



# FEMTOCLOCKS™ CRYSTAL-TO-LVDS FREQUENCY SYNTHESIZER

ICS844002

## GENERAL DESCRIPTION



The ICS844002 is a 2 output LVDS Synthesizer optimized to generate Fibre Channel reference clock frequencies and is a member of the HiPerClocks™ family of high performance clock solutions from IDT. Using a 26.5625MHz 18pF parallel resonant crystal, the following frequencies can be generated based on the 2 frequency select pins (F\_SEL[1:0]): 212.5MHz, 187.5MHz, 159.375MHz, 106.25MHz and 53.125MHz. The ICS844002 uses IDT's 3<sup>rd</sup> generation low phase noise VCO technology and can achieve <1ps typical rms phase jitter, easily meeting Fibre Channel jitter requirements. The ICS844002 is packaged in a small 20-pin TSSOP package.

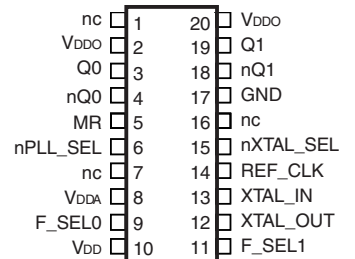
## FEATURES

- Two LVDS outputs
- Selectable crystal oscillator interface or LVCMOS/LVTTL single-ended input
- Supports the following output frequencies: 212.5MHz, 187.5MHz, 159.375MHz, 106.25MHz and 53.125MHz
- VCO range: 560MHz - 680MHz
- RMS phase jitter @ 212.5MHz, using a 26.5625MHz crystal (637kHz - 10MHz): 0.65ps (typical)
- Full 3.3V or 2.5V supply modes
- 0°C to 70°C ambient operating temperature
- Available in both standard (RoHS 5) and lead-free (RoHS 6) packages

FREQUENCY SELECT FUNCTION TABLE

Input Frequency (MHz)	Inputs					Output Frequency (MHz)
	F_SEL1	F_SEL0	M Divider Value	N Divider Value	M/N Divider Value	
26.5625	0	0	24	3	8	212.5 (default)
26.5625	0	1	24	4	6	159.375
26.5625	1	0	24	6	4	106.25
26.5625	1	1	24	12	2	53.125
23.4375	0	0	24	3	8	187.5 (default)

## PIN ASSIGNMENT



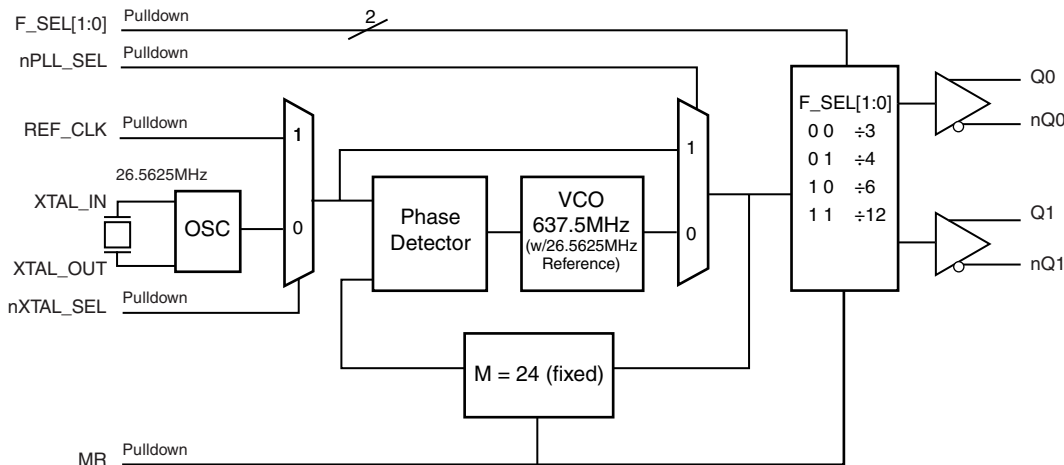
### ICS844002

20-Lead TSSOP

6.5mm x 4.4mm x 0.925mm package body

G Package  
Top View

## BLOCK DIAGRAM



**TABLE 1. PIN DESCRIPTIONS**

Number	Name	Type		Description
1, 7	nc	Unused		No connect.
2, 20	V <sub>DDO</sub>	Power		Output supply pins.
3, 4	Q0, nQ0	Ouput		Differential output pair. LVDS interface levels.
5	MR	Input	Pulldown	Active HIGH Master Reset. When logic HIGH, the internal dividers are reset causing the true outputs Qx to go low and the inverted outputs nQx to go high. When logic LOW, the internal dividers and the outputs are enabled. LVCMOS/LVTTL interface levels.
6	nPLL_SEL	Input	Pulldown	Selects between the PLL and REF_CLK as input to the dividers. When LOW, selects PLL (PLL Enable). When HIGH, deselects the reference clock (PLL Bypass). LVCMOS/LVTTL interface levels.
8	V <sub>DDA</sub>	Power		Analog supply pin.
9, 11	F_SEL0, F_SEL1	Input	Pulldown	Frequency select pins. LVCMOS/LVTTL interface levels.
10	V <sub>DD</sub>	Power		Core supply pins.
12, 13	XTAL_OUT, XTAL_IN	Input		Parallel resonant crystal interface. XTAL_OUT is the output, XTAL_IN is the input.
14	REF_CLK	Input	Pulldown	LVCMOS/LVTTL reference clock input.
15	nXTAL_SEL	Input	Pulldown	Selects between crystal or REF_CLK inputs as the the PLL Reference source. Selects XTAL inputs when LOW. Selects REF_CLK when HIGH. LVCMOS/LVTTL interface levels.
16	nc	Unused		No connect.
17	GND	Power		Power supply ground.
18, 19	nQ1, Q1	Output		Differential output pair. LVDS interface levels.

NOTE: *Pulldown* refers to internal input resistors. See Table 2, Pin Characteristics, for typical values.

**TABLE 2. PIN CHARACTERISTICS**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
C <sub>IN</sub>	Input Capacitance			4		pF
R <sub>PULLDOWN</sub>	Input Pulldown Resistor			51		kΩ

## ABSOLUTE MAXIMUM RATINGS

Supply Voltage, $V_{DD}$	4.6V
Inputs, $V_I$	-0.5V to $V_{DD} + 0.5V$
Outputs, $I_O$	
Continuous Current	10mA
Surge Current	15mA
Package Thermal Impedance, $\theta_{JA}$	73.2°C/W (0 lfpm)
Storage Temperature, $T_{STG}$	-65°C to 150°C

NOTE: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

**TABLE 3A. POWER SUPPLY DC CHARACTERISTICS,  $V_{DD} = V_{DDO} = 3.3V \pm 5\%$ ,  $T_A = 0^\circ C$  TO  $70^\circ C$**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{DD}$	Core Supply Voltage		3.135	3.3	3.465	V
$V_{DDA}$	Analog Supply Voltage		$V_{DD} - 0.12$	3.3	$V_{DD}$	V
$V_{DDO}$	Output Supply Voltage		3.135	3.3	3.465	V
$I_{DD}$	Power Supply Current				105	mA
$I_{DDA}$	Analog Supply Current				12	mA
$I_{DDO}$	Output Supply Current				120	mA

**TABLE 3B. POWER SUPPLY DC CHARACTERISTICS,  $V_{DD} = V_{DDO} = 2.5V \pm 5\%$ ,  $T_A = 0^\circ C$  TO  $70^\circ C$**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{DD}$	Core Supply Voltage		2.375	2.5	2.625	V
$V_{DDA}$	Analog Supply Voltage		$V_{DD} - 0.10$	2.5	$V_{DD}$	V
$V_{DDO}$	Output Supply Voltage		2.375	2.5	2.625	V
$I_{DD}$	Power Supply Current				95	mA
$I_{DDA}$	Analog Supply Current				10	mA
$I_{DDO}$	Output Supply Current				90	mA

**TABLE 3C. LVCMOS / LVTTTL DC CHARACTERISTICS,  $V_{DD} = V_{DDO} = 3.3V \pm 5\%$  OR  $2.5V \pm 5\%$ ,  $T_A = 0^\circ C$  TO  $70^\circ C$**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{IH}$	Input High Voltage	$V_{DD} = 3.3V$	2		$V_{DD} + 0.3$	V
		$V_{DD} = 2.5V$	1.7		$V_{DD} + 0.3$	V
$V_{IL}$	Input Low Voltage	$V_{DD} = 3.3V$	-0.3		0.8	V
		$V_{DD} = 2.5V$	-0.3		0.7	V
$I_{IH}$	Input High Current	REF_CLK, MR, F_SEL0, F_SEL1, nPLL_SEL, nXTAL_SEL, $V_{DD} = V_{IN} = 3.465$ or 2.5V			150	$\mu A$
$I_{IL}$	Input Low Current	REF_CLK, MR, F_SEL0, F_SEL1, nPLL_SEL, nXTAL_SEL, $V_{DD} = 3.465V$ or 2.5V, $V_{IN} = 0V$	-150			$\mu A$

**TABLE 3D. LVDS DC CHARACTERISTICS,  $V_{DD} = V_{DDO} = 3.3V \pm 5\%$ ,  $T_A = 0^\circ C$  TO  $70^\circ C$** 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{OD}$	Differential Output Voltage		300	450	600	mV
$\Delta V_{OD}$	$V_{OD}$ Magnitude Change				50	mV
$V_{OS}$	Offset Voltage		1.4	1.525	1.65	V
$\Delta V_{OS}$	$V_{OS}$ Magnitude Change				50	mV

**TABLE 3E. LVDS DC CHARACTERISTICS,  $V_{DD} = V_{DDO} = 2.5V \pm 5\%$ ,  $T_A = 0^\circ C$  TO  $70^\circ C$** 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{OD}$	Differential Output Voltage		250	400	550	mV
$\Delta V_{OD}$	$V_{OD}$ Magnitude Change				50	mV
$V_{OS}$	Offset Voltage		1.0		1.4	V
$\Delta V_{OS}$	$V_{OS}$ Magnitude Change				50	mV

**TABLE 4. CRYSTAL CHARACTERISTICS**

Parameter	Test Conditions	Minimum	Typical	Maximum	Units
Mode of Oscillation		Fundamental			
Frequency		23.33	26.5625	28.33	MHz
Equivalent Series Resistance (ESR)				50	$\Omega$
Shunt Capacitance				7	pF
Drive Level				1	mW

NOTE: Characterized using an 18pF parallel resonant crystal.

**TABLE 5A. AC CHARACTERISTICS,  $V_{DD} = V_{DDO} = 3.3V \pm 5\%$ ,  $T_A = 0^\circ C$  TO  $70^\circ C$**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$f_{OUT}$	Output Frequency	F_SEL[1:0] = 00	186.67		226.66	MHz
		F_SEL[1:0] = 01	140		170	MHz
		F_SEL[1:0] = 10	93.33		113.33	MHz
		F_SEL[1:0] = 11	46.67		56.66	MHz
$t_{sk(o)}$	Output Skew; NOTE 1, 2			15	ps	
$f_{jit}(\emptyset)$	RMS Phase Jitter (Random); NOTE 3	212.5MHz, (637kHz - 10MHz)		0.65		ps
		159.375MHz, (637kHz - 10MHz)		0.61		ps
		106.25MHz, (637kHz - 10MHz)		0.74		ps
		53.125MHz, (637kHz - 10MHz)		0.64		ps
		187.5MHz, (637kHz - 10MHz)		0.80		ps
$t_R / t_F$	Output Rise/Fall Time	20% to 80%	250		500	ps
odc	Output Duty Cycle	F_SEL[1:0] $\neq$ $\div 3$	48		52	%
		F_SEL[1:0] = $\div 3$	45		55	%

NOTE 1: Defined as skew between outputs at the same supply voltages and with equal load conditions.

Measured at the differential cross points.

NOTE 2: This parameter is defined in accordance with JEDEC Standard 65.

NOTE 3: Please refer to the Phase Noise Plot.

**TABLE 5B. AC CHARACTERISTICS,  $V_{DD} = V_{DDO} = 2.5V \pm 5\%$ ,  $T_A = 0^\circ C$  TO  $70^\circ C$**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$f_{OUT}$	Output Frequency	F_SEL[1:0] = 00	186.67		226.66	MHz
		F_SEL[1:0] = 01	140		170	MHz
		F_SEL[1:0] = 10	93.33		113.33	MHz
		F_SEL[1:0] = 11	46.67		56.66	MHz
$t_{sk(o)}$	Output Skew; NOTE 1, 2			15	ps	
$f_{jit}(\emptyset)$	RMS Phase Jitter (Random); NOTE 3	212.5MHz, (637kHz - 10MHz)		0.65		ps
		159.375MHz, (637kHz - 10MHz)		0.61		ps
		106.25MHz, (637kHz - 10MHz)		0.74		ps
		53.125MHz, (637kHz - 10MHz)		0.64		ps
		187.5MHz, (637kHz - 10MHz)		0.80		ps
$t_R / t_F$	Output Rise/Fall Time	20% to 80%	250		500	ps
odc	Output Duty Cycle	F_SEL[1:0] $\neq$ $\div 3$	48		52	%
		F_SEL[1:0] = $\div 3$	45		55	%

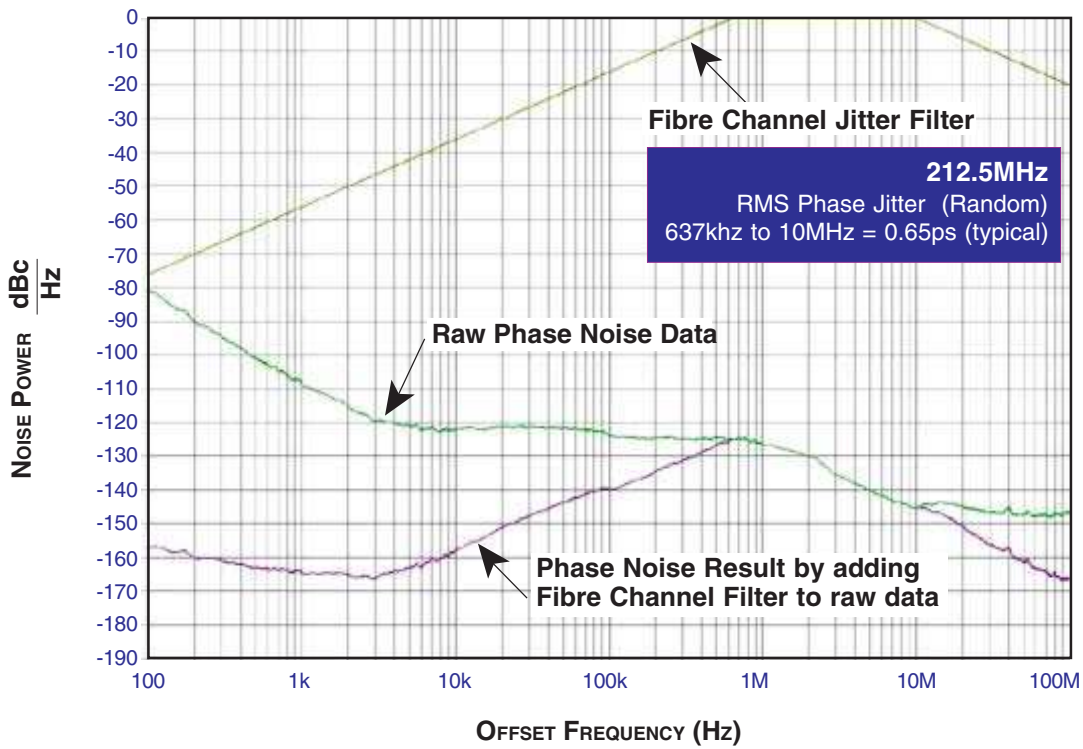
NOTE 1: Defined as skew between outputs at the same supply voltages and with equal load conditions.

Measured at the differential cross points.

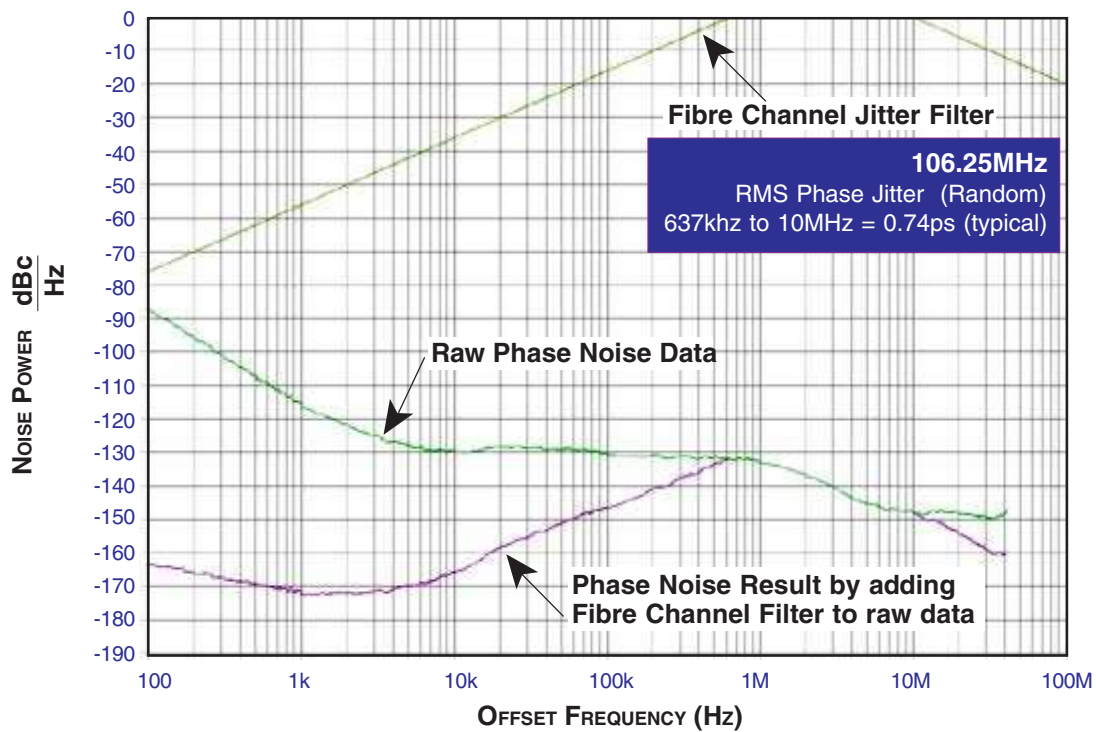
NOTE 2: This parameter is defined in accordance with JEDEC Standard 65.

NOTE 3: Please refer to the Phase Noise Plot.

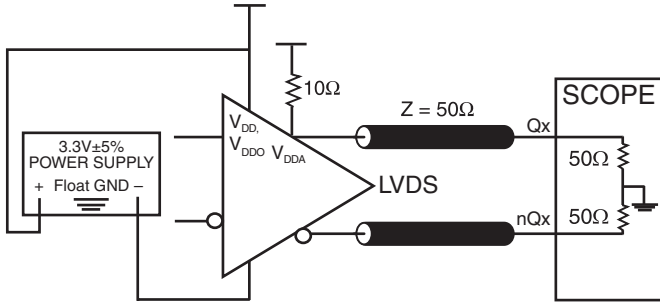
### TYPICAL PHASE NOISE AT 212.5MHz



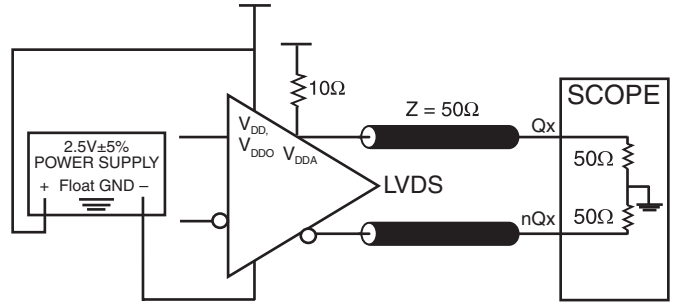
### TYPICAL PHASE NOISE AT 106.25MHz



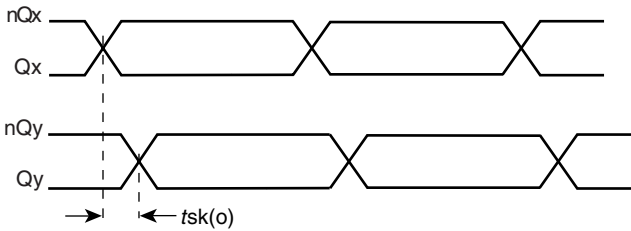
# PARAMETER MEASUREMENT INFORMATION



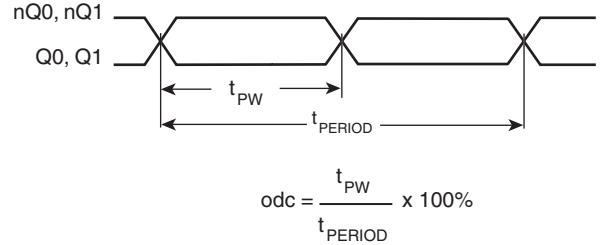
**3.3V CORE/3.3V OUTPUT LOAD AC TEST CIRCUIT**



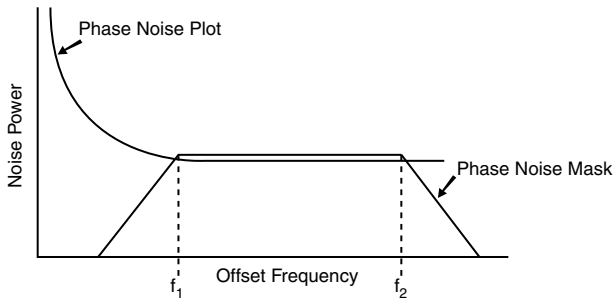
**2.5V CORE/2.5V OUTPUT LOAD AC TEST CIRCUIT**



**OUTPUT SKEW**

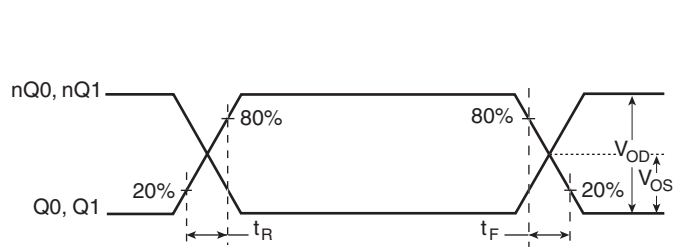


**OUTPUT DUTY CYCLE/PULSE WIDTH/PERIOD**

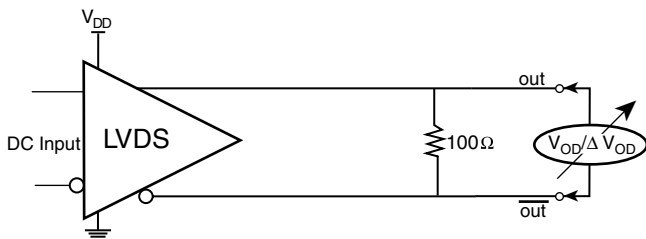


$$\text{RMS Jitter} = \sqrt{\text{Area Under the Masked Phase Noise Plot}}$$

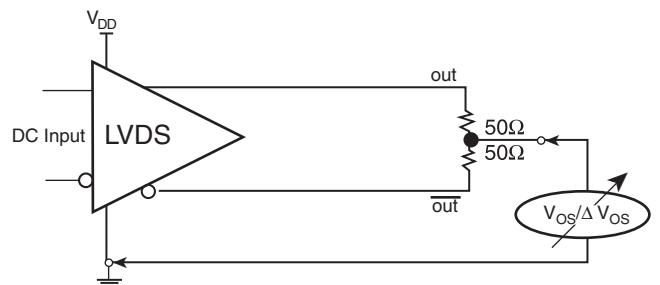
**RMS PHASE JITTER**



**OUTPUT RISE/FALL TIME**



**DIFFERENTIAL OUTPUT VOLTAGE SETUP**



**OFFSET VOLTAGE SETUP**

# APPLICATION INFORMATION

## POWER SUPPLY FILTERING TECHNIQUES

As in any high speed analog circuitry, the power supply pins are vulnerable to random noise. To achieve optimum jitter performance, power supply isolation is required. The ICS844002 provides separate power supplies to isolate any high switching noise from the outputs to the internal PLL.  $V_{DD}$ ,  $V_{DDA}$  and  $V_{DDO}$  should be individually connected to the power supply plane through vias, and  $0.01\mu\text{F}$  bypass capacitors should be used for each pin. *Figure 1* illustrates this for a generic  $V_{CC}$  pin and also shows that  $V_{DDA}$  requires that an additional  $10\Omega$  resistor along with a  $10\mu\text{F}$  bypass capacitor be connected to the  $V_{DDA}$  pin.

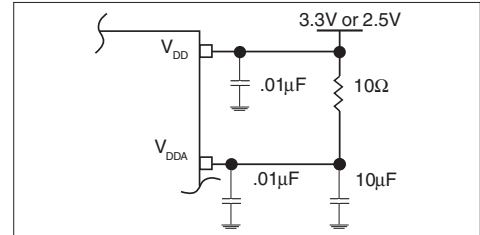


FIGURE 1. POWER SUPPLY FILTERING

## CRYSTAL INPUT INTERFACE

The ICS844002 has been characterized with  $18\text{pF}$  parallel resonant crystals. The capacitor values shown in *Figure 2* below

were determined using a  $26.5625\text{MHz}$ ,  $18\text{pF}$  parallel resonant crystal and were chosen to minimize the ppm error.

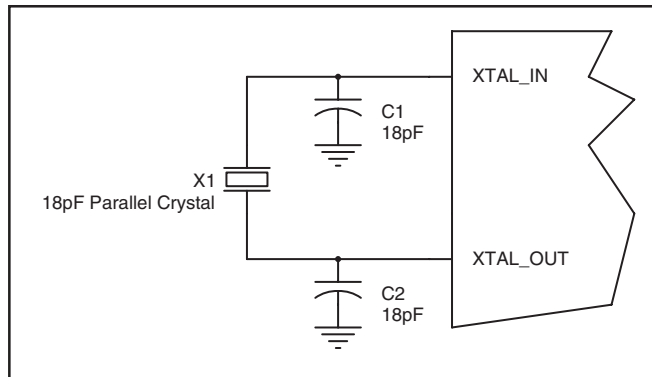


FIGURE 2. CRYSTAL INPUT INTERFACE

## LVC MOS TO XTAL INTERFACE

The XTAL\_IN input can accept a single-ended LVC MOS signal through an AC coupling capacitor. A general interface diagram is shown in *Figure 3*. The XTAL\_OUT pin can be left floating. The input edge rate can be as slow as 10ns. For LVC MOS inputs, it is recommended that the amplitude be reduced from full swing to half swing in order to prevent signal interference with the power rail and to reduce noise. This configuration requires that the output

impedance of the driver ( $R_o$ ) plus the series resistance ( $R_s$ ) equals the transmission line impedance. In addition, matched termination at the crystal input will attenuate the signal in half. This can be done in one of two ways. First,  $R_1$  and  $R_2$  in parallel should equal the transmission line impedance. For most  $50\Omega$  applications,  $R_1$  and  $R_2$  can be  $100\Omega$ . This can also be accomplished by removing  $R_1$  and making  $R_2$   $50\Omega$ .

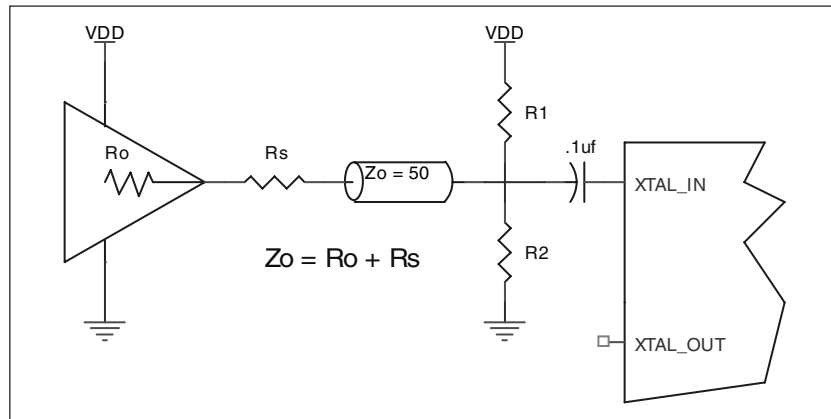


FIGURE 3. GENERAL DIAGRAM FOR LVC MOS DRIVER TO XTAL INPUT INTERFACE

## RECOMMENDATIONS FOR UNUSED INPUT AND OUTPUT PINS

### INPUTS:

#### CRYSTAL INPUTS

For applications not requiring the use of the crystal oscillator input, both XTAL\_IN and XTAL\_OUT can be left floating. Though not required, but for additional protection, a  $1k\Omega$  resistor can be tied from XTAL\_IN to ground.

#### REF\_CLK INPUT

For applications not requiring the use of the reference clock, it can be left floating. Though not required, but for additional protection, a  $1k\Omega$  resistor can be tied from the REF\_CLK to ground.

#### LVC MOS CONTROL PINS

All control pins have internal pulldowns; additional resistance is not required but can be added for additional protection. A  $1k\Omega$  resistor can be used.

### OUTPUTS:

#### LVDS OUTPUTS

All unused LVDS output pairs can be either left floating or terminated with  $100\Omega$  across. If they are left floating, we recommend that there is no trace attached.

### 3.3V, 2.5V LVDS DRIVER TERMINATION

A general LVDS interface is shown in *Figure 4*. In a 100Ω differential transmission line environment, LVDS drivers require a matched load termination of 100Ω across near

the receiver input. For a multiple LVDS outputs buffer, if only partial outputs are used, it is recommended to terminate the unused outputs.

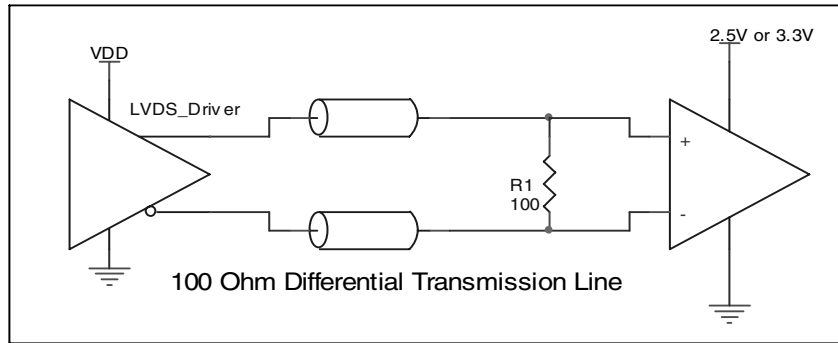


FIGURE 4. TYPICAL LVDS DRIVER TERMINATION

# POWER CONSIDERATIONS

This section provides information on power dissipation and junction temperature for the ICS844002. Equations and example calculations are also provided.

## 1. Power Dissipation.

The total power dissipation for the ICS844002 is the sum of the core power plus the power dissipated in the load(s). The following is the power dissipation for  $V_{DD} = 3.3V + 5\% = 3.465V$ , which gives worst case results.

- Power (core)<sub>MAX</sub> =  $V_{DD\_MAX} * (I_{DD\_MAX} + I_{DDA\_MAX}) = 3.465V * (105mA + 12mA) = 405.4mW$
- Power (outputs)<sub>MAX</sub> =  $V_{DDO\_MAX} * I_{DDO\_MAX} = 3.465V * 120mA = 415.8mW$

$$\text{Total Power}_{MAX} = 405.4mW + 415.8mW = 821.2mW$$

## 2. Junction Temperature.

Junction temperature,  $T_j$ , is the temperature at the junction of the bond wire and bond pad and directly affects the reliability of the device. The maximum recommended junction temperature for HiPerClockS™ devices is 125°C.

The equation for  $T_j$  is as follows:  $T_j = \theta_{JA} * Pd\_total + T_A$

$T_j$  = Junction Temperature

$\theta_{JA}$  = Junction-to-Ambient Thermal Resistance

$Pd\_total$  = Total Device Power Dissipation (example calculation is in section 1 above)

$T_A$  = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance  $\theta_{JA}$  must be used. Assuming a moderate air flow of 200 linear feet per minute and a multi-layer board, the appropriate value is 66.6°C/W per Table 6 below.

Therefore,  $T_j$  for an ambient temperature of 70°C with all outputs switching is:

$$70^\circ\text{C} + 0.821\text{W} * 66.6^\circ\text{C/W} = 124.6^\circ\text{C}. \text{ This is below the limit of } 125^\circ\text{C}.$$

This calculation is only an example.  $T_j$  will obviously vary depending on the number of loaded outputs, supply voltage, air flow, and the type of board (single layer or multi-layer).

**TABLE 6. THERMAL RESISTANCE  $\theta_{JA}$  FOR 20-LEAD TSSOP, FORCED CONVECTION**

$\theta_{JA}$ by Velocity (Linear Feet per Minute)			
	0	200	500
Single-Layer PCB, JEDEC Standard Test Boards	114.5°C/W	98.0°C/W	88.0°C/W
Multi-Layer PCB, JEDEC Standard Test Boards	73.2°C/W	66.6°C/W	63.5°C/W

**NOTE:** Most modern PCB designs use multi-layered boards. The data in the second row pertains to most designs.

# RELIABILITY INFORMATION

TABLE 7.  $\theta_{JA}$  VS. AIR FLOW TABLE FOR 20 LEAD TSSOP

	$\theta_{JA}$ by Velocity (Linear Feet per Minute)		
	0	200	500
Single-Layer PCB, JEDEC Standard Test Boards	114.5°C/W	98.0°C/W	88.0°C/W
Multi-Layer PCB, JEDEC Standard Test Boards	73.2°C/W	66.6°C/W	63.5°C/W

**NOTE:** Most modern PCB designs use multi-layered boards. The data in the second row pertains to most designs.

## TRANSISTOR COUNT

The transistor count for ICS844002 is: 2914

PACKAGE OUTLINE - G SUFFIX FOR 20 LEAD TSSOP

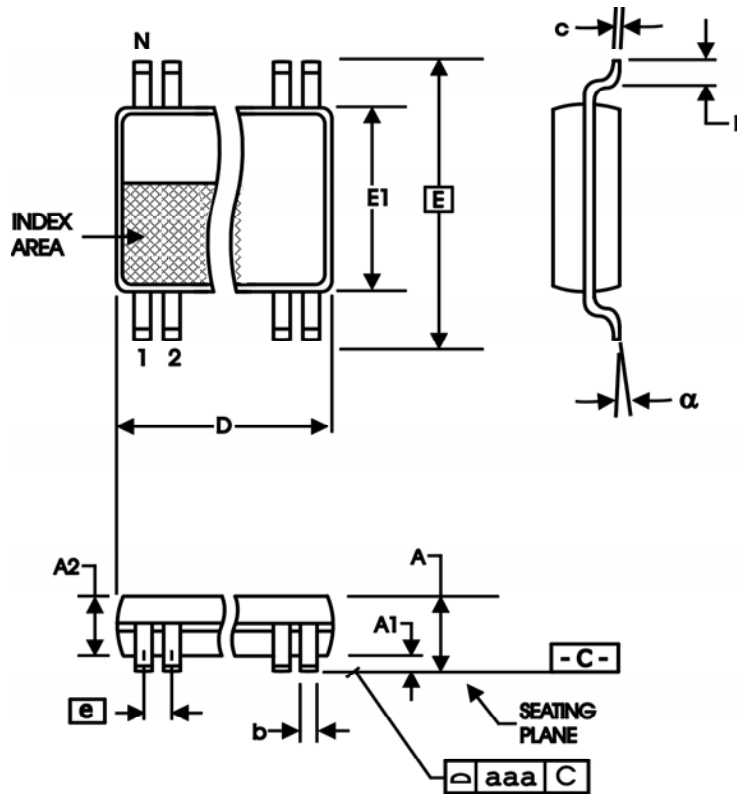


TABLE 8. PACKAGE DIMENSIONS

SYMBOL	Millimeters	
	MIN	MAX
N	20	
A	--	1.20
A1	0.05	0.15
A2	0.80	1.05
b	0.19	0.30
c	0.09	0.20
D	6.40	6.60
E	6.40 BASIC	
E1	4.30	4.50
e	0.65 BASIC	
L	0.45	0.75
alpha	0°	8°
aaa	--	0.10

Reference Document: JEDEC Publication 95, MO-153

**TABLE 9. ORDERING INFORMATION**

<b>Part/Order Number</b>	<b>Marking</b>	<b>Package</b>	<b>Shipping Packaging</b>	<b>Temperature</b>
844002AG	ICS844002AG	20 Lead TSSOP	tube	0°C to 70°C
844002AGT	ICS844002AG	20 Lead TSSOP	2500 tape & reel	0°C to 70°C
844002AGLF	ICS844002AGL	20 Lead "Lead-Free" TSSOP	tube	0°C to 70°C
844002AGLFT	ICS844002AGL	20 Lead "Lead-Free" TSSOP	2500 tape & reel	0°C to 70°C

NOTE: Parts that are ordered with an "LF" suffix to the part number are the Pb-Free configuration and are RoHS compliant.

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**REVISION HISTORY SHEET**

<b>Rev</b>	<b>Table</b>	<b>Page</b>	<b>Description of Change</b>	<b>Date</b>
A		6	Added Phase Noise Plots.	7/24/06
A		11	Power Consideraitons - corrected sentence after the Tj calculation.	1/19/07
A	T1	1 2	Pin Assignment - corrected Pin 16 from V <sub>DD</sub> to nc. Pin Description Table - deleted number 16 from VDD row and added row Pin 16 as a "nc".	9/28/07
B	T3A, T3B T5A, T5B	3 5 7 8 8	Power Supply DC Characteristics Tables - changed VDDA max. from 3.465V to V <sub>DD</sub> . AC Characteristics Tables - corrected NOTE 1. Corrected Output Rise/Fall Time Diagram. Updated Power Supply Filtering Techniqie paragraph text. Corrected Crystal Input Interface Diagram from 33/27p to 18/18p.	7/2/08

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