 V/div).

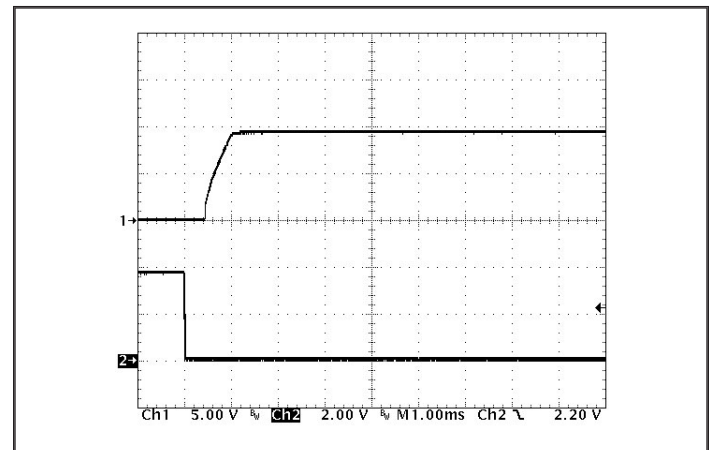


Figure 10: Turn-on transient at zero load (1 ms/div). Load cap: 3 mF output capacitance. Ch 1:  $V_{out}$  (5 mV/div), Ch 2: ON/OFF input (2 V/div).

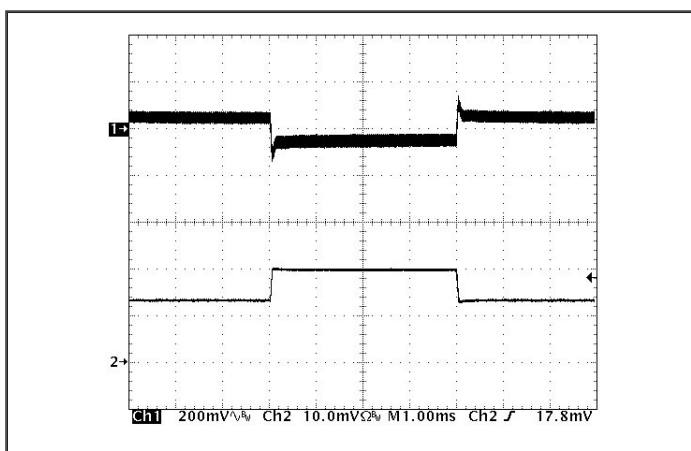


Figure 11: Output voltage response to step-change in load current (50%-75%-50% of  $I_{out(max)}$ ;  $dI/dt = 0.1A/\mu s$ ). Load cap: 15  $\mu F$  tantalum cap and 1  $\mu F$  ceramic cap. Ch 1:  $V_{out}$  (200 mV/div), Ch 2:  $I_{out}$  (10 A/div).

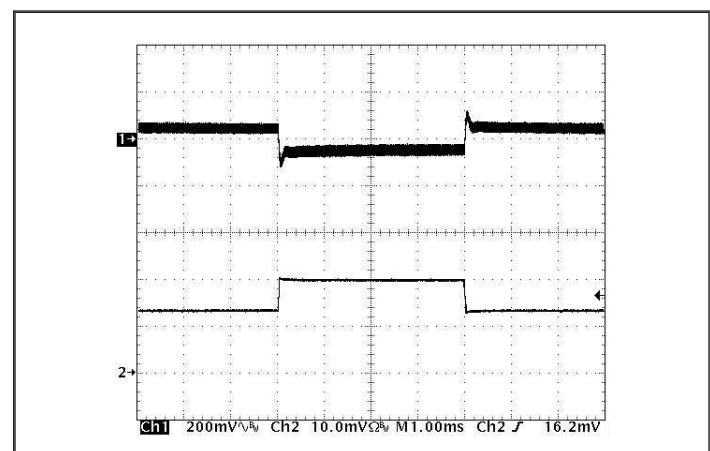


Figure 12: Output voltage response to step-change in load current (50%-75%-50% of  $I_{out(max)}$ ;  $dI/dt = 1A/\mu s$ ). Load cap: 470  $\mu F$ , 30 m $\Omega$  ESR tantalum cap and 1  $\mu F$  ceramic cap. Ch 1:  $V_{out}$  (200 mV/div), Ch 2:  $I_{out}$  (10 A/div).

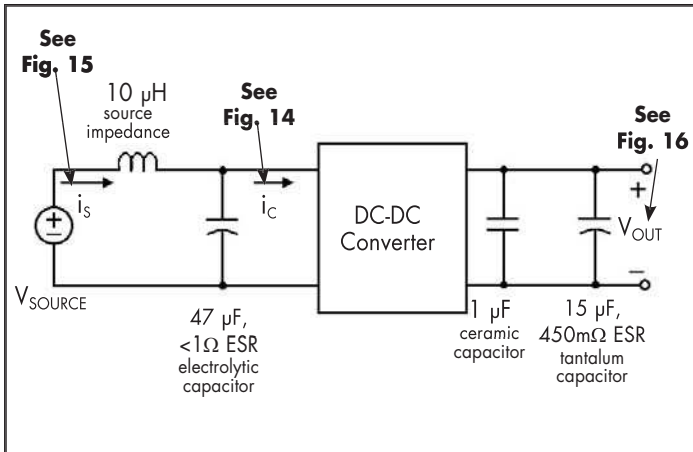


Figure 13: Test set-up diagram showing measurement points for Input Terminal Ripple Current (Figure 14), Input Reflected Ripple Current (Figure 15) and Output Voltage Ripple (Figure 16).

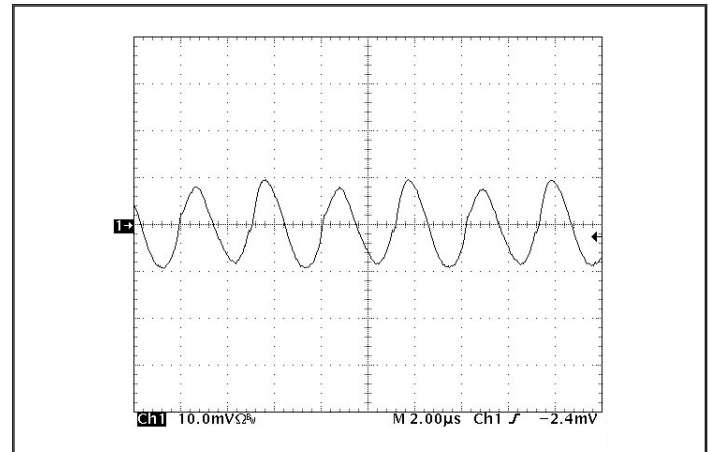


Figure 14: Input Terminal Ripple Current,  $i_c$ , at full rated output current and nominal input voltage with  $10\ \mu\text{H}$  source impedance and  $47\ \mu\text{F}$  electrolytic capacitor (200 mA/div). See Figure 13.

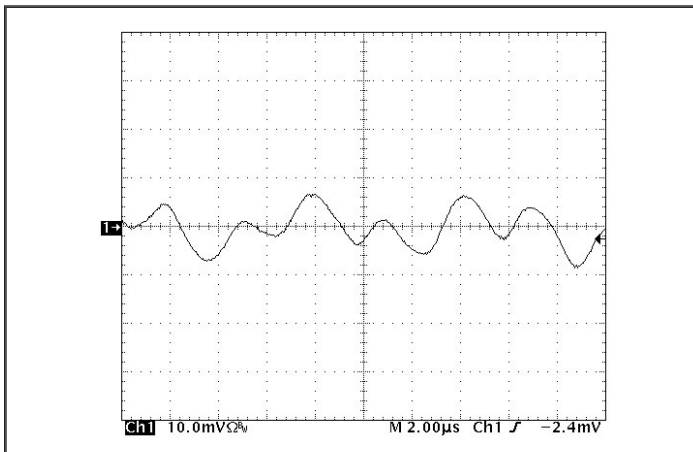


Figure 15: Input reflected ripple current,  $i_s$ , through a  $10\ \mu\text{H}$  source inductor, using a  $47\ \mu\text{F}$  electrolytic input capacitor (20 mA/div). See Figure 13.

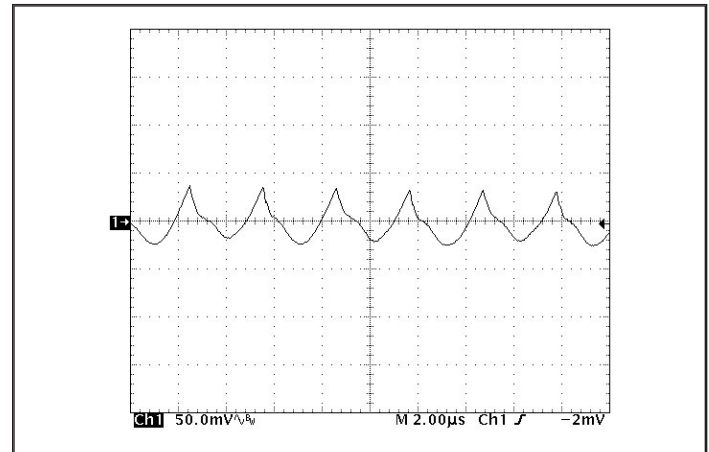


Figure 16: Output voltage ripple at nominal input voltage and rated load current (50 mV/div). Load capacitance:  $1\ \mu\text{F}$  ceramic capacitor and  $15\ \mu\text{F}$  tantalum capacitor. Bandwidth: 20 MHz. See Figure 13.

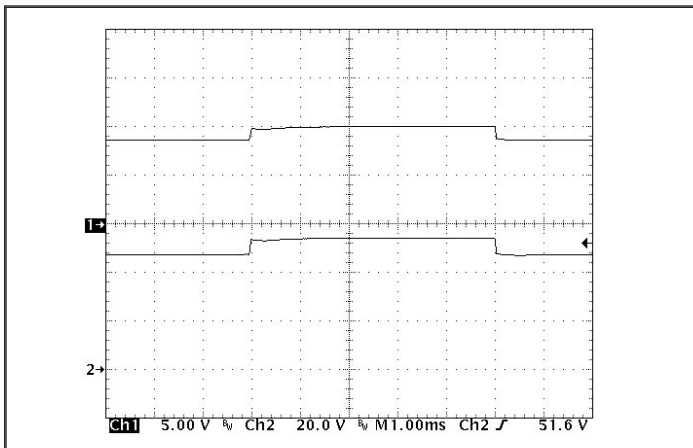


Figure 17: Output voltage response to step-change in input voltage (48V to 55V), at rated load current. Load cap:  $15\ \mu\text{F}$  tantalum capacitor and  $1\ \mu\text{F}$  ceramic cap. Ch 1:  $V_{out}$  (5 V/div), Ch 2:  $V_{in}$  (20V/div).

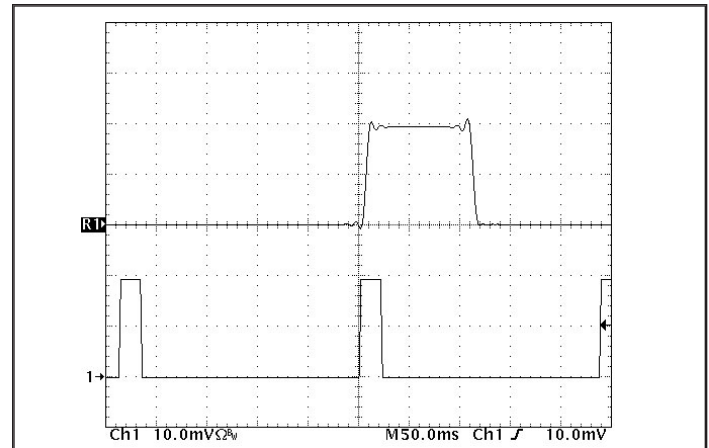


Figure 18: Load current (50 A/div) as a function of time when the converter attempts to turn on into a  $1\ \text{m}\Omega$  short circuit. Top trace (10 ms/div) is an expansion of the on-time portion of the bottom trace.



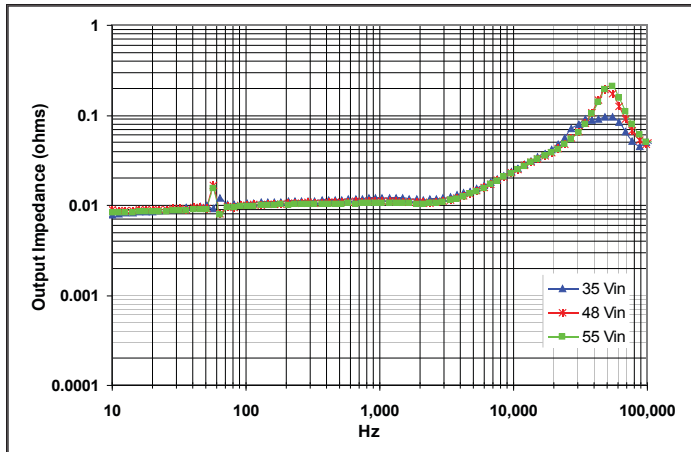


Figure 19: Magnitude of incremental output impedance ( $Z_{out} = V_{out}/I_{out}$ ) for minimum, nominal, and maximum input voltage at full rated power.

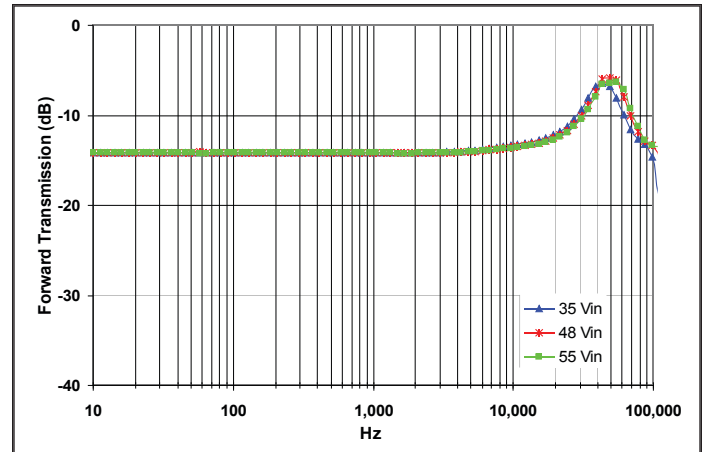


Figure 20: Magnitude of incremental forward transmission ( $FT = V_{out}/V_{in}$ ) for minimum, nominal, and maximum input voltage at full rated power.

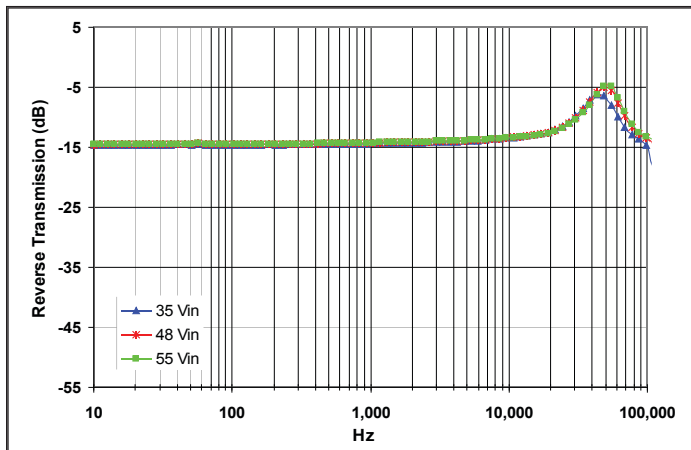


Figure 21: Magnitude of incremental reverse transmission ( $RT = I_{in}/I_{out}$ ) for minimum, nominal, and maximum input voltage at full rated power.

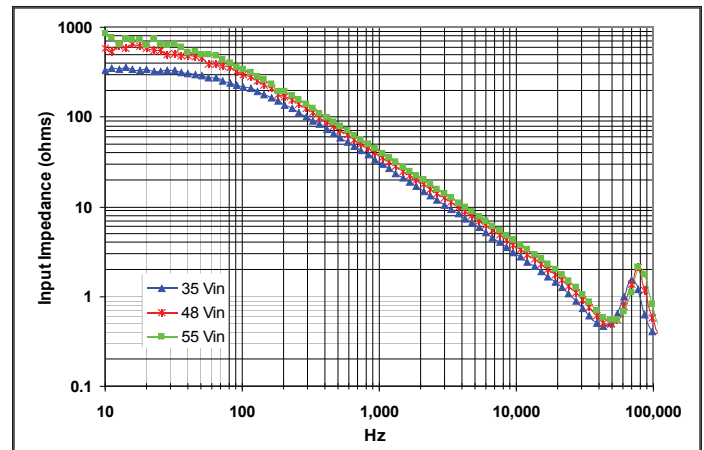


Figure 22: Magnitude of incremental input impedance ( $Z_{in} = V_{in}/I_{in}$ ) for minimum, nominal, and maximum input voltage at full rated power.

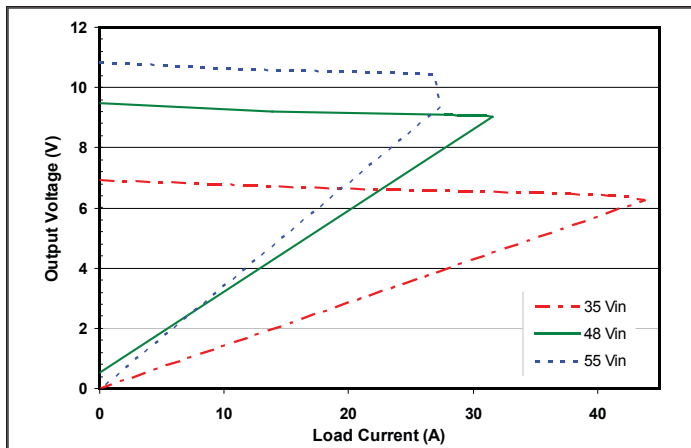


Figure 23: Output voltage vs. load current showing typical current limit curves and converter shutdown points.

## BASIC OPERATION AND FEATURES

With voltages dropping and currents rising, the economics of an Intermediate Bus Architecture (IBA) are becoming more attractive, especially in systems requiring multiple low voltages. IBA systems separate the role of isolation and voltage scaling from regulation and sensing. The BusQor series bus converter provides isolation and a coarse voltage step down in one compact module, leaving regulation to simpler, less expensive non-isolated converters.

In Figure A below, the BusQor module provides the isolation stage of the IBA system. The isolated bus then distributes power to the non-isolated buck regulators to generate the required voltage levels at the point of load. In this case, the bucks are represented with SynQor's NiQor series of non-isolated dc/dc converters. In many applications requiring multiple low voltage outputs, significant savings can be achieved in board space and overall system costs.

When designing an IBA system with bus converters, the designer can select from a variety of bus voltages. While there is no universally ideal bus voltage, most designs employ one of the following: 12V, 9V, 7.5V, 5V, or 3.3V. Higher bus voltages can lead to lower efficiency for the buck regulators but are more efficient for the bus converter and provide lower board level distribution current. Lower bus voltages offer the opposite trade offs.

SynQor's 9.6 Vout BusQor module acts as a true dc transformer. The output voltage is proportional to the input voltage, with a specified "turns ratio" or voltage ratio, plus minor drop from the internal resistive losses in the module. When used in IBA systems, the output variation of the BusQor must be in accordance with the input voltage range of the non-isolated converters being employed.

The BusQor architecture is very scalable, meaning multiple bus converters can be connected directly in parallel to allow current sharing for higher power applications.

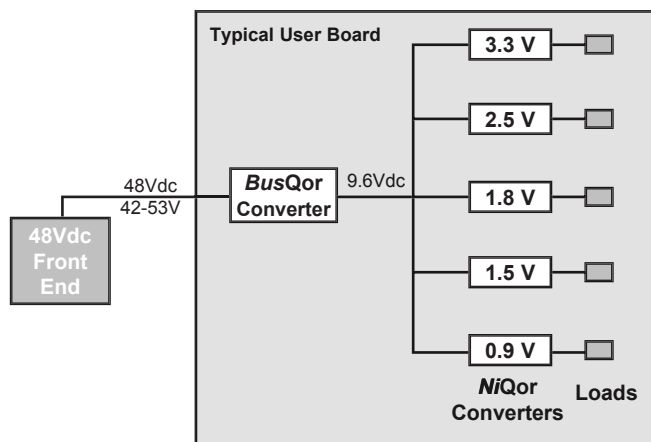


Figure A: Example of Intermediate Bus Architecture using BusQor bus converter and NiQor non-isolated converters

## CONTROL FEATURES

**REMOTE ON/OFF (Pin 2):** The ON/OFF input, Pin 2, permits the user to control when the converter is on or off. This input is referenced to the return terminal of the input bus,  $V_{in(-)}$ . There are two versions of the converter that differ by the sense of the logic used for the ON/OFF input.

In the positive logic version, the ON/OFF input is active high (meaning that a high turns the converter on). In the negative logic version, the ON/OFF signal is active low (meaning that a low turns the converter on). Figure B is a detailed look of the internal ON/OFF circuitry. Positive logic enabled modules have N-channel FET populated. Negative logic enabled modules have P-channel FET populated.

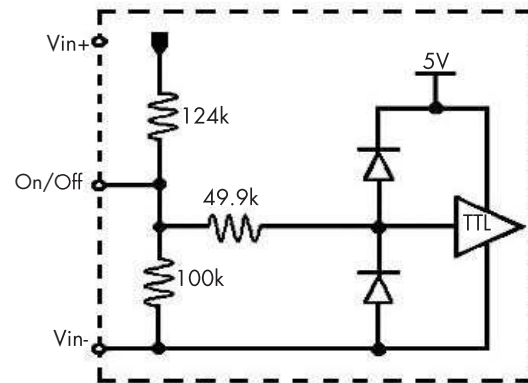


Figure B: Internal ON/OFF pin circuitry

## PROTECTION FEATURES

**Input Under-Voltage Lockout:** The converter is designed to turn off when the input voltage is too low, helping avoid an input system instability problem, described in more detail in the application note titled "Input System Instability". The lockout circuitry is a comparator with DC hysteresis. When the input voltage is rising, it must exceed the typical Turn-On Voltage Threshold value (listed on the specification page) before the converter will turn on. Once the converter is on, the input voltage must fall below the typical Turn-Off Voltage Threshold value before the converter will turn off. Also see Figure E.

**Output Current Limit:** The output of the BusQor module is electronically protected against output overloads. When an overload current greater than the "DC Current-Limit Inception" specification is drawn from the output, the output shuts down to zero volt in a period of 1ms typical (see Figure C). The shutdown period lasts for a typical period of 200 ms (Figure D) after which the BusQor tries to power up again. If the overload persists, the output voltage will go through repeated cycles of shutdown and restart with a duty cycle of 20 ms (On) and 200 ms (Off) respectively. The BusQor module returns (auto resetting) to normal operation once the overload is removed. The BusQor is designed to survive in this mode indefinitely without damage and without human intervention.

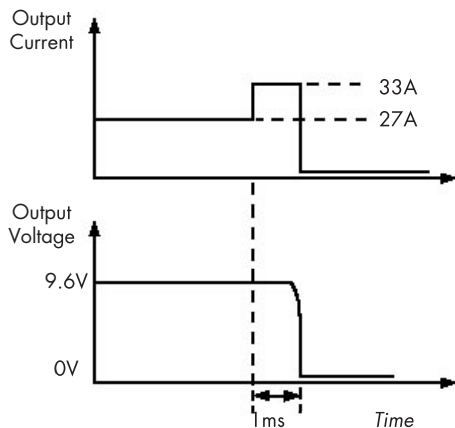


Figure C: Output Overload protection diagram (not to scale)

**Output Short Circuit Protection:** When the output of the BusQor module is shorted, a peak current of typically 95 A will flow into the short circuit for a period of about 20ms. The output of the BusQor will shutdown to zero volts for the same period (Figure E). The shutdown period lasts for a period of 200ms, at the end of which the BusQor module tries to power up again. If the short circuit persists, the output voltage will go through repeated cycles of shutdown and restart with a duty cycle of 20ms (On) and 200ms (Off) respectively. The BusQor module returns (auto resetting) to normal operation once the short circuit is removed. The BusQor is designed to survive in this mode indefinitely without damage and without human intervention. In the Auto resetting mode, also referred to as "Hiccup" mode, the power drawn from the 48V input is about 5 Watts, most of which is dissipated into the external fault. It is important that copper traces and pads from the output circuit be designed to withstand the short term peaks, although the average current into the fault may be as low as 100mA typical. See Figure 18 for appropriate waveform.

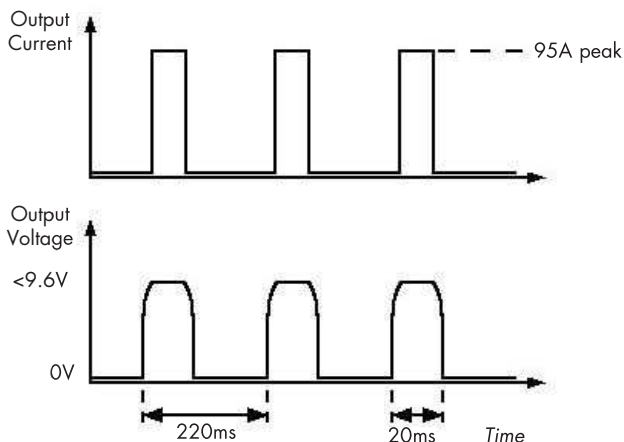


Figure D: Output Short Circuit and Auto-Resetting protection diagram (not to scale)

**Over-Temperature Shutdown:** A temperature sensor on the converter senses the average temperature of the module. The thermal shutdown circuit is designed to turn the converter off when the temperature at the sensed location reaches the Over-Temperature Shutdown value. It will allow the converter to turn on again when the temperature of the sensed location falls by the amount of the Over-Temperature Shutdown Restart Hysteresis value.

## APPLICATION CONSIDERATIONS

**Start-Up Inhibit Period:** Figure E details the Start-Up Inhibit Period for the BusQor module. At time  $t_0$ , when  $V_{in}$  is applied with On/Off pin asserted (enabled), the BusQor output begins to build up. Before time  $t_1$ , when the input voltage is below the UVL threshold, the unit is disabled by the Input Under-Voltage Lockout feature. When the input voltage rises above the UVL threshold, the Input Under-Voltage Lockout is released, and a typical Startup Inhibit Period of 3ms is initiated. The output builds up to 90% or higher of the nominal voltage and stabilizes at the nominal value of 9.6 V in a period of 3.6ms typical. The rise time from 0V to 90% is 600 $\mu$ s typical (with 3mF external output capacitance at no load). At time  $t_2$ , when the On/Off pin is de-asserted (disabled), the BusQor output instantly drops to 0V. Fall time from 9.6V to 0V is dependent on output capacitance and any parasitic trace inductance in the output load circuit. At time  $t_3$ , when the On/Off pin is re-asserted (enabled), the BusQor module output begins to build up after the inhibit period of 2.5ms typical. Refer to the Control Features section of the data sheet for details on enabling and disabling methods for N and P logic type modules.

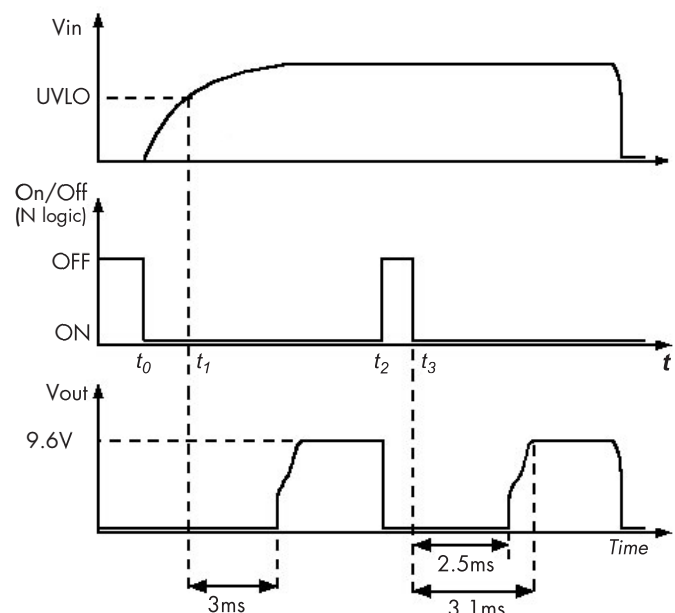


Figure E: Power Up/Down Diagram (not to scale) showing Start-Up Inhibit Period.

### Keep Out Areas:

**Component Keep Out Area:** Keep out areas for components not referenced to the Primary circuit are shown in shaded areas in Figure F. The keep out areas shown are consistent with UL's requirements for Basic Insulation of 0.04" (40 mils) for Pollution degree 2. User should consult UL standards for other insulation classes and operating environments

For applications that require mounting parts BELOW the BusQor module, one should be aware of potential high levels of electromagnetic interference, in addition to safety keep out. Users are advised to consult SynQor Applications engineering in such applications.

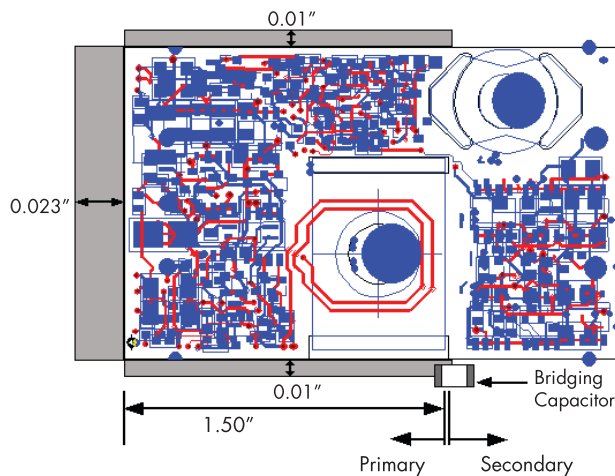


Figure F: Keep Out Areas for BusQor module

**Copper Keep Out Area:** Keep out areas shown in Figure F are to be observed for Top layer copper traces and vias. Internal layers buried one or more layers may be exempt, depending on the PCB material grade and thickness. Users are advised to consult UL standards for details. All layers including top and bottom, are subject to the keep out areas shown around Primary pins of BusQor module. Actual keep outs along the surface (Creepage) may vary depending on the PCB material CTI. Users are advised to consult UL standards for details.

**Bridging Components:** Bridging components like EMI filter capacitors required to be placed as close as possible to the BusQor module for optimum performance must observe the clearance/creepage requirements of 0.04"(40 mils) between pads to maintain compliance to UL standards for the overall power system.

**Note:** Referenced keep out widths are adequate to withstand UL's Basic Insulation Dielectric strength tests for approved PCB materials. Applications requiring Double or Reinforced insulation must double the keep out widths shown in Figure F. Keep out areas shown have standard margins above UL's minimum requirements.

**Output Load Current Calculation:** The BusQor series allows the converter output load to be measured without adding a current loop or external shunt resistor to the designer's PCB board under test. On the top side of the BusQor converter is a current sense resistor as shown in Figure G. The output load current is proportional to the voltage drop across this sense resistor. This calculation is detailed below.

A current sense resistor referenced to the primary input is used in the equation below to calculate the output load current.

$$I_{load} = (VR_{sense}(load) - VR_{sense}(no\ load)) \times 833$$

where:

$$I_{load} = \text{output load current}$$

$$VR_{sense}(load) = \text{voltage across the sense resistor with converter under load}$$

$$VR_{sense}(no\ load) = \text{voltage across the sense resistor with converter at zero load}$$

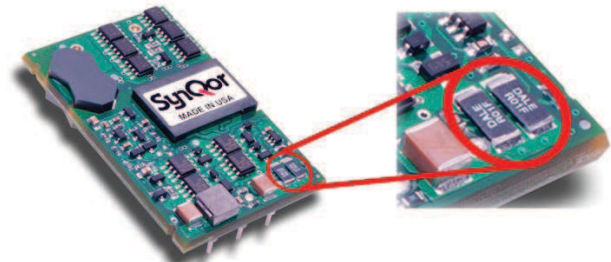


Figure G: Location of Sense Resistors

The number 833 is a "Load Current Scale Factor" for this product, valid only for  $V_{in} = 48V$ . Hence, measurement should be made at this input voltage. The voltage drop across the sense resistor is about 32.4mV at full load of 27A. Therefore, proper measuring techniques must be used to ensure accuracy. A calibrated DMM set to 300mV is acceptable. Since this measurement is done on the Primary area of the converter, a slight non-linearity may be observed over the load current range. Using the Scale Factor referenced above, the error may be on the order of 4%. For more detailed information consult the application note on SynQor's web site titled "Output Load Current Calculations".

**Current Sharing:** BusQor modules are designed to operate in parallel without the use of any external current share circuitry. A typical circuit for paralleling two BusQor modules is shown in Figure H. An output capacitor is recommended across each module and located close to the converter for optimum filtering and noise control performance. Dedicated input inductors are recommended but are considered optional. Input capacitors must be located close to the converter module. PCB layout in the input circuit should be such that high frequency ripple currents of each



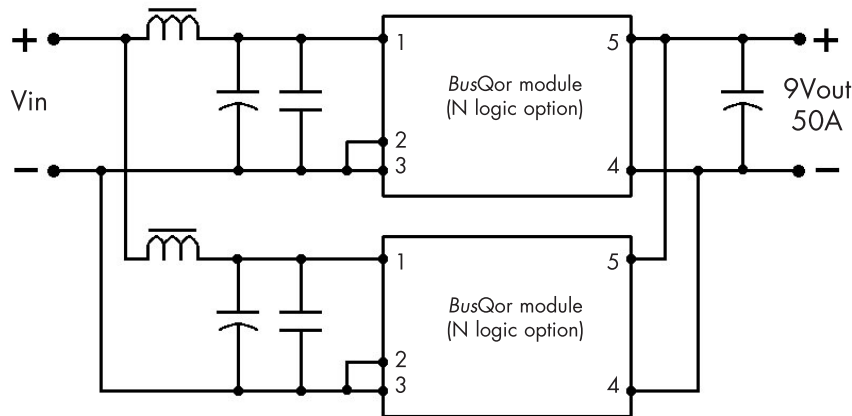


Figure H: BusQor Output Paralleling for increased current output.

module is restricted to a loop formed by the input capacitors and the input terminals of the BusQor module. See Figure J for details on PCB layout. Contact SynQor application engineering

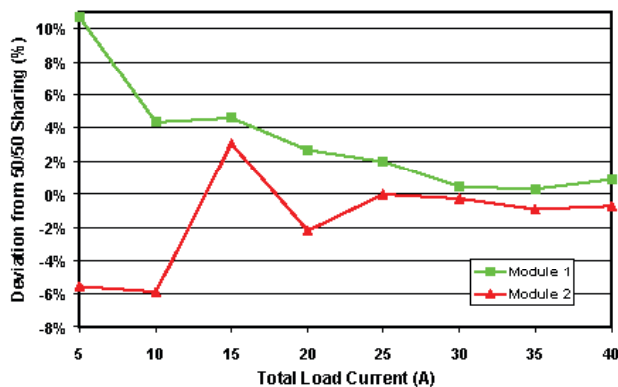


Figure I: Current share performance of 2 paralleled modules

for further assistance on PCB trace design.

The current share performance of two paralleled modules is illustrated in the graph in Figure I. In this graph the percent deviation from ideal sharing (50%) is plotted for each module versus the total output load current at 48V<sub>in</sub>.

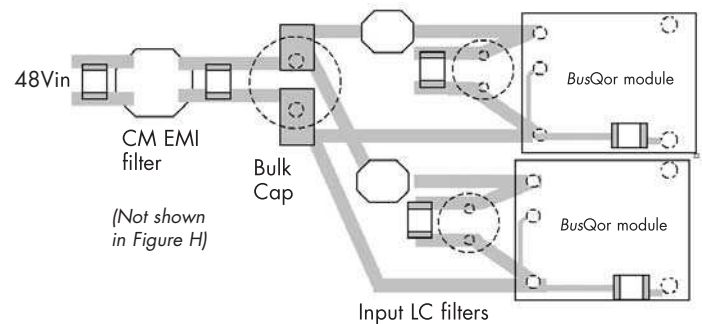


Figure J: Recommended PCB layout for input circuit



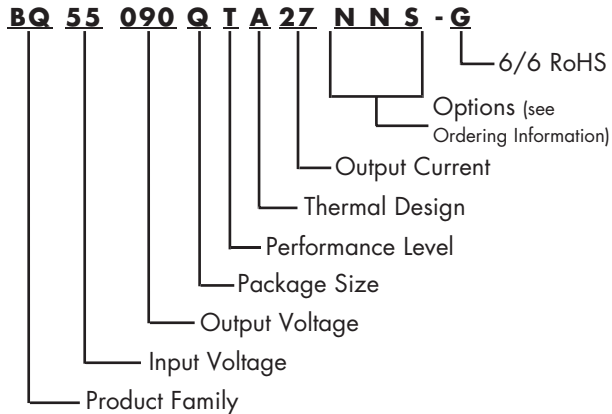


## Technical Specification

**Input:** 36-55 V  
**Output:** 9.6 V  
**Current:** 27A  
**Package:** Quarter-brick

### PART NUMBERING SYSTEM

The part numbering system for SynQor's *BusQor* DC bus converters follows the format shown in the example below.



The first 12 characters comprise the base part number and the last 3 characters indicate available options. Although there are no default values for enable logic and pin length, the most common options are negative logic and 0.145" pins. These part numbers are more likely to be readily available in stock for evaluation and prototype quantities. A "-G" suffix indicates the product is 6/6 RoHS compliant.

### PATENTS

SynQor holds the following U.S. patents, one or more of which apply to each product listed in this document. Additional patent applications may be pending or filed in the future.

5,999,417	6,222,742	6,545,890	6,577,109	6,594,159	6,731,520	6,894,468
6,896,526	6,927,987	7,050,309	7,072,190	7,085,146	7,119,524	7,269,034
7,272,021	7,272,023	7,558,083	7,564,702	7,765,687	7,787,261	8,023,290

### Contact SynQor for further information:

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**Toll Free:** 888-567-9596  
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**E-mail:** power@synqor.com  
**Web:** www.synqor.com  
**Address:** 155 Swanson Road  
 Boxborough, MA 01719  
 USA

### ORDERING INFORMATION

The tables below show the valid model numbers and ordering options for converters in this product family. When ordering SynQor converters, please ensure that you use the complete 15 character part number consisting of the 12 character base part number and the additional characters for options. A "-G" suffix indicates the product is 6/6 RoHS compliant.

Model Number	Input Voltage	Output Voltage	Max Output Current
BQ55090QTA27xyz-G	36-55 V	9.6 V	27A

The following option choices must be included in place of the x y z spaces in the model numbers listed above.

Options Description: x y z		
Enable Logic*	Pin Length	Feature Set
P - Positive N - Negative	K - 0.110" N - 0.145" R - 0.180" Y - 0.250"	S - Auto Recovery

### Application Notes

A variety of application notes and technical white papers can be downloaded in pdf format from our website.

**RoHS Compliance:** The EU led RoHS (Restriction of Hazardous Substances) Directive bans the use of Lead, Cadmium, Hexavalent Chromium, Mercury, Polybrominated Biphenyls (PBB), and Polybrominated Diphenyl Ether (PBDE) in Electrical and Electronic Equipment. This SynQor product is available as 6/6 RoHS compliant. For more information please refer to SynQor's RoHS addendum available at our RoHS Compliance / Lead Free Initiative web page or e-mail us at rohs@synqor.com.

### Warranty

SynQor offers a three (3) year limited warranty. Complete warranty information is listed on our website or is available upon request from SynQor.

Information furnished by SynQor is believed to be accurate and reliable. However, no responsibility is assumed by SynQor for its use, nor for any infringements of patents or other rights of third parties which may result from its use. No license is granted by implication or otherwise under any patent or patent rights of SynQor.