



FEATURES

- 14-bit resolution
- 1MHz sampling rate
- Functionally complete; No missing codes
- Edge-triggered; No pipeline delays
- Small 24-pin DDIP or SMT package
- Low power, 1.9 Watts maximum
- Operates from ±15V or ±12V supplies & +5V supply
- Unipolar 0 to +10V input range

PRODUCT OVERVIEW

The ADS-917 is a high-performance, 14-bit, 1MHz sampling A/D converter. This device samples input signals up to Nyquist frequencies with no missing codes. The ADS-917 features outstanding dynamic performance including a THD of –80dB.

Housed in a small 24-pin DDIP or SMT (gull-wing) package, the functionally complete ADS-917 contains a fast-settling sample-hold amplifier, a subranging (two-pass) A/D converter, a precise voltage reference, timing/control logic, and error-

correction circuitry. Digital input and output levels are TTI

Requiring $\pm 15V$ (or $\pm 12V$) and $\pm 5V$ supplies, the ADS-917 dissipates 1.9W (1.6W for $\pm 12V$) maximum. The unit is offered with a unipolar input (0 to $\pm 10V$). Models are available for use in either commercial (0 to $\pm 70^{\circ}C$) or HI-REL ($\pm 55^{\circ}C$) operating temperature ranges. Applications include radar, sonar, spectrum analysis, and graphic/medical imaging.

	INPUT/OUTPUT CONNECTIONS								
PIN	FUNCTION	PIN	FUNCTION						
1	BIT 14 (LSB)	24	-12V/-15V SUPPLY						
2	BIT 13	23	ANALOG GROUND						
3	BIT 12	22	+12V/+15V SUPPLY						
4	BIT 11	21	+10V REFERENCE OUT						
5	BIT 10	20	ANALOG INPUT						
6	BIT 9	19	ANALOG GROUND						
7	BIT 8	18	BIT 1 (MSB)						
8	BIT 7	17	BIT 2						
9	BIT 6	16	START CONVERT						
10	BIT 5	15	EOC						
11	BIT 4	14	DIGITAL GROUND						
12	BIT 3	13	+5V SUPPLY						

BLOCK DIAGRAM

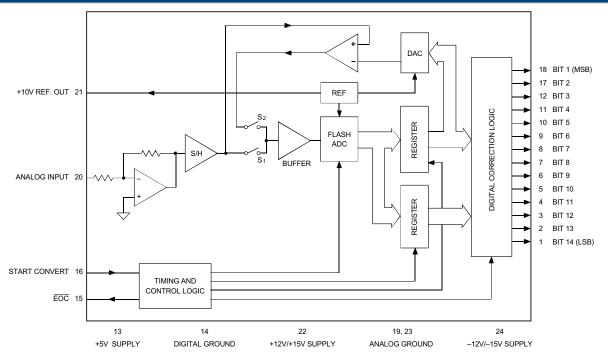


Figure 1. ADS-917 Functional Block Diagram



ABSOLUTE MAXIMUM RATINGS						
PARAMETERS	LIMITS	UNITS				
+12V/+15V Supply (Pin 22)	0 to +16	Volts				
-12V/-15V Supply (Pin 24)	0 to −16	Volts				
+5V Supply (Pin 13)	0 to +6	Volts				
Digital Input (Pin 16)	-0.3 to $+V_{DD}$ $+0.3$	Volts				
Analog Input (Pin 20)	-4 to +17	Volts				
Lead Temperature (10 second	ds) +300	°C				

PHYSICAL/ENVIRONMENTAL									
PARAMETERS	MIN.	TYP.	MAX.	UNITS					
Operating Temp. Range, Case	Operating Temp. Range, Case								
ADS-917MC, GC	0	_	+70	°C					
ADS-917MM, GM	-55	_	+125	°C					
Thermal Impedance	Thermal Impedance								
θјс	_	6	_	°C/Watt					
θса	_	24	_	°C/Watt					
Storage Temperature Range	-65	_	+150	°C					
Package Type 24-pin, metal-sealed, ceramic DDIP or SMT									
Weight	0.42 ounces (12 grams)								

FUNCTIONAL SPECIFICATIONS

 $(T_A = +25^{\circ}C, \pm V_{CC} = \pm 15V)$ (or $\pm 12V$), $\pm V_{DD} = +5V$, 1MHz sampling rate, and a minimum 1 minute warmup ① unless otherwise specified.)

		+25°C			0 TO +70°C			–55 TO +125°	С	
ANALOG INPUT	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	UNITS
Input Voltage Range ②	_	0 to +10	_	_	0 to +10	_	_	0 to +10	_	Volts
Input Resistance	_	1	_	_	1	_	_	1	_	kΩ
Input Capacitance	_	7	15	_	7	15	_	7	15	pF
DIGITAL INPUT										
Logic Levels										
Logic "1"	+2.0	_	_	+2.0	_	_	+2.0	_	_	Volts
Logic "0"	_	_	+0.8	_	_	+0.8	_	_	+0.8	Volts
Logic Loading "1"	_	_	+20	_	_	+20	_	_	+20	μΑ
Logic Loading "0"	_	_	-20		_	-20		_	-20	μΑ
Start Convert Positive Pulse Width ③	20	200	_	20	200	_	20	200	_	ns
STATIC PERFORMANCE										
Resolution	_	14	_	_	14	_	_	14	_	Bits
Integral Nonlinearity (fin = 10kHz)	_	±0.5	_	_	±0.75	_	_	±1.5	_	LSB
Differential Nonlinearity (fin = 10kHz)	_	±0.5	+0.95	_	±0.5	±0.95	-0.95	±0.75	+1.25	LSB
Full Scale Absolute Accuracy	_	±0.05	±0.1	_	±0.1	±0.2	_	±0.15	±0.4	%FSR
Unipolar Offset Error (Tech Note 2)	_	±0.1	±0.2	_	±0.1	±0.2	_	±0.15	±0.4	%FSR
Gain Error (Tech Note 2)		±0.1	±0.25		±0.1	±.025		±0.25	±0.4	%
No Missing Codes (fin = 10kHz)	14	_	_	14	_	_	14	_	_	Bits
DYNAMIC PERFORMANCE										
Peak Harmonics (-0.5dB)										
dc to 100kHz	_	-87	-82	_	-87	-82	_	-85	-80	dB
100kHz to 500kHz		-81	-76		-81	-76		- 79	-74	dB
Total Harmonic Distortion (–0.5dB)										
dc to 100kHz	_	-85	-82	_	-85	-82	_	-84	-80	dB
100kHz to 500kHz		-80	-76		-80	-76		-79	-74	dB
Signal-to-Noise Ratio (w/o distortion, -0.5dB)										
dc to 100kHz	75	79		75	79		73	77		dB
100kHz to 500kHz	73	78		73	78		73 72	7 <i>7</i>		dB
Signal-to-Noise Ratio 4	73	70		13	70		12	70		ub
(& distortion, -0.5dB)										
dc to 100kHz	74	77	_	74	77	_	72	76	_	dB
100kHz to 500kHz	72	76	_	72	76		71	75	_	dB
Two-tone Intermodulation				·						
Distortion ($f_{in} = 100kHz$,										
$500kHz$, $f_s = 1MHz - 0.5dB$)		-87			-86			-85		dB
Noise	_	300	_	_	400	_		600	_	μVrms
Input Bandwidth (–3dB)										
Small Signal (-20dB input)	_	7	_	_	7	_	_	7	_	MHz
Large Signal (-0.5dB input)		5			5			5		MHz
Feedthrough Rejection (fin = 500kHz)		84			84			84		dB
Slew Rate		±60			±60			±60		V/µs
Aperture Delay Time		±20			±20			±20		ns
Aperture Uncertainty		5			5			5		ps rms
S/H Acquisition Time	F00	F70	010	F00	F70	010	F00	F70	010	
(to ±0.003%FSR, 10V step)	530	570	610	530	570	610	530	570	610	ns
Overvoltage Recovery Time (5)		400	1000		400	1000		400	1000	ns Mu-
A/D Conversion Rate	I			11			1			MHz

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		+25°C			0 to +70°C		-	-55 to +125°	C	
ANALOG OUTPUT	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	UNITS
Internal Reference										
Voltage	+9.95	+10.0	+10.05	+9.95	+10.0	+10.05	+9.95	+10.0	+10.05	Volts
Drift	_	±5	_	_	±5	_	_	±5	_	ppm/°C
External Current	_	_	1.5	_	_	1.5	_	_	1.5	mA
DIGITAL OUTPUTS										
Logic Levels										
Logic "1"	+2.4		_	+2.4	_	_	+2.4	_	_	Volts
Logic "0"	_	_	+0.4	_	_	+0.4	_	_	+0.4	Volts
Logic Loading "1"	_	_	-4	_	_	-4	_	_	-4	mA
Logic Loading "0"	_	_	+4	_	_	+4	_	_	+4	mA
Delay, Falling Edge of EOC to										
Output Data Valid	_	_	35	_	_	35	_	_	35	ns
Output Coding				5	Straight Binar	у				
POWER REQUIREMENTS , ±15V										
Power Supply Ranges										
+15V Supply	+14.5	+15.0	+15.5	+14.5	+15.0	+15.5	+14.5	+15.0	+15.5	Volts
-15V Supply	-14.5	-15.0	-15.5	-14.5	-15.0	-15.5	-14.5	-15.0	-15.5	Volts
+5V Supply	+4.75	+5.0	+5.25	+4.75	+5.0	+5.25	+4.75	+5.0	+5.25	Volts
Power Supply Currents										
+15V Supply	_	+50	+65	_	+50	+65	_	+50	+65	mA
-15V Supply	_	-41	-50	_	-41	-50	_	-41	-50	mA
+5V Supply	_	+70	+85		+70	+85	_	+70	+85	mA
Power Dissipation	_	1.7	1.9	_	1.7	1.9	_	1.7	1.9	Watts
Power Supply Rejection	_	_	±0.01	_	_	±0.01	_	_	±0.01	%FSR/%V
POWER REQUIREMENTS, ±12V										
Power Supply Ranges										
+12V Supply	+11.5	+12.0	+12.5	+11.5	+12.0	+12.5	+11.5	+12.0	+12.5	Volts
–12V Supply	-11.5	-12.0	-12.5	-11.5	-12.0	-12.5	-11.5	-12.0	-12.5	Volts
+5V Supply	+4.75	+5.0	+5.25	+4.75	+5.0	+5.25	+4.75	+5.0	+5.25	Volts
Power Supply Currents										
+12V Supply	_	+50	+65	_	+50	+65	_	+50	+65	mA
-12V Supply	_	-40	-48	_	-40	-48	_	-40	-48	mA
+5V Supply	_	+70	+80	_	+70	+80	_	+70	+80	mA
Power Dissipation	_	1.4	1.6	_	1.4	1.6	_	1.4	1.6	Watts
Power Supply Rejection		_	±0.01			±0.01			±0.01	%FSR/%V

Footnotes:

① All power supplies must be on before applying a start convert pulse. All supplies and the clock (START CONVERT) must be present during warmup periods. The device must be continuously converting during this time. There is a slight degradation in performance when using ±12V supplies.

- ② See Ordering Information for availability of ±5V input range. Contact DATEL for availability of other input voltage ranges.
- ③ A 1MHz clock with a 200ns wide start convert pulse is used for all production testing. See Timing Diagram for more details.

Effective bits is equal to:

⑤ This is the time required before the A/D output data is valid once the analog input is back within the specified range.

TECHNICAL NOTES

Obtaining fully specified performance from the ADS-917 requires careful attention to pc-card layout and power supply decoupling. The device's analog and digital ground systems are connected to each other internally. For optimal performance, tie all ground pins (14, 19 and 23) directly to a large analog ground plane beneath the package.

Bypass all power supplies and the REFERENCE OUTPUT (pin 21) to ground with $4.7\mu\text{F}$ tantalum capacitors. Locate the bypass capacitors as close to the unit as possible. If the user-installed offset and gain adjusting circuit shown in Figure 2 is used, also locate it as close to the ADS-917 as possible.

2. The ADS-917 achieves its specified accuracies without the need for external calibration. If required, the device's small initial offset and gain errors

can be reduced to zero using the input circuit of Figure 2. When using this circuit, or any similar offset and gain-calibration hardware, make adjustments following warmup. To avoid interaction, always adjust offset before gain.

- When operating the ADS-917 from ±12V supplies, do not drive external
 circuitry with the REFERENCE OUTPUT. The reference's accuracy and drift
 specifications may not be met, and loading the circuit may cause accuracy
 errors within the converter.
- Applying a start convert pulse while a conversion is in progress (EOC = logic "1") initiates a new and inaccurate conversion cycle. Data from the interrupted and subsequent conversions will be invalid.

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CALIBRATION PROCEDURE (Refer to Figures 2 and 3)

Any offset and/or gain calibration procedures should not be implemented until devices are fully warmed up. To avoid interaction, offset must be adjusted before gain. The ranges of adjustment for the circuit of Figure 2 are guaranteed to compensate for the ADS-917's initial accuracy errors and may not be able to compensate for additional system errors.

All fixed resistors in Figure 2 should be metal-film types, and multiturn potentiometers should have TCR's of 100ppm/°C or less to minimize drift with temperature.

A/D converters are calibrated by positioning their digital outputs exactly on the transition point between two adjacent digital output codes. This can be accomplished by connecting LED's to the digital outputs and adjusting until certain LED's "flicker" equally between on and off. Other approaches employ digital comparators or microcontrollers to detect when the outputs change from one code to the next.

For the ADS-917, offset adjusting is normally accomplished at the point where the output bits are 0's and the LSB just changes from a 0 to a 1. This digital output transition ideally occurs when the applied analog input is $\pm 1/2$ LSB ($\pm 305 \mu V$).

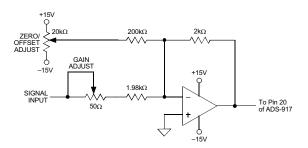


Figure 2. ADS-917 Calibration Circuit

Gain adjusting is accomplished when all bits are 1's and the LSB just changes from a 1 to a 0. This transition ideally occurs when the analog input is at +full scale minus $1\frac{1}{2}$ LSB's (+9.999085V).

Zero/Offset Adjust Procedure

- Apply a train of pulses to the START CONVERT input (pin 16) so the converter is continuously converting. If using LED's on the outputs, a 200kHz conversion rate will reduce flicker.
- 2. Apply +305μV to the ANALOG INPUT (pin 20).
- Adjust the offset potentiometer until the output bits are all 0's and the LSB flickers between 0 and 1.

Gain Adjust Procedure

- 1. Apply +9.999085V to the ANALOG INPUT (pin 20).
- Adjust the gain potentiometer until the output bits are all 1's and the LSB flickers between 1 and 0.

Table 1. Zero and Gain Adjust

INPUT VOLT-	ZERO ADJUST	GAIN ADJUST		
AGE RANGE	+½ LSB	+FS –1½ LSB		
0 to +10V	+305µV	+9.999085V		

Table 2. Output Coding

INPUT VOLTAGE		DIGITAL OUTPUT			
(0 TO +10V)	UNIPOLAR SCALE	OFFSET MSB	BINARY LSB		
+9.999390	+FS -1LSB	11 1111	1111 1111		
+7.500000	+3/4 FS	11 0000	0000 0000		
+5.000000	+1/2 FS	10 0000	0000 0000		
+2.500000	+1/4 FS	01 0000	0000 0000		
+0.000610	+1LSB	00 0000	0000 0001		
0.000000	0	00 0000	0000 0000		

Coding is straight binary; $1LSB = 610\mu V$.

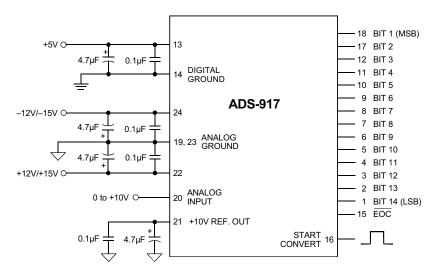


Figure 3. Typical ADS-917 Connection Diagram



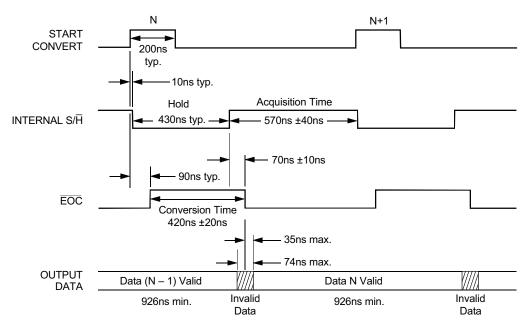
THERMAL REQUIREMENTS

All DATEL sampling A/D converters are fully characterized and specified over operating temperature (case) ranges of 0 to $+70^{\circ}$ C and -55 to $+125^{\circ}$ C. All room-temperature (TA = $+25^{\circ}$ C) production testing is performed without the use of heat sinks or forced-air cooling. Thermal impedance figures for each device are listed in their respective specification tables.

These devices do not normally require heat sinks, however, standard precautionary design and layout procedures should be used to ensure

devices do not overheat. The ground and power planes beneath the package, as well as all pcb signal runs to and from the device, should be as heavy as possible to help conduct heat away from the package.

Electrically-insulating, thermally-conductive "pads" may be installed underneath the package. Devices should be soldered to boards rather than "socketed," and of course, minimal air flow over the surface can greatly help reduce the package temperature.



Notes: 1. $f_s = 500kHz$.

2. The ADS-917 is an edge-triggered device. All internal operations are triggered by the rising edge of the start convert pulse, which may be as narrow as 20nsec. All production testing is performed at a 1MHz sampling rate with 200nsec wide start pulses. For lower sampling rates, wider start pulses may be used, however, a minimum pulse width low of 20nsec must be maintained.

Figure 4. ADS-917 Timing Diagram



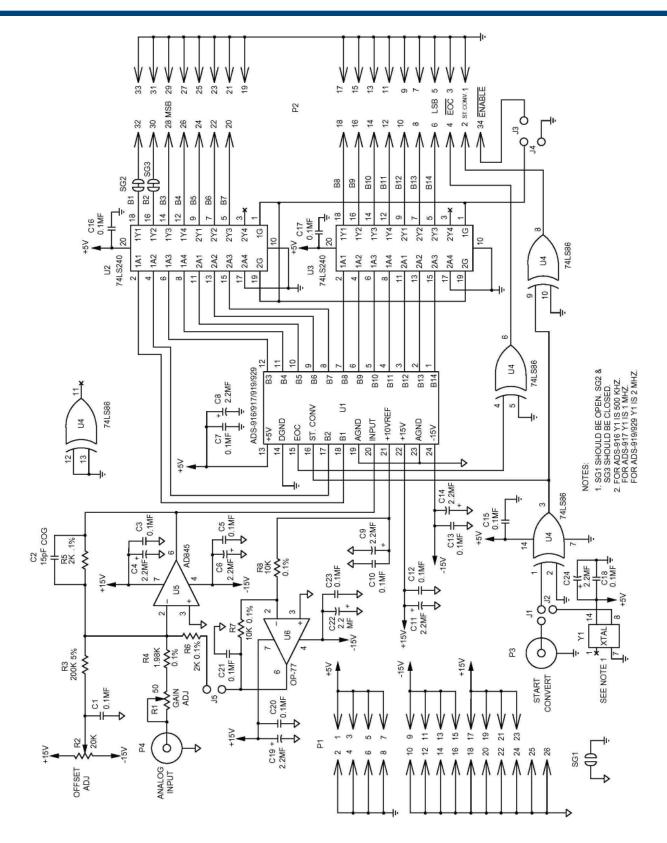


Figure 5. ADS-917 Evaluation Board Schematic



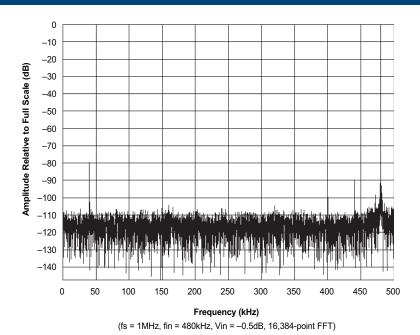


Figure 6. ADS-917 FFT Analysis

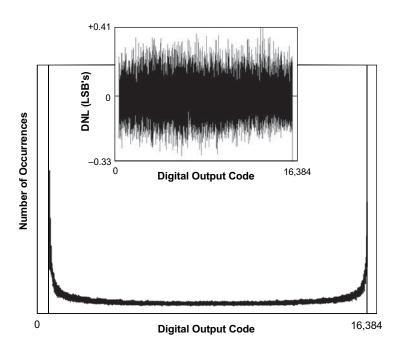
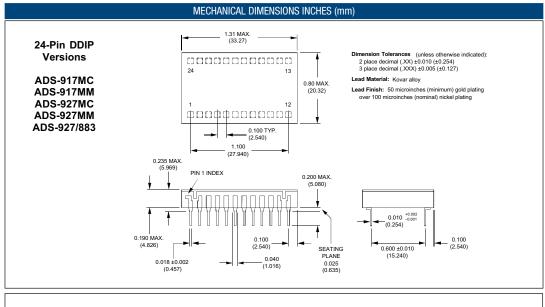
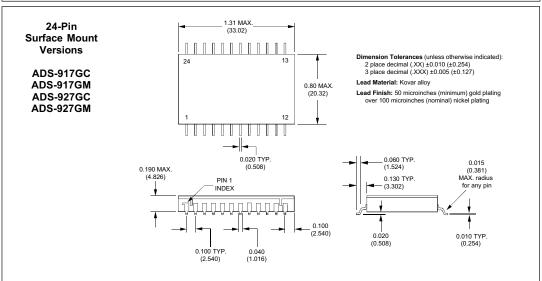


Figure 7. ADS-917 Histogram and Differential Nonlinearity







ORDERING INFORMATION									
MODEL NUMBER	OPERATING TEMP. RANGE	ANALOG INPUT	ACCESSORIES						
ADS-917MC	0 to +70°C	Unipolar (0 to +10V)	ADS-B916/917	Evaluation Board (without ADS-917)					
ADS-917MM	−55 to +125°C	Unipolar (0 to +10V)	HS-24	Heat Sinks for all ADS-917/927 DDIP models					
ADS-917GC	0 to +70°C	Unipolar (0 to +10V)							
ADS-917GM	−55 to +125°C	Unipolar (0 to +10V)	Receptacles for PC board mounting can be ordered through AMP In Part #3-331272-8 (Component Lead Socket), 24 required. For MIL-STD-883 product specifications, contact DATEL. * For information, see ADS-927 data sheet.						
ADS-927MC	0 to +70°C	Bipolar (±5V)*							
ADS-927MM	−55 to +125°C	Bipolar (±5V)*							
ADS-927GC	0 to +70°C	Bipolar (±5V)*							
ADS-927GM	−55 to +125°C	Bipolar (±5V)*	1						
ADS-927/883	−55 to +125°C	Bipolar (±5V)*							

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